

# ARMATURE WINDINGS

OF

# ELECTRIC MACHINES

BY

## H. F. PARSHALL

MEMBER AMERICAN INSTITUTE ELECTRICAL ENGINEERS, MEMBER INSTITUTION ELECTRICAL ENGINEERS GREAT BRITAIN, MEMBER AMERICAN SOCIETY OF MECHANICAL ENGINEERS, ETC.

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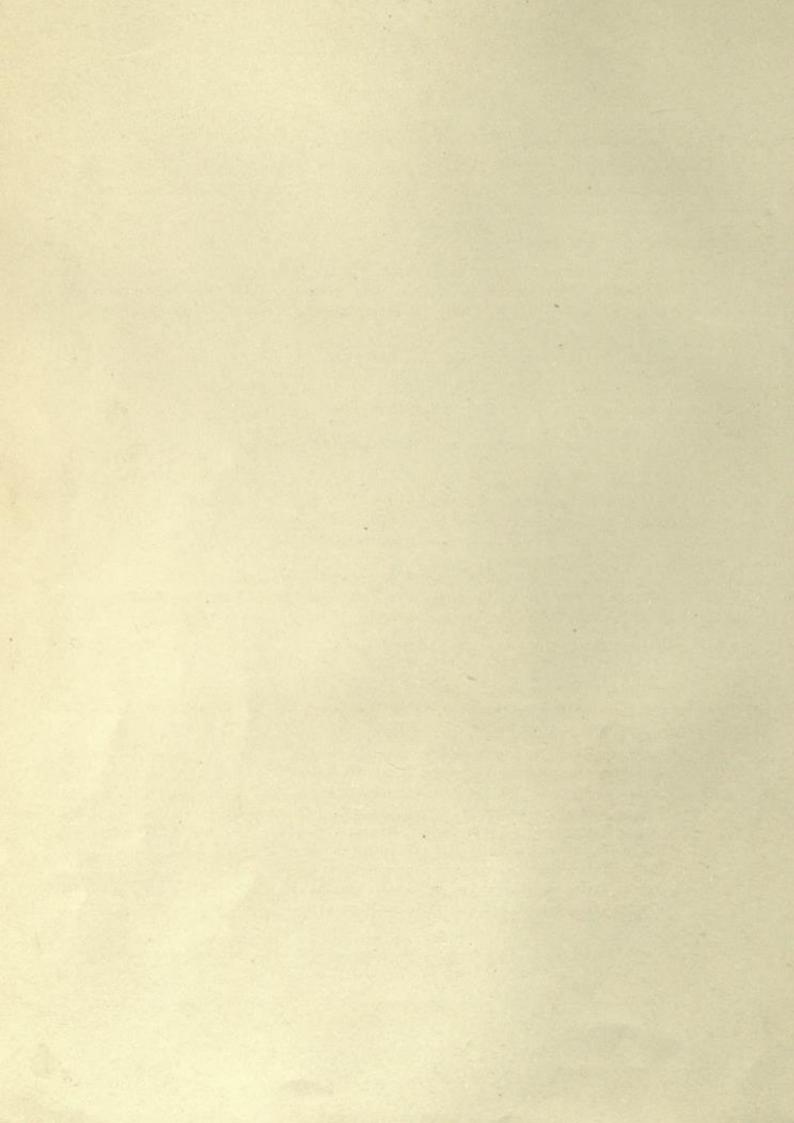
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45 Multipolar drum - Two-circuit, single winding .		8	5	2-0	= y.	Cross-connected	commutator	PAGE 97
46 Multipolar drum - Two-circuit, single winding .		8	4	8	5 & 7	Cross-connected	commutator	98
47 Multipolar drum - Two-circuit, single winding .		8	5	6	7 & 21	Cross-connected	commutator	101
48 Multipolar drum - Two-circuit, single winding .		8	50	2	7 & 19	Cross-connected	The state of the s	102
49 Multipolar drum - Two-circuit, single winding - F	our-	pole wire-	wound	armatur				105
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						7.77		
			No	of poles	Pitch = y	No. of conduc- tors = C.	No. of commu- tator segments.	
50. — Multipolar drum — Two-circuit, single winding .				6	13	80	80	106
51. — Multipolar drum — Two-circuit, single winding .				6	7	44	66	109
52. — Multipolar drum — Two-circuit, single winding .				8	5	42	42	110
53. — Multipolar drum — Two-circuit, single winding .				8	5	42	84	113
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			No. of poles	No. of conductors = C.	No. of windings = m.	Re-entrancy.	Pitch-y.	
54 Multipolar drum - Two-circuit, singly re-entrant, do	ouble	winding,	4	32	2	0	7	115
55 Multipolar drum - Two-circuit, singly re-entrant, do	ouble	winding,	4	32	2	0	7	116
56 Multipolar drum - Two-circuit, singly re-entrant, to	riple	winding .	4	70	3	@	15 & 17	119
57 Multipolar drum - Two-circuit, triply re-entrant, tri	iple :	winding .	4	66	3	000	15	120
58 Multipolar drum - Two-circuit, singly re-entrant, de	onble	winding,	6	58	2	0	9	123
59 Multipolar drum - Two-circuit, doubly re-entrant, de				52	2	00	7 & 9	124
60 Multipolar drum - Two-circuit, triply re-entrant, tri		THE OWNER OF THE OWNER OWNER OF THE OWNER OWNE	6	60	3	000	9	127
61 Multipolar drum - Two-circuit, singly re-entrant, tr	A 11.00		6	54	3	6	7 & 9	128
62 Multipolar drum - Two-circuit, triply re-entrant, tr				78	3	000	11 & 13	131
63 Multipolar drum - Two-circuit, singly re-entrar	7						-	

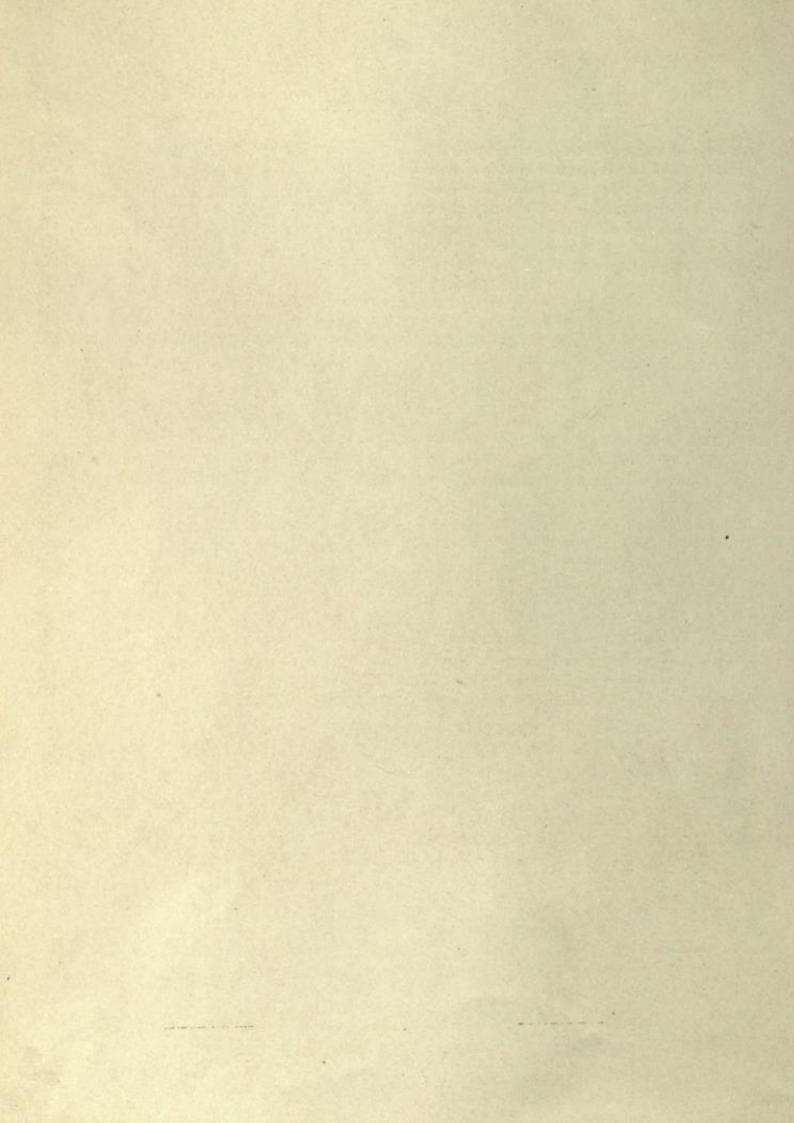
54. — Multipolar drum — Two-circuit, singly re-entrant, double winding,	4	32	2	0	7	115
55. — Multipolar drum — Two-circuit, singly re-entrant, double winding,	4	32	2	0	7	116
56. — Multipolar drum — Two-circuit, singly re-entrant, triple winding .	4	70	3	@	15 & 17	119
57. — Multipolar drum — Two-circuit, triply re-entrant, triple winding .	4	66	3	000	15	120
58 Multipolar drum - Two-circuit, singly re-entrant, double winding,	6	58	2	0	9	123
59 Multipolar drum - Two-circuit, doubly re-entrant, double winding,	6	52	2	00	7 & 9	124
60. — Multipolar drum — Two-circuit, triply re-entrant, triple winding .	6	60	3	000	9	127
61 Multipolar drum - Two-circuit, singly re-entrant, triple winding .	6	54	3	@	7 & 9	128
62 Multipolar drum - Two-circuit, triply re-entrant, triple winding .	6	78	3	000	11 & 13	131
63 Multipolar drum - Two-circuit, singly re-entrant, quadruple						
winding	6	50	4	000	7	132
64 Multipolar drum - Two-circuit, quadruply re-entrant, quadruple						
winding	6	56	4	0000	7 & 9	135
65 Multipolar drum - Two-circuit, doubly re-entrant, quadruple						
winding	6	68	4	00	9 & 11	136
66. — Multipolar drum — Two-circuit, quadruply re-entrant, quadruple						
winding ,	6	80	4	0000	11 & 13	139
67 Multipolar drum - Two-circuit, quadruply re-entrant, quadruple						
winding	6	104	4	0000	15 & 17	140
68. — Multipolar drum — Two-circuit, quadruply re-entrant, quadruple						
winding	6	88	4	0000	15 & 17	143
69 Multipolar drum - Two-circuit, triply re-entrant, sextuple winding,	6	66	6	000	9	144
70 Multipolar drum - Two-circuit, doubly re-entrant, sextuple						
winding	6	72	6	@@	9 & 11	147
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#### LIST OF DIAGRAMS.

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74. — Multipolar drum — Two-circuit, singly re-entrant, double winding		84	2	0	11	155
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The present treatise is the outcome of an investigation made a number of years ago, before the principles of the armature winding of multipolar commutating dynamos were generally understood by electricians. At that time it appeared that the demand for dynamos of greater current output could only be met satisfactorily by dynamos of the multipolar type, since with bipolars beyond a certain output the number of commutator segments compatible with freedom from sparking was found to be incompatible with the maximum armature reaction which experience has shown to be permissible. After some study it was concluded the only feature of the multipolar dynamo requiring special study was that of the armature windings.

A considerable number of diagrams were prepared and classified; the advantages and disadvantages of each, and the comparative fitness of these windings for different purposes, noted. Inasmuch as it was found convenient to refer to this data frequently, and on account of the comparative inaccessibility of such information when in the form of notes, we decided that it would be a great convenience to electricians generally if our notes were published in book form. We therefore proceeded to do this; but owing to the intervention of certain circumstances contingent to our position in an industrial concern, it became necessary to lay aside this work until those competent to judge of its nature should feel able to permit us to proceed as we had wished. The delay has not been disadvantageous, since in the meantime we have not laid the work aside; on the contrary, we have made a study of the properties of a number of the more important windings, so that the original manuscript has been largely added to.

In the section on continuous-current armature windings our endeavor has been to include only those windings that possess some practical merit, and we have frequently pointed out the advantages and disadvantages peculiar to certain classes of windings. The thought will probably occur to the reader, which one of these windings should be selected for a given voltage after the number of poles and the magnitude of the magnetic flux at the poles have been assigned a proper value. We cannot point out the fitness of each winding for a given purpose, since this is more or less dependent upon the magnetic characteristics peculiar to any particular design. Thus in some machines of particularly good characteristics two-circuit windings have been used in the generation of comparatively large currents with some success, when had the magnetic characteristics of the dynamos been ordinarily good, the use of the two-circuit winding would have been attended with results entirely unsatisfactory.

In general, we may state, the type of winding should be determined with reference to the magnitude of the current to be generated. Any deviation from a perfectly symmetrical arrangement of the armature conductors should be inversely proportional to the magnitude of the currents to be generated. When the currents to be generated are large, the coils should be similarly situated with respect to each other, and should all have the same resistance and inductance. It has been frequently found that when the conductors are dissimilarly situated with respect to each other or to any other body that can affect the armature conductors inductively, the wearing away of the commutator is uneven, the trouble increasing more and more as the currents in the conductors are increased, or the resistance of the collecting brushes diminished. Especially in armatures in which there are more than two coils in a slot this uneven wearing away of the commutator has been noticed. In this case the coils are of slightly unequal area, due to the progression of the winding from slot to slot.

In gramme windings the lack of symmetry may be due to some of the coils being longer than the others, or carried near the spider arms.

It may, therefore, be stated generally that when a given result has to be obtained without experimenting, such windings as these are to be avoided when the currents in the conductors have to be of any considerable magnitude.

The utility of the double, triple, and quadruple windings shown and described depends very largely upon the maximum are upon the commutator over which uniform contact resistance can be obtained. With the thickness of segments now common in practice, only double and triple windings appear to be of practical value, since, in general, brushes cannot be relied upon to maintain a uniform contact over an arc of much more than three-quarters of an inch in width. When the width of the brush has to exceed this amount, it is found that it bridges imperfectly from commutator bar to commutator bar in the same winding, thereby causing sparking.

A feature peculiar to these windings, as well as to some of the two-circuit single windings, is that the voltage between adjacent commutator sections is affected by the angular distance between the different sets of collecting brushes. With some of these windings the voltage between adjacent commutator sections varies simply according to the field strength when the angle between the different sets of brushes corresponds to the angle between the centers of the poles. In other windings the voltage between adjacent commutator sections varies by jumps, but may be made to vary according to the field strength by slightly varying the position of some one set of brushes with respect to the other sets. This feature of the different windings is a subject for special investigation, and is of more or less importance, according to the nature of the winding and the average voltage between commutator bars.

We have frequently made mention of the number of slots. With respect to slotted armatures in general, it is to be remembered that an additional condition to that for smooth-core armatures has to be fulfilled; i.e. the total number of the conductors to suit the equations for re-entrancy has to be divisible by the number of conductors possible to place in a slot, this number being dependent upon the number of poles. The number of conductors permissible per slot for two-circuit windings for different numbers of poles is shown in a table.

We have omitted any reference to mechanical details of construction of armature windings, since these permit of great variety, without in any way modifying the results. Further, they are a part of the stock in trade of the electrical manufacturer.

The drum windings considered are principally those in which the end connections are interchangeable, and

are in the form of evolutes, as in the Eickemeyer and Hopkinson windings, description of which will be found in Weymouth's "Drum Armatures and Commutators" ("The Electrician" Printing and Publishing Company, London, 1893). In general, such windings possess the advantages that all coils are of equal inductance and resistance, are equally accessible, have equal radiating surfaces, and are most easily repaired. When a coil consists of a number of conductors, bound together so as to be considered a single unit mechanically, it is so considered in the text, and in the formulæ for the arrangement of conductors.

These windings appear to have been invented by Bollmann, Desroziers, Fritsche, Pischon, Eickemeyer, and others; but inasmuch as it is a disputed question as to which of these inventors has the right to claim priority, and as there may be more or less litigation before the question is settled, we have considered it best to omit all discussion as to who may have invented any of the windings. Where with a winding is given the name of a supposed inventor, it is simply because that winding has been known under that name, and not because the writers possess any special evidence to show by whom the winding was invented. After the possibility of litigation has ceased we hope to do justice to all inventors concerned, giving to each his proper proportion of credit for the work he has done.

We believe that the tables on drum windings are a feature that should meet with especial favor, since after the number of conductors required for a given type of winding has been determined, the proper pitches for any style of winding can be found in the tables. Further, by referring opposite to this number of conductors in the different tables it may be ascertained at a glance whether, by slightly changing the end connections, the winding may be adapted to some other voltage. Such features, peculiar to certain numbers of conductors, are frequently in practice of the greatest importance. As a practical example take the following case: In a six-pole machine with 104 armature conductors, the winding may be connected for a two-circuit single winding by making the pitch 17 on each end, or for a two-circuit, doubly re-entrant double winding, by making the pitch 17 on one end and 19 on the other; this second arrangement being suitable for the same watt output as the first, at one-half the voltage.

In the section on alternate-current armature windings are included a number of windings that have now only a limited application in practice, as it is thought that, on account of the very limited literature on this subject, a description of all windings of any practical use will be appreciated.

With respect to the work in general, we should be glad to receive the suggestions and criticisms of all who are interested in this subject.

The following articles on armature windings have been consulted in the preparation of this book, and are mentioned here for reference: —

ARNOLD — Die Ankerwicklungen der Gleichstrom-Dynamomaschinen. Berlin, 1891.

Fritsche — Die Gleichstrom-Dynamomaschine. Berlin, 1889.

Kapp — Practical Electrical Engineering, Vol. II., p. 43. London, 1893.

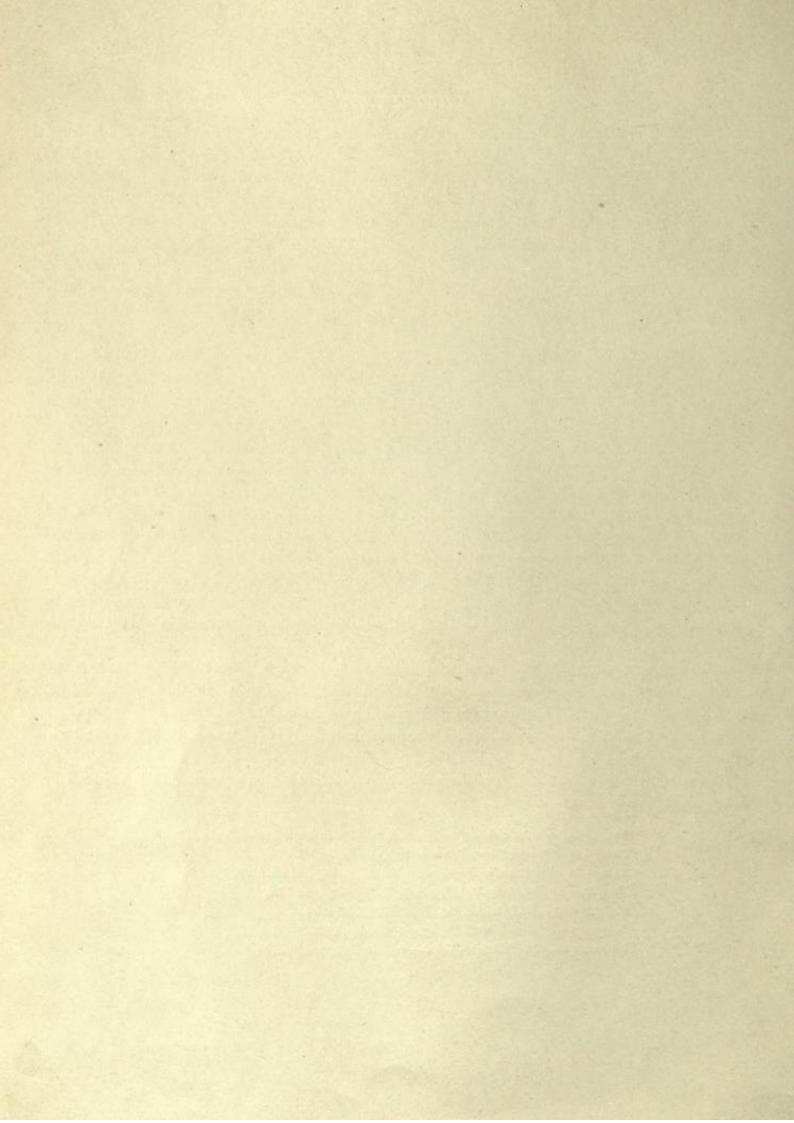
Kittler — Handbuch der Elektrotechnik, Vol. I. Stuttgart, 1892.

Rechniewski — L'Electricien, Vol. V. Jan. 14, 1893 et seq.

Thompson — Dynamo-Electric Machinery. London, 1892.

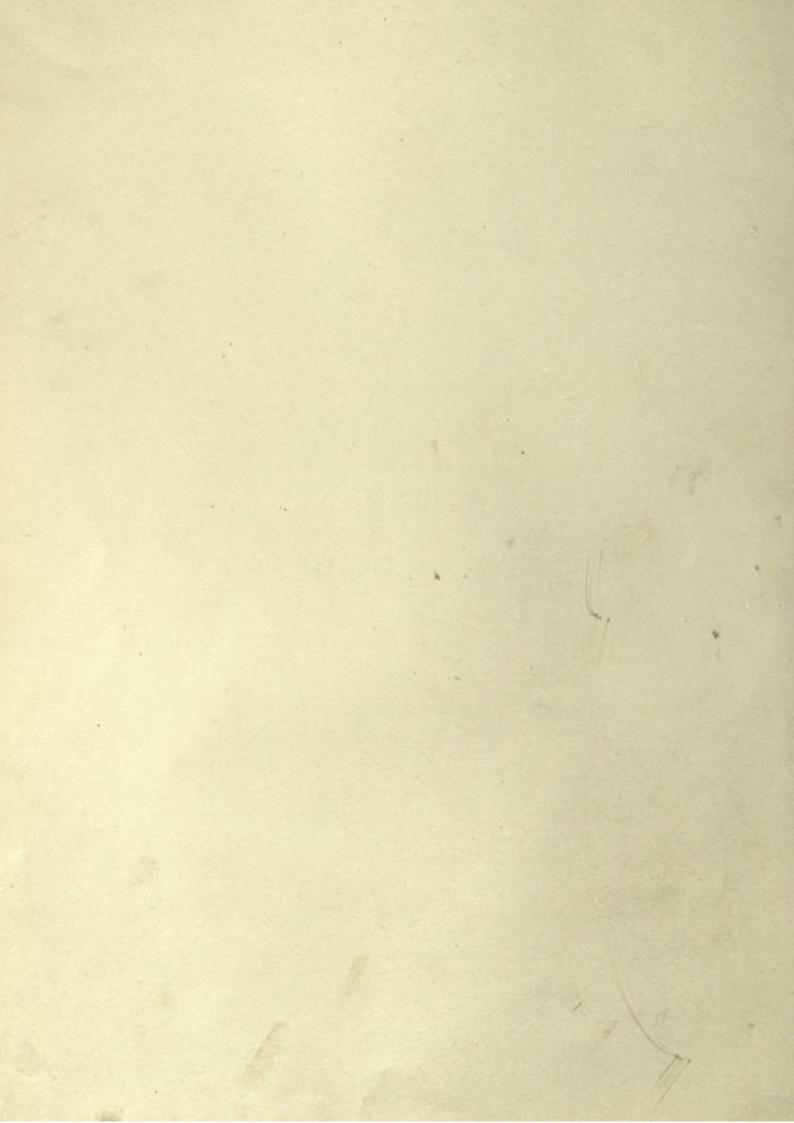
Weymouth — The Electrician, Vol. XXV. Nov. 7 to Dec. 19, 1890.





# PART I.

CONTINUOUS-CURRENT ARMATURE WINDINGS.





#### CHAPTER I.

#### SINGLE-WOUND GRAMME RINGS.

These are the simplest windings in use, and will require only a very few diagrams and explanations. Many complex connections have been proposed, but only such forms will be discussed as are of general practical use.

The plain gramme ring, with a single winding, is shown in Figs. 1 and 2, from which it may be seen that the construction, as far as concerns location of coils, connectors, and commutator segments, is independent of the number of poles. The number of coils should be a multiple of the number of poles in order to maintain

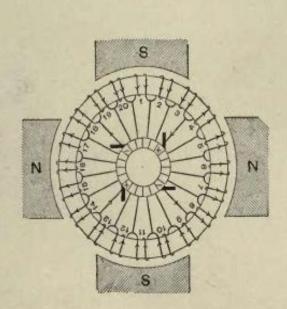


Fig. I
FOUR-CIRCUIT, SINGLE-WINDING.

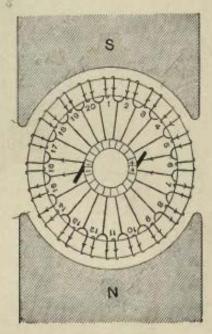


Fig. 2
TWO-CIRCUIT, SINGLE-WINDING.

symmetry among all the branches from brush to brush. The number of commutator segments is equal to the number of coils. It is desirable to minimize the turns per coil, and consequently the inductance of the short-circuited elements, by as large a number of segments as practicable.

A further discussion of these two diagrams would be superfluous, beyond calling attention to the progressive nature of the rise of potential around the ring, whereby the contiguous wires have only the small difference of potential of one turn, making the question of insulation very simple. In cases where it is desirable to use but two brushes in multipolar rings with more than two circuits, the method of cross-connecting, shown in Fig. 3, may be used. The number of commutator segments remains equal to the number of coils. An inspection of the diagram will show that it really consists in connecting in parallel those coils occupying corresponding positions in the various fields.

It would seldom be desirable to utilize this method of connection, except in very small machines, as the use of only one pair of sets of brushes would necessitate lengthening the commutator in order to retain the proper extent of brush contact surfaces.

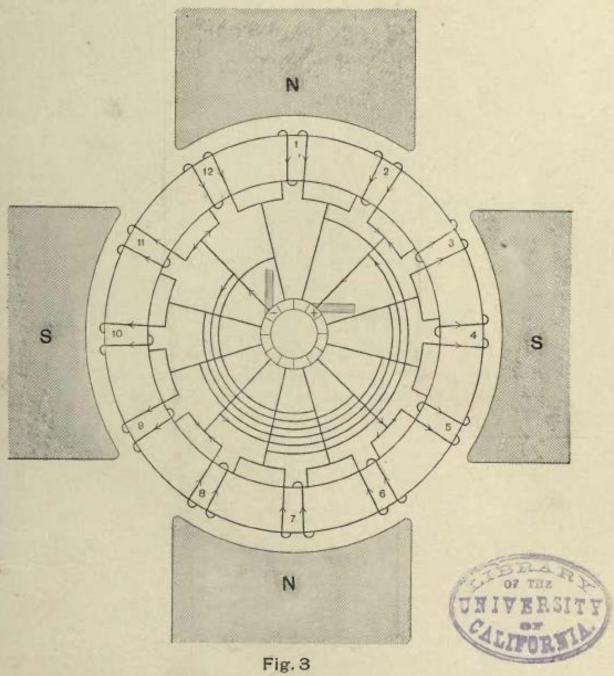


Fig. 3
FOUR CIRCUIT, SINGLE WINDING.

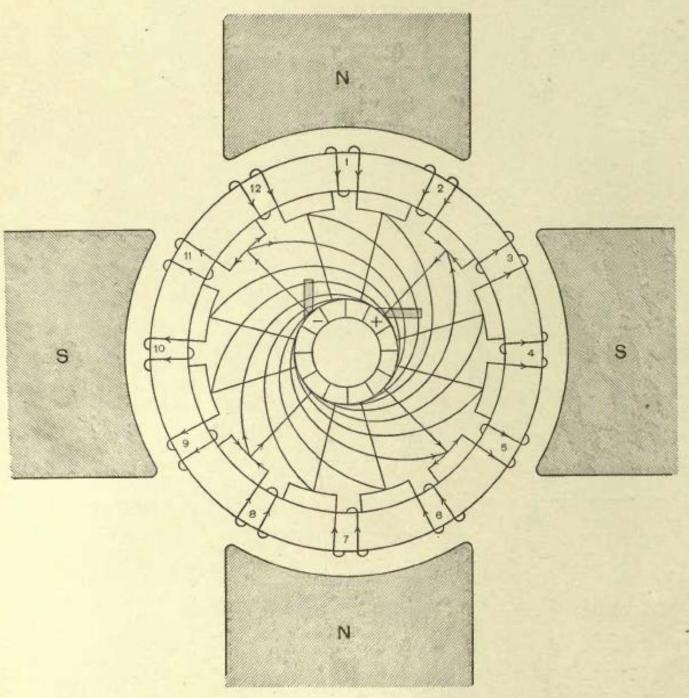


Fig. 4
FOUR CIRCUIT SINGLE WINDING

Figure 4 differs from Fig. 3 only in the use of two crossconnecting leads instead of one. This diagram would sometimes be of advantage, inasmuch as it utilizes the available space more completely and symmetrically. Each crossconnecting conductor could be of smaller cross-section than if only one were used.

Both this and the preceding method have the disadvantage that the two parallel sections have unequal resistance, due to one section having the long cross-connecting leads in series with it, and the other merely the regular short leads to the commutator.

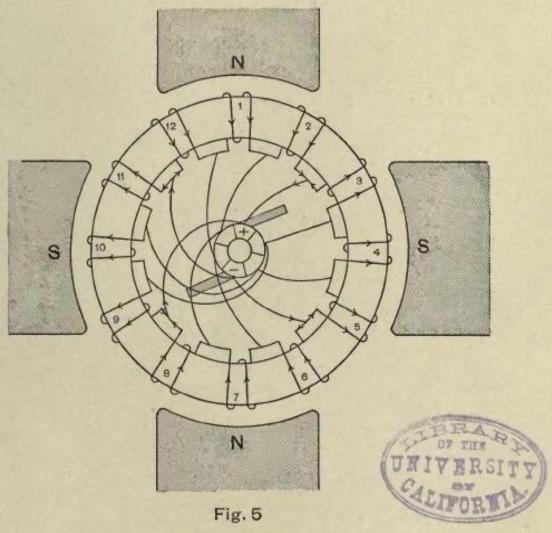
Failure to give due attention to this point often causes serious trouble. Figure 5 gives a winding which is wrong, but which has been given in the treatises of many of the specialists on windings, none of whom, except Herr Arnold, criticise it.

The fault is that the positions of the coils bear such a relation to the positions of their respective commutator segments, that during each revolution of the armature the position given in the figure is the only one in which the brushes are properly placed with regard to the diameter of commutation. In order that the brushes should always be in a position to properly perform their commutative function, they would have to be revolved in a direction opposite to that of the armature, and with a velocity equal to it.

The characteristic of the winding is that it brings together into one segment each pair of cross-connected segments of the previous diagram. As above stated, however, this diagram is worthless, except to call attention to its character, so that the text-books in which it is described shall not be misleading.

See Arnold — Die Ankerwicklungen der Gleichstrom-Dynamomaschinen, Fig. 42.

KITTLER — Handbuch der Elektrotechnik, 1892, Fig. 401 C. FRITSCHE — Die Gleichstrom-Dynamomaschinen, Fig 64.



FOUR CIRCUIT SINGLE WINDING.

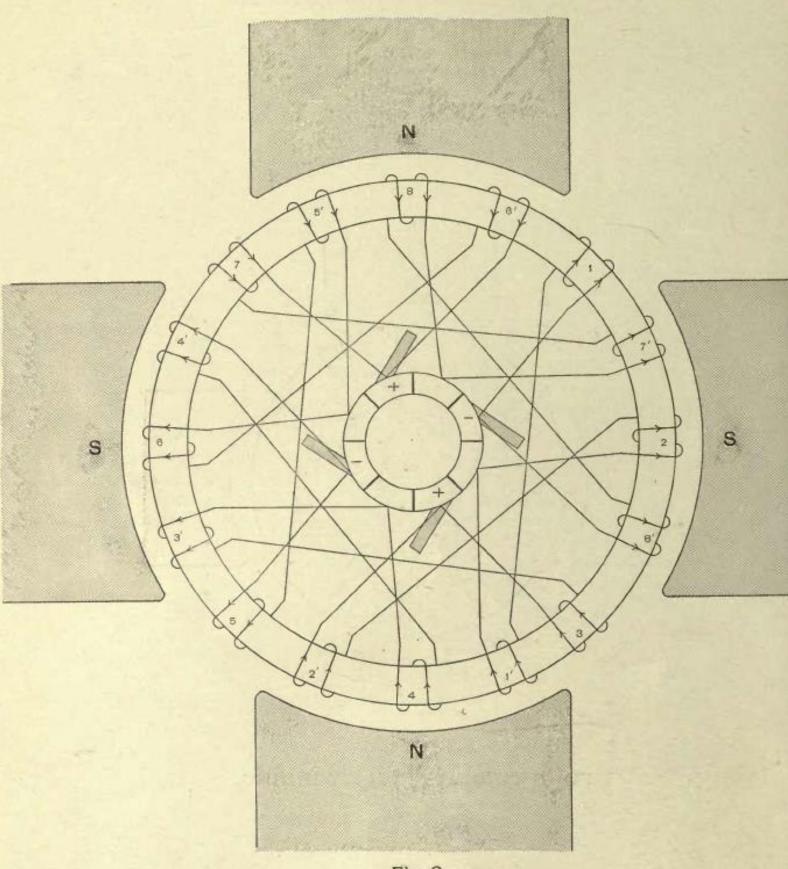


Fig. 6
FOUR CIRCUIT SINGLE WINDING.

In Fig. 6 the number of commutator segments is made equal to half the number of coils by connecting two coils in series between each pair of adjacent segments. The coils so connected in series are situated in adjoining fields of opposite polarity. This winding has the disadvantage that coils at quite different potentials are adjacent, as may be seen by following through the various armature circuits from brush to brush. This increases the difficulty of insulating. The volts per bar also, for the same number of conductors per coil, are twice as high as in the simple gramme ring. If it is necessary, for any reason, to halve the number of bars, it would be preferable to combine two adjacent coils into one, and retain the advantages of the simple gramme ring connection.

But in cases where the shape of the frame necessitates somewhat unequal magnetic circuits, this connection averages up the unequal induction in the various coils, and therefore tends to diminish the sparking which might, with a simple gramme ring in such an unbalanced magnetic system, be considerable.

If s=number of coils, and n=number of poles, then any coil is connected across to one  $\left(\frac{s}{n}\pm 1\right)$  in advance of it, and the two free ends of this pair of coils are connected to adjacent commutator segments.

Figure 7 is merely a step in advance of Fig. 6, and the advantages and disadvantages pointed out in the discussion of Fig. 6 apply in still greater degree to Fig. 7.

It will be seen that the number of commutator segments is reduced to one-fourth of the number of coils by the connecting in series of four coils, one in each field, between two adjacent segments of the commutator.

As in the previous figure, the rule for connecting the coils is to connect each coil to one  $\left(\frac{s}{n} \pm 1\right)$  in advance.

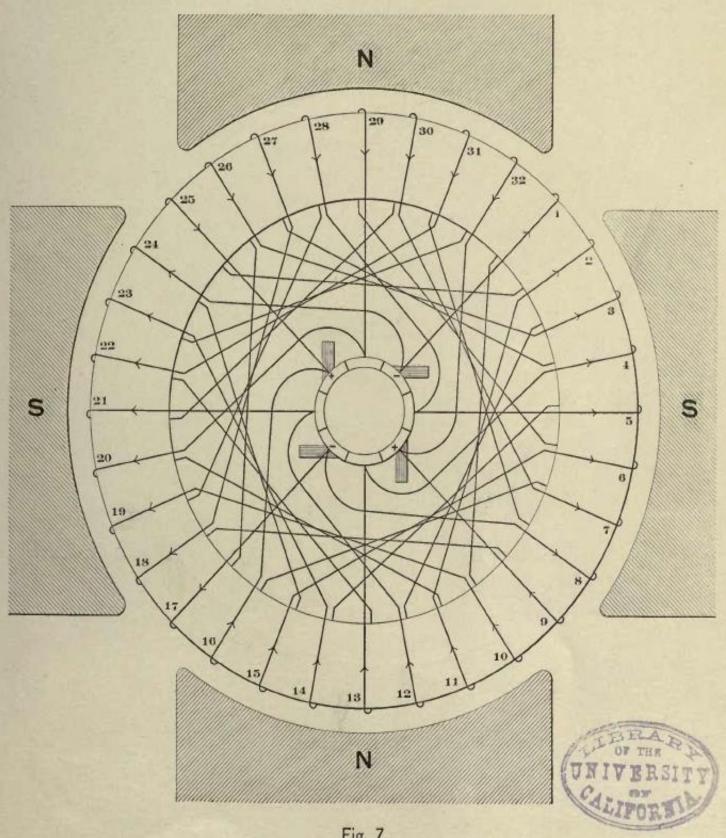
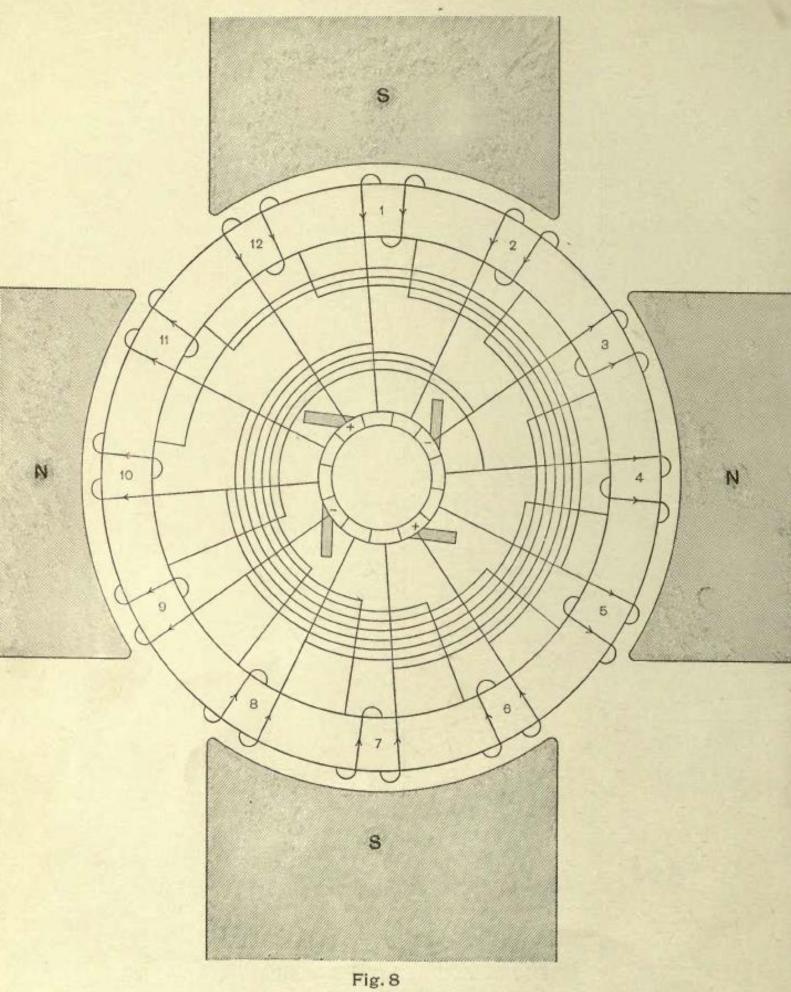


Fig. 7
FOUR CIRCUIT SINGLE WINDING.



FOUR CIRCUIT, SINGLE WINDING.

Figure 8 represents a winding in which the coils of one circuit, from brush to brush, instead of being adjacent to each other, are situated in different fields. For instance, the circuits through the armature in the position shown are,—

$$\rightarrow \begin{array}{c} -\left\{ \begin{array}{ccc} 3 & 10 & 5 \\ 8 & 1 & 6 \end{array} \right\} + \\ -\left\{ \begin{array}{ccc} 2 & 7 & 12 \\ 9 & 4 & 11 \end{array} \right\} + \end{array} \rightarrow$$

It is important to note that when the armature has entered the position in which four coils are short-circuited, the short-circuiting of any coil occurs, not at any one brush, but through the pair of brushes of like polarity. This would enable sparking to be diminished by connecting the two positive brushes together through a suitable resistance (ohmic or inductive), and leading off to the load from the middle point of this resistance. The magnitude of the resistance, if ohmic, would be limited only by the permissible loss therein. High resistance leads to the commutator, and high-resistance brushes have been used with considerable success; but in both of these cases heat has to be developed in undesirable localities. But in the above method of connection, the insertion of this resistance externally to the brushes will not increase the heating of the machine. This resistance is also so located that it could be adjusted in experimental work, and the difference in sparking noted by having a short-circuiting switch shunted around the resistance.

Another advantage of this winding is that pointed out in the remarks on Fig. 6, that in cases where the shape of the frame necessitates somewhat unequal magnetic circuits, this connection will average up the unequal induction in the various coils, and thereby diminish the sparking that would otherwise occur.

# CHAPTER II.

### DOUBLE-WOUND GRAMME RINGS.

FIGURE 9 and the immediately following diagrams relate to a class of very great importance, which are known as double, triple, quadruple, etc., windings.

Very satisfactory results have been attained by the use of windings of this class. The most important advantage of the double winding is that the current is commutated at two different parts of the bearing surface of the brush; each independent volume of current being, therefore, only one-half of what it would be for a single winding. The importance of this feature has in practice been found to be very great.

Another important feature of this winding is that the successive commutator bars of one winding are not adjacent to each other, but alternate with the bars of the other winding; the two windings being put in parallel by the use of wide brushes. The result is that a section is very unlikely to be short-circuited by dirt or an arc. It also makes a very flexible winding, owing to the readiness with which any number of parallels may be arranged. Thus, in a six-pole field, we may have four, six, eight, etc., parallels.

It is necessary for a double winding that the brush should bear over a surface greater than the width of one segment (plus insulation); for a triple winding, greater than the width of two segments, etc.

In Fig. 9, which represents a two-circuit, doubly re-entrant, double-wound, simple gramme ring, the circuits through the armature are, —

After the armature has revolved through  $\frac{360}{20 \times 2} = 9^{\circ}$ , coils 3 and 8 will be short-circuited, and the circuits through the armature will become,—

Thus it will be seen that there will be a lack of balance between the two windings. First they will be of equal length; after 9° revolution, one will have one less section in series between the brushes; 9° later they will be equal again; and after still another 9° the other winding will have the smaller number of turns. This lack of symmetry will be less apparent as the number of sections is increased, and becomes of very little importance with the large numbers of conductors employed in practical work.

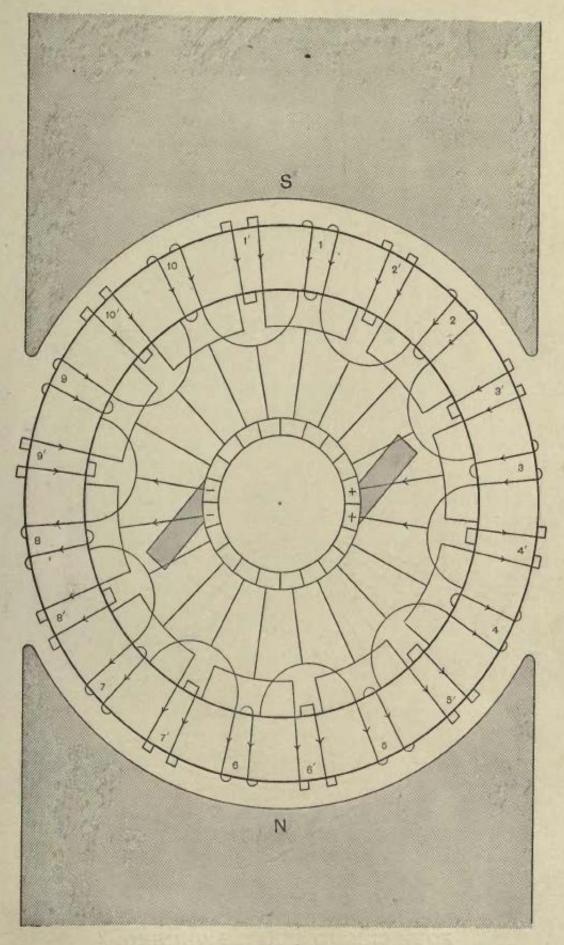
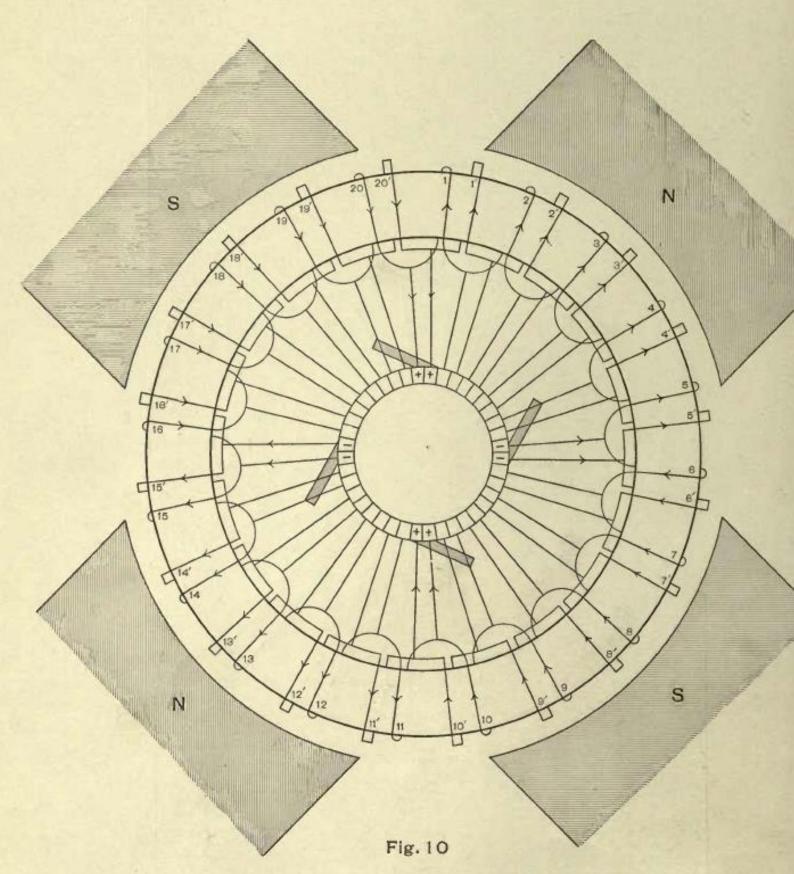


Fig. 9
TWO CIRCUIT DOUBLE WINDING





FOUR CIRCUIT DOUBLE WINDING.

Figure 10 shows a similar winding in a four-pole field. The circuit through the armature in the position shown is,—

$$\rightarrow \begin{bmatrix} -\left\{ \begin{matrix} 16 & 17 & 18 & 19 & 20 \\ 15 & 14 & 13 & 12 & 11 \end{matrix} \right. \\ -\left\{ \begin{matrix} 16' & 17' & 18' & 19' & 20' \end{matrix} \right. \\ +\left\{ \begin{matrix} 15' & 14' & 13' & 12' & 11' \end{matrix} \right. \\ -\left\{ \begin{matrix} 6 & 7 & 8 & 9 & 10 \\ 5 & 4 & 3 & 2 & 1 \end{matrix} \right. \\ -\left\{ \begin{matrix} 6' & 7' & 8' & 9' & 10' \end{matrix} \right. \\ -\left\{ \begin{matrix} 6' & 7' & 8' & 9' & 10' \end{matrix} \right. + \\ -\left\{ \begin{matrix} 6' & 7' & 8' & 9' & 10' \end{matrix} \right. \end{bmatrix} + \end{bmatrix}$$

After turning through  $\frac{360}{40 \times 2} = 4.5^{\circ}$ , coils 15', 20', 5', and 10' will be short-circuited, and the circuits through the armature will be,—

$$\rightarrow \begin{cases} -\left\{ \begin{matrix} 16 & 17 & 18 & 19 & 20 \\ 15 & 14 & 13 & 12 & 11 \end{matrix} \right. \\ -\left\{ \begin{matrix} 16' & 17' & 18' & 19' \\ 14' & 13' & 12' & 11' \end{matrix} \right. \\ -\left\{ \begin{matrix} 6 & 7 & 8 & 9 & 10 \\ 5 & 4 & 3 & 2 & 1 \\ -\left\{ \begin{matrix} 6' & 7' & 8' & 9' \\ 4' & 3' & 2' & 1' \end{matrix} \right. \right. \end{cases} \rightarrow$$

Here can be seen again the lack of symmetry noted in remarks on Fig. 9.

A very useful winding is that shown in Fig. 11. It, also, is a four-circuit double winding. It is one of a class with very interesting properties. It differs from the double winding shown in Fig. 10, in that the two windings are components of one re-entrant system. Any one section is no longer exclusively an element of one of two windings, but changes from one winding to the other four times per revolution, being short-circuited at the neutral point for a brief period at the occurrence of each of these transfers. These features are secured by adding one section to the doubly re-entrant double winding shown in Fig. 10, and, as in that figure, making the connections, not between adjacent sections, but always by passing over one section. The number of sections being odd, it will be seen that after having progressed twice around the ring, all sections will have been passed through, and the winding will have arrived at the other terminal of the section from which it started.

Triple, quadruple, and higher orders of windings may be treated analogously. The circuits through the armature in the position shown in Fig. 11 are,—

Coil 10 is, at this instant, short-circuited. An instant later coil 10 becomes active, and coil 2 becomes short-circuited. The circuits through the armature then become, —

The order in which the various coils will be short-circuited is 10, 2, 15, 7, 20, 12, 4, 17, etc., so that the 21 coils will each have been short-circuited once when the armature shall have revolved through  $\frac{360^{\circ}}{4} = 90^{\circ}$ . Therefore the angular interval between corresponding positions of two successive short circuits is  $\frac{90^{\circ}}{21} = 4.28^{\circ}$ .

Such windings will be designated as singly re-entrant, to distinguish them from others, such as those of Figs. 9 and 10, which are doubly re-entrant.

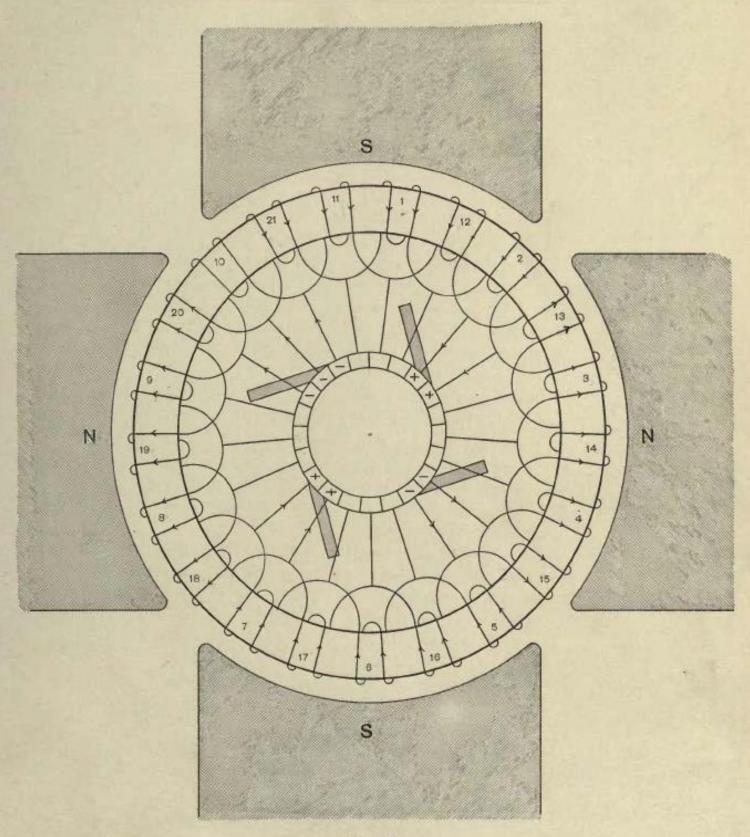


Fig. 11
FOUR CIRCUIT DOUBLE WINDING



All of the windings so far described have as many circuits through the armature as there are pole pieces, and form a class by themselves known as multiple-circuit windings. Four-pole fields have usually been considered, but the modifications of the diagrams and text to apply them to larger numbers of poles, are obvious.

In general, the number of sets of brushes equals the number of poles and the number of circuits through the armature. Different numbers of segments and brushes are due to modifications, and do not affect the underlying character of the windings as a class. Some of these modifications have been described. Others can be worked out as the occasion requires.

Too much importance cannot be attached to the general rule that interpolations and cross-connections are almost always very undesirable.

### CHAPTER III.

### TWO-CIRCUIT, SINGLE-WOUND, MULTIPOLAR RINGS.

The next windings to be considered form a class which, independently of the number of 'poles, have only two circuits through the armature. These are known as two-circuit windings. Such windings possess the practical advantage that the number of conductors, as compared with multiple-circuit windings, is only  $\frac{2}{N}$  as great, hence the space required for insulation is only  $\frac{2}{N}$  as great as with the multiple-circuit windings, in consequence of which the diameter of the armature, or the depth of space occupied by the armature conductors, may be less than with the multiple-circuit windings, thereby diminishing the cost of material.

Further, on account of the lesser number of conductors, the cost of the labor of winding is correspondingly diminished.

In practice, the two-circuit gramme windings have been applied only to armatures of small output, under which condition lack of symmetry of the armature coils with respect to the points of commutation is not particularly objectionable. Only two sets of collecting brushes are necessary for the collection of current; in practice generally but two sets have been used.

In the "short-connection" type of two-circuit gramme windings, the circuits from brush to brush consist of conductors influenced by all the poles, so that the electromotive forces generated in the two circuits are necessarily equal, a feature that may prove advantageous when the depth of air-gap is so small that any slight eccentricity of the armature affects the magnetic flux at the different poles.

In the "long-connection" type of two-circuit gramme winding, the two circuits from brush to brush consist of conductors influenced by only one-half of the poles, so that the electromotive forces generated in the two circuits are unequal, unless the sum of the lines at the poles of the same sign is equal to the sum of the lines at the poles of the opposite sign. In magnetic circuits of ordinarily good design this condition is fulfilled even though the fluxes at the different poles are unequal. So the winding is practically as good as the "short-connection" winding, and possesses certain other advantages stated in the text, that make its use preferable.

For armatures the outputs of which are so great that several sets of collecting brushes are required, these windings possess the same disadvantages as two-circuit drum windings, a discussion of which is to be found under that caption.

<sup>&</sup>lt;sup>1</sup> Called "short-connection" type because coils in adjacent fields are connected together. This distinguishes it from the "long-connection" type, in which coils twice as far apart are connected together.

Figure 12 represents one of the most practicable two-circuit windings for multipolar-ring armatures. It may be designated as the long-connection type of the two-circuit gramme winding, and one of its chief advantages is, that no great differences of potential exist between adjacent coils.

In the figure is shown the case of a four-pole, two-circuit, single-wound, long-connection ring armature. In the position chosen, the circuits through the armature are,—

Coils 3 and 10, in series, are at this instant short-circuited by the negative brush. A little later, coils 7 and 15 will be short-circuited by the positive brush. When this occurs, the negative brush will bear upon the middle of a segment.

The number of commutator segments is equal to the number of coils, and must be odd for armatures with an even number of pairs of poles; but may be odd or even for armatures with an odd number of pairs of poles. The relation that must subsist in two-circuit, multipolar-ring, long-connection windings, between the number of coils (s) and the number of poles (n), is,—

$$s = \frac{n}{2} y \pm 1,$$

where y = pitch. (The pitch is the number of coils to be advanced through in arranging the end connections. In the diagram, for instance, the pitch y = 7, and the end of coil 1 is joined to the beginning of coil 1 + 7 = 8; the end of 8 to the beginning of 8 + 7 = 15; the end of 15 to the beginning of 15 + 7 = 22 (or 7), etc.) Mr. Gisbert Kapp has prepared the following table for two-circuit, multipolar-ring, long-connection windings by substituting numerical values for n in the above formula:—

# TWO-CIRCUIT, MULTIPOLAR-RING, LONG-CONNECTION WINDINGS.

	MACHINE HAS							
The number of coils must be equal to				10 poles 5 y ± 1				

For two-circuit, multipolar-ring machines with long-connection windings, y, the pitch, may be any integer.

(Note that these conditions do not hold for drum windings.)

Mr. Kapp has also prepared the following table, showing the practicable choice of angular distances between brushes in these two-circuit, multipolar windings:—

NUMBER OF POLES.	A	NGULAR DIST	ANCE BETW	EEN BRUSH	es.
2	180				
4	90				
6	60	180			
8	45	135			
10	36	108	180		
12	30	90	150		
14	25.7	77	128	180	
16	22.5	67.5	112	158	
18	20.	60	100	140	180
20	18	54	90	126	162

The smaller possible angles, namely, 20° for 18 poles, and 18° for 20 poles, are in practice too small to be admissible, and are, therefore, not given in the table.

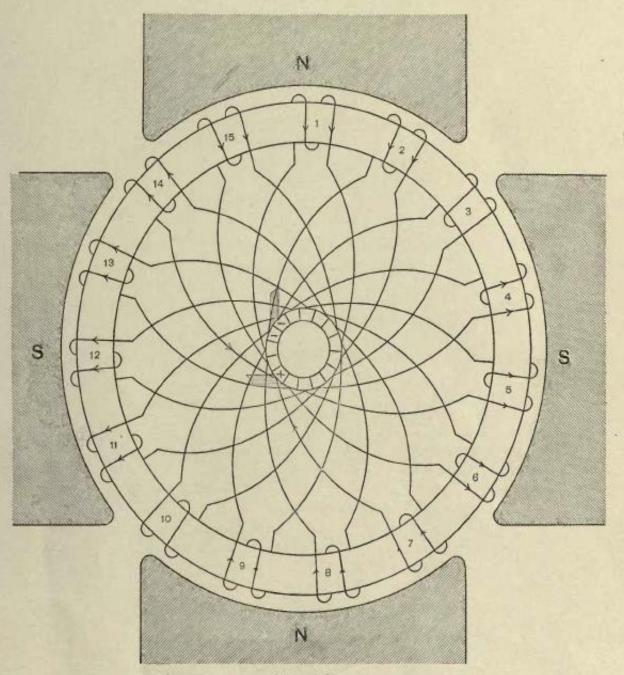
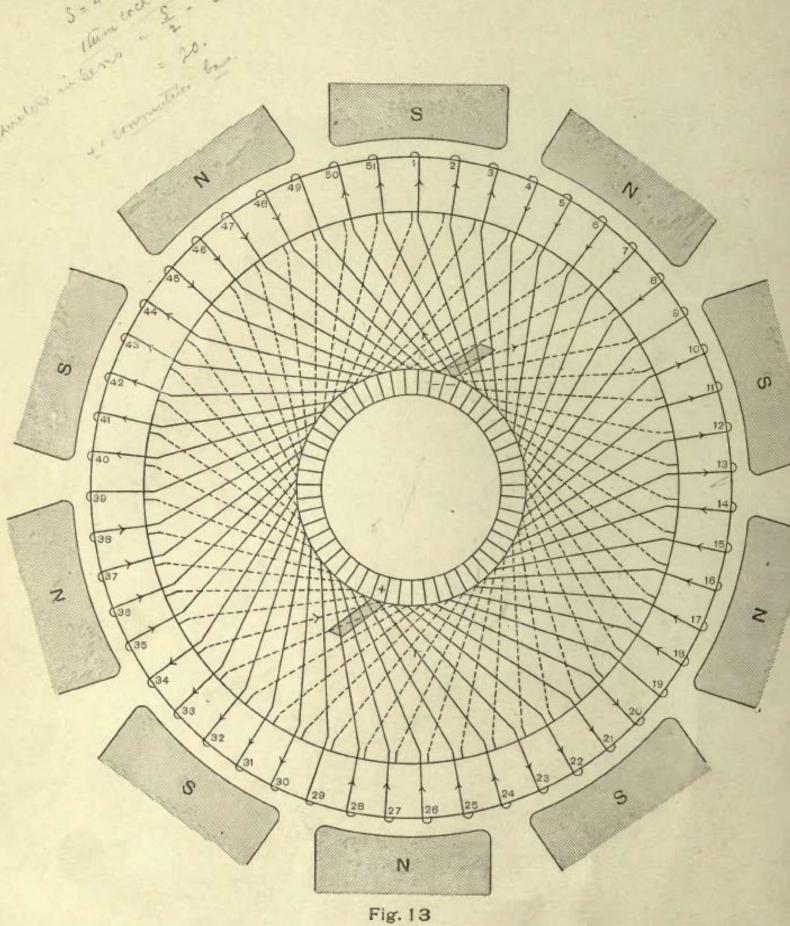


Fig. 12

TWO CIRCUIT, SINGLE WINDING.



TWO CIRCUIT, SINGLE WINDING.

Figure 13 represents a two-circuit, single-wound, longconnection, ten-pole ring armature. Substituting in the formula for the number of coils

$$s = \frac{n}{2} \ y \pm 1$$

the pitch, y = 10, and the number of poles, n = 10, gives  $s = 10 \cdot 10 \pm 1 = 51$  or 49. 51 coils are taken in this case. The end of coil 1 is joined to the beginning of coil 1 + 10 = 11; the end of 11 to the beginning of 21, etc.

The brushes are shown 180° apart, and at the position given the negative brush short-circuits the coils 9, 19, 29, 39, and 49. The circuits through the armature are,—

This diagram and table show very clearly that with an odd number of pairs of poles and an odd number of coils, an odd number of coils are short-circuited at one time, so that, as the total number of coils is odd, an even number is left to be divided between the two armature circuits, which are, therefore, equal. Referring back to Fig. 12, it will be seen that in the case of an even number of pairs of poles, an even number of coils are short-circuited, and as the total number of coils is necessarily odd, an odd number remains to be divided between the two armature circuits, so that these are necessarily unequal.



If, however, in Fig. 13 the brushes are put 108° apart instead of 180°, coil 24 would be taken from the circuit given in the upper line of numbers and put in the other circuit. There would then be 24 coils in one circuit, and 22 in the other, instead of 23 in both. With the large number of coils used in practice, however, these slight inequalities cause no trouble.

If y were chosen odd, 9 for instance, s would equal 46 or 44.

$$S = \frac{n}{2} \cdot y \pm 1 = \frac{10}{2} \cdot 9 \pm 1 = 46$$
 or 44.

This is in accordance with the observation made above, that in the case of an odd number of pairs of poles the number of coils may be even. The diagram for this case is given in Fig. 14, where s=46, n=10, y=9. In the position shown, coils 8, 17, 26, 35, and 44 are short-circuited by the negative brush, and coils 31, 40, 3, 12, and 21 by the positive brush. The circuits through the armature are,—

$$\longrightarrow \quad - \left\{ \begin{array}{c} 7 \text{--}16 \text{--}25 \text{--}34 \text{--}43 \text{--} 6 \text{--}15 \text{--}24 \text{--}33 \text{--}42 \text{--}5 \text{--}14 \text{--}23 \text{--}32 \text{--}41 \text{--}4 \text{--}13 \text{--}22} \\ 45 \text{--}36 \text{--}27 \text{--}18 \text{--} 9 \text{--}46 \text{--}37 \text{--}28 \text{--}19 \text{--}10 \text{--}1 \text{--}38 \text{--}29 \text{--}20 \text{--}11 \text{--}2 \text{--}39 \text{--}30} \end{array} \right\} + \quad \longrightarrow \quad - \left\{ \begin{array}{c} 7 \text{--}16 \text{--}25 \text{--}34 \text{--}43 \text{--}6 \text{--}15 \text{--}24 \text{--}33 \text{--}42 \text{--}5 \text{--}14 \text{--}23 \text{--}32 \text{--}41 \text{--}4 \text{--}13 \text{--}22} \\ 45 \text{--}36 \text{--}27 \text{--}18 \text{--}9 \text{--}46 \text{--}37 \text{--}28 \text{--}19 \text{--}10 \text{--}1 \text{--}38 \text{--}29 \text{--}20 \text{--}11 \text{--}2 \text{--}39 \text{--}30} \end{array} \right\} + \quad \longrightarrow \quad - \left\{ \begin{array}{c} 7 \text{--}16 \text{--}25 \text{--}34 \text{--}43 \text{--}43 \text{--}24 \text{--}33 \text{--}42 \text{--}5 \text{--}14 \text{--}23 \text{--}32 \text{--}41 \text{--}4 \text{--}13 \text{--}22} \\ 45 \text{--}36 \text{--}27 \text{--}18 \text{--}9 \text{--}46 \text{--}37 \text{--}28 \text{--}19 \text{--}10 \text{--}1 \text{--}38 \text{--}29 \text{--}20 \text{--}11 \text{--}2 \text{--}39 \text{--}30} \end{array} \right\} + \quad \longrightarrow \quad - \left\{ \begin{array}{c} 7 \text{--}16 \text{--}25 \text{--}34 \text{--}43 \text{$$

giving, as in Fig. 13, two equal paths through the armature.

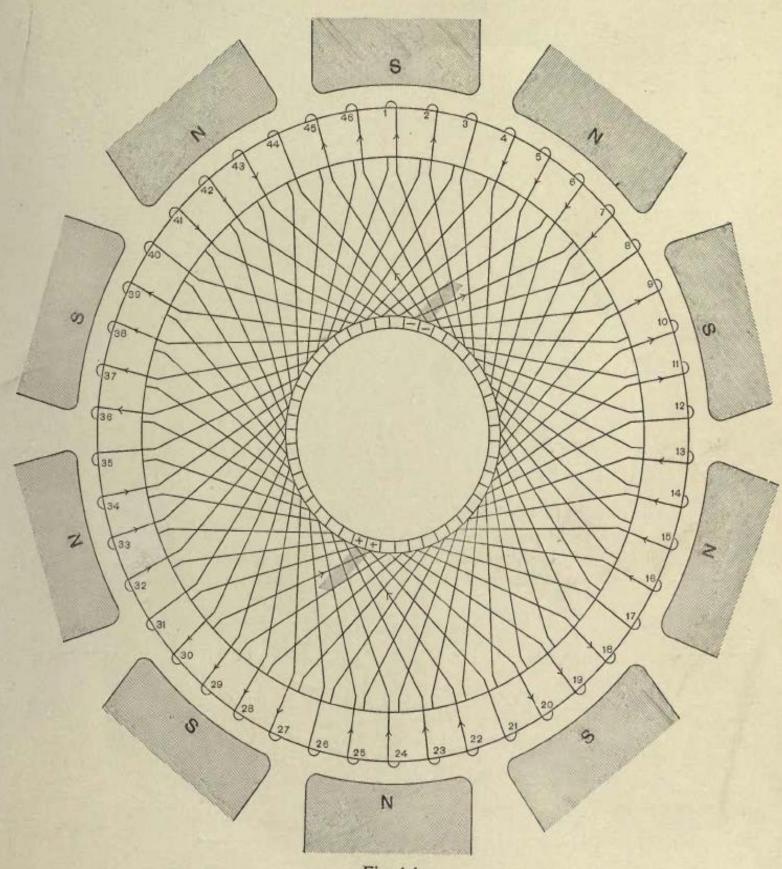


Fig. 14
TWO CIRCUIT, SINGLE WINDING.

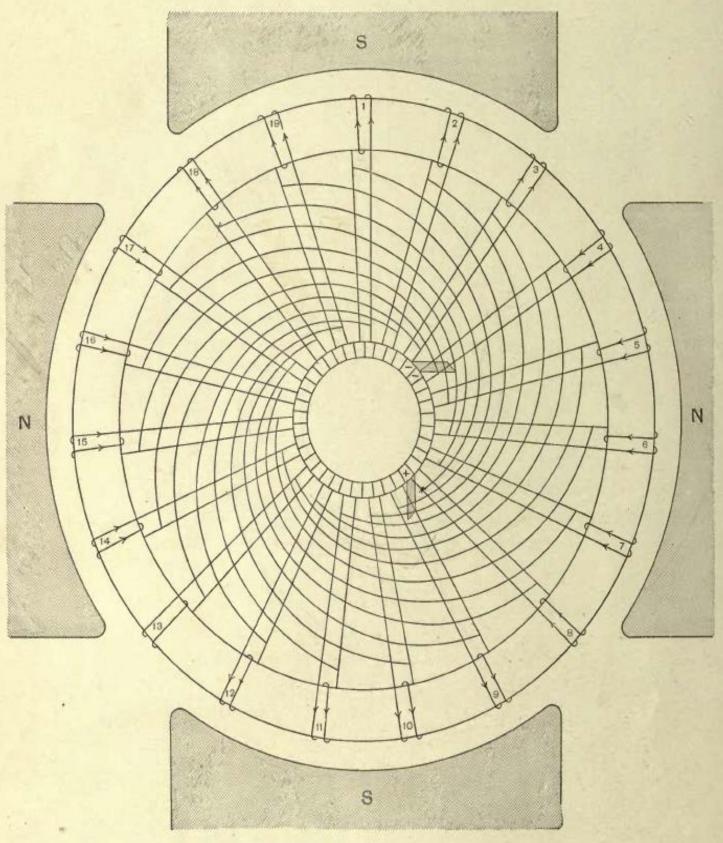


Fig. 15
TWO CIRCUIT, SINGLE WINDING.

In Fig. 15 is given a winding that has been used in practice with considerable success, owing partly to the extreme regularity of all connections, and still more to the fact that it involves the use of twice as many commutator segments as coils. Only one coil in series is short-circuited at each brush, and the volts per segment are one-half what they would be in the unmodified long-connection winding. The number of coils to be used is, as in the unmodified winding,  $s = \frac{n}{2} \cdot y \pm 1$ . Thus, in Fig. 15, n = 4, y = 9,  $s = \frac{4}{2} \cdot 9 + 1 = 19$ . Coil 1 is connected to coil 10, etc.

It will also be noted that those segments  $\left[\frac{360}{\frac{n}{2}}\right]^{\circ}$  from each

other are connected together. The number of segments =  $\frac{n}{2} \cdot s$ , of which  $\frac{n}{2}$ , at distances of  $\left[\frac{360}{n}\right]^{\circ}$  from each other, are

connected together. If every other one of the radial connections from the coils to the commutator are discarded, the winding becomes once more the plain, long-connection, twocircuit, gramme winding.

At the position shown, coil 13 is short-circuited by the negative brush, and the circuits through the armature are, —

$$\longrightarrow$$
  $-\left\{ \begin{array}{lll} 3-12-2-11-1-10-19-&9-18\\ 4-14-5-15-6-16-&7-17-&8 \end{array} \right\} + \longrightarrow$ 



Figure 16 is an application of the same type of winding to a six-pole gramme ring. n=6, y=6,  $s=\frac{n}{2}y\pm 1=\frac{6}{2}\cdot 6+1=19$ . There are  $19\times\frac{6}{2}=57$  segments. All segments distant from each other by  $\frac{360}{n}=120^{\circ}$  should be connected together. Some of the cross-connections are shown inside the armature.

At the position shown, coil 12 is short-circuited by the positive brush. The circuits through the armature are—

If the connections shown inside the commutator, together with one-third of the segments, had been omitted, there would have been an unequal distribution of potential about the commutator. Between two segments would be found a certain voltage, V, and between the next two 2 V; then V again, etc.

If it should be desirable to diminish the number of commutator segments to one-half the number of coils, it may be done by the method of connection shown in Fig. 17, page 34, which will be recognized at once as the multipolar ring counterpart of the two-circuit winding as applied to multipolar drums. This winding will be referred to as a "short connection," two-circuit gramme winding. In the "long-connection" type, examples of which have just been given, connection has been made between coils situated in fields of like polarity. But in the "short-connection" type, connection is made between coils in adjacent fields. Both methods are feasible in ring windings, because the two ends of a coil located at a certain point of the periphery are accessible for connection at the commutator end if desired, but in drum windings only one end of a conductor located at a given point of the periphery is accessible at the commutator end, the other end of the conductor being necessarily connected across at the opposite end of the armature, and in consequence, also, must be connected over to a conductor in an adjacent field of unlike polarity, in order that the electromotive force, which is, say, from front to back in the first conductor, may add itself to that in the second conductor, which must therefore be from back to front; that is, the second conductor must be situated in a field of opposite polarity. Thus there are two sub-classes of two-circuit, multipolar ring windings, in the first of which (the "long-connection" winding) coils in fields of like polarity are connected in succession, and in the second of which, as in the two-circuit, multipolar drum winding, the conductors immediately succeeding each other are situated in fields of opposite polarity.

In this "short-connection" winding for two-circuit multipolar rings the formula for determining the proper number of coils, s, for any number of poles, n, is —

$$s = ny \pm 2$$
,

where y, the pitch, may equal any integer, odd or even.

In connecting up this "short-connection" type of winding the following additional rule should be borne in mind in the interpretation and application of the meaning of the pitch, y: The number of coils in this winding, being from the formula always even, if y is also even, it is necessary in connecting up to use as the pitch, alternately, (y-1) and (y+1) instead of always y. Otherwise, if the coils are numbered successively

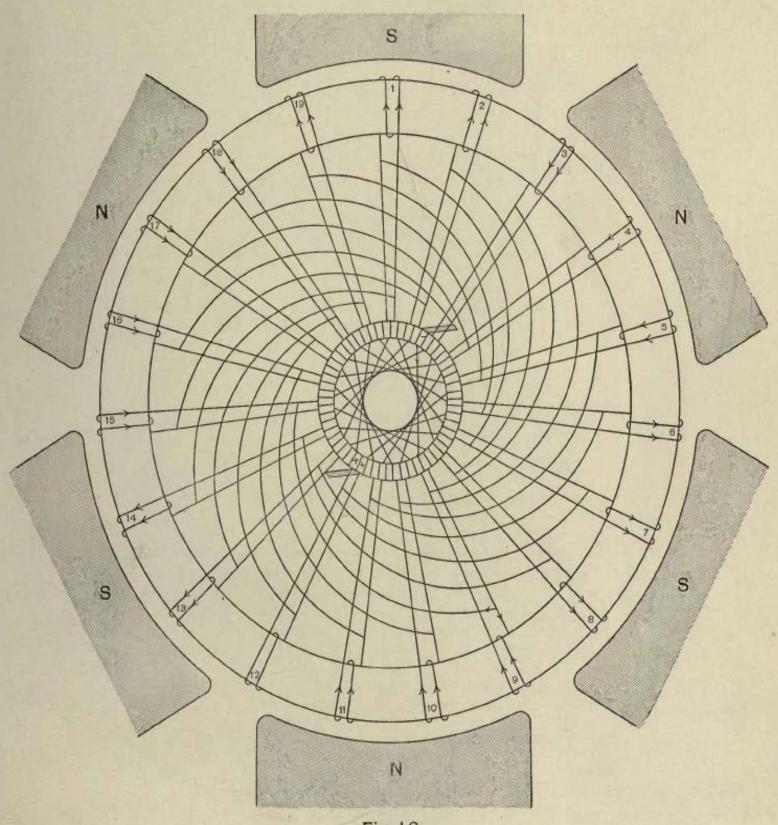


Fig. 16
TWO CIRCUIT, SINGLE WINDING.

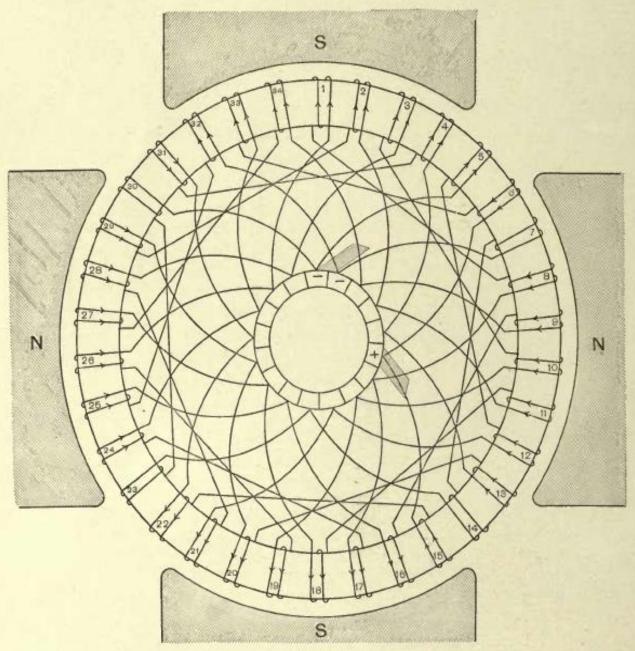


Fig. 17
TWO CIRCUIT, SINGLE WINDING.

from No. 1 on, the even-numbered coils would never be touched, if an odd-numbered conductor were started with, and vice versa. If y were used every time as the pitch, a double winding would be obtained. This case will be treated later.

It may also be well to note that (y-3) and (y+3) could be used alternately as the pitch. It is thought, however, that no advantages, and several disadvantages, would result from such a choice of pitches.

Figure 17 represents a two-circuit, single-wound, four-pole ring of the "short-connection" type just described.

$$n = 4$$
,  $y = 8$ ,  $s = ny \pm 2 = 4 \times 8 + 2 = 34$ .

This is the case referred to above, in which, s being even and also y, (y-1) and (y+1) must be used alternately as the pitch in connecting up. The sequence of connections will be seen in the figure to be 1, 1+7=8, 8+9=17, 17+7=24, etc.

Number of commutator segments =  $\frac{3}{0}$  = 17.

In the position shown, coils 7, 14, 23, and 30, in series, are short-circuited at the negative brush, and the circuits through the armature are, —

There are 14 coils in one path and 16 in the other. A little later, coils 6, 33, 24, and 17, in series, will be short-circuited by the positive brush, and coils 7, 14, 23, and 30 will take their place, the circuits through the armature then becoming, —

$$\longrightarrow \ - \left\{ \begin{smallmatrix} 7-14-23-30-&5-12-21-28-3-10-19-26-1-&8-----\\ 32-25-16-&9-34-27-18-11-2-29-20-13-4-31-22-15 \end{smallmatrix} \right\} + \ \longrightarrow$$

A further inspection of the diagram will show the unsymmetrical arrangement of the short-circuited and adjacent coils, causing the induction in some coils to act in opposition to that in others with which it is in series. This is less marked with large numbers of coils.

The chief disadvantages of the "short-connection" winding are that adjacent coils have between them, periodically, the full E.M.F. of the armature, and that the end windings are complicated.



Figure 18 represents another two-circuit, single-wound, "short-connection" gramme winding, in which  $s = ny \pm 2$  =  $4 \times 5 \pm 2 = 22$ . In this case y, the pitch, is odd, and consequently the sequence of connections is 1, 1+5=6, 6+5=11, 11+5=16, etc., thus advancing each time by 5, and not, as in the case of Fig. 17, page 34, where y was even, alternately by (y+1) and (y-1). Corresponding ends of coils are connected together; thus, the end of 1 and the end of 6, the beginning of 6 and the beginning of 11, etc.

At the position shown, coils 5, 10, 15, and 20 are short-circuited by the negative brush, and the circuits through the armature are,—

$$\longrightarrow \ - \left\{ \begin{smallmatrix} 22-17-12-&7-2-19-14-&9-\\ 3-&8-13-18-1-&6-11-16-21-4 \end{smallmatrix} \right\} + \ \longrightarrow$$

The winding is subject to the disadvantages noted in connection with Fig. 17, page 34.

Instead of having the objectionable crossings at the terminals of the coils, as shown in Fig. 18, page 37, alternate coils should be wound right and left handedly. This would only be useful in cases where all the connecting is done at one end, which should be avoided when possible.

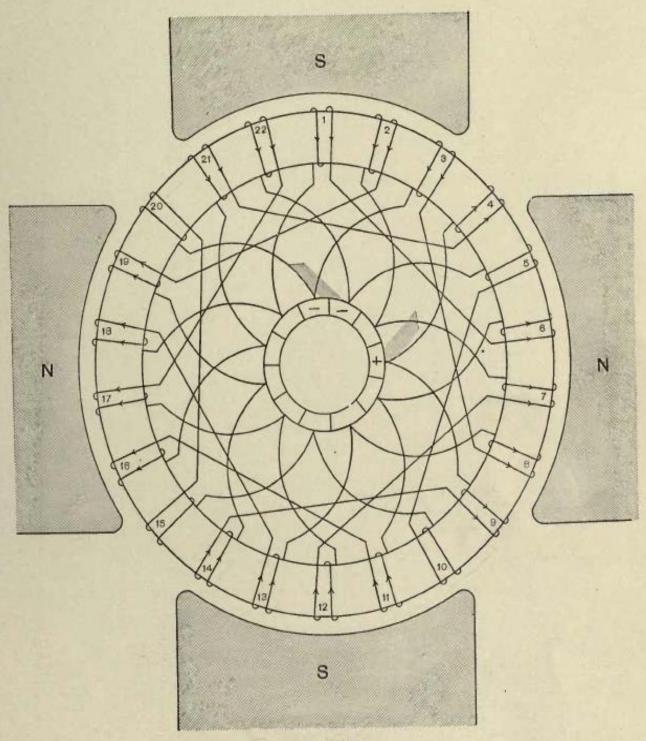


Fig. 18
TWO CIRCUIT, SINGLE WINDING.

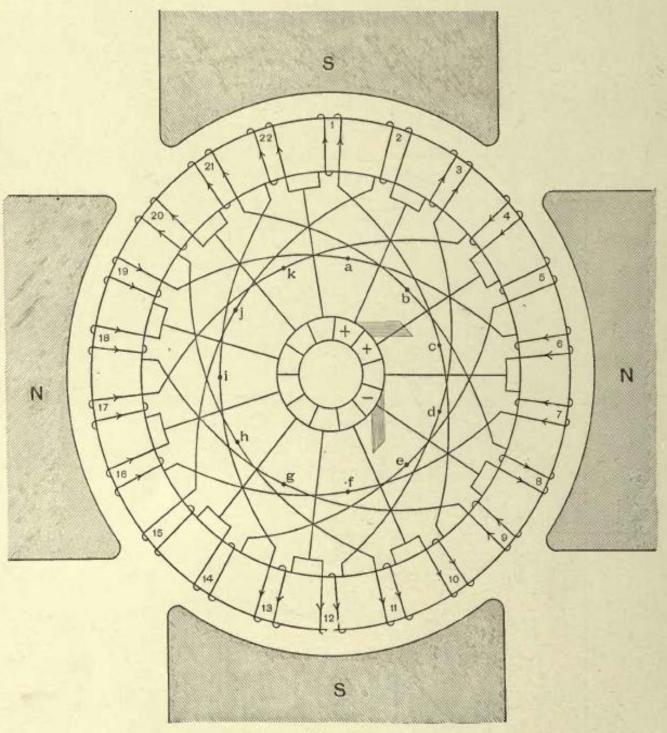


Fig. 19
TWO CIRCUIT, SINGLE WINDING.

Instead of connecting together in pairs coils lying in fields of opposite polarity, as in Figs. 17 and 18, adjacent coils may be connected together as shown in Fig. 19, and these connected across to coils in the nearest field of like polarity. The number of commutator segments is equal to one-half of the number of coils. The inherent identity of this and the "long-connection" winding may be seen by doing away with the leads to the commutator segments, and substituting leads from the eleven points lettered a, b, c, d, etc. The result will be a simple "long-connection" gramme winding, with half as many coils of twice as many turns each.

Therefore, the best way of laying out such a winding is to apply the rules for the "long-connection" winding, and make the connections shown in Fig. 19, instead of those of the regular "long-connection" gramme winding.

This winding gives half as many commutator segments as coils.

In the position shown, coils 5, 14, 15, and 2 are short-circuited by the positive brush, and the circuits through the armature are, —

$$\longrightarrow$$
  $-\left\{ \begin{array}{llll} 8-21-20-11-10-&1-22-13-12-3 \\ 9-18-19-&6-&7-16-17-&4---- \end{array} \right\}+$   $\longrightarrow$ 



# CHAPTER IV.

## TWO-CIRCUIT, MULTIPLE-WOUND, MULTIPOLAR RINGS.

THE next class is that of the two-circuit, multiple-wound, long-connection ring windings. The general formula is, —

 $s = \frac{n}{2} \times y \pm m$ ,

where

s = number of coils,

n = number of poles,

y = pitch,

m = number of windings.

The "m" windings will consist of a number of independently re-entrant windings equal to the greatest common factor of "y" and "m."

Therefore, when it is desired that the "m" windings shall combine to form one re-entrant system, it will be necessary that the G.C.F. of "y" and "m" shall be made equal to 1.

Figure 20 represents a two-circuit, doubly re-entrant, double-wound ring armature.

$$s=26,$$
  $n=4,$   $m=2.$  
$$s=\frac{n}{2}\times y\pm m,$$
  $26=\frac{4}{2}\times y+2,$   $\therefore y=12.$ 

Greatest common factor of y (12) and m (2) is 2. Therefore the winding will be doubly re-entrant.

At the position shown, coils 24 and 12, in series, are short-circuited by the negative brush. The circuits through the armature are,—

$$\rightarrow \begin{cases} -\left\{ \frac{25-13-1-15-3-17-}{26-14-2-16-4-18-} \right\} + \\ -\left\{ \frac{10-22-8-20-6-}{11-23-9-21-7-19-5} \right\} + \end{cases}$$

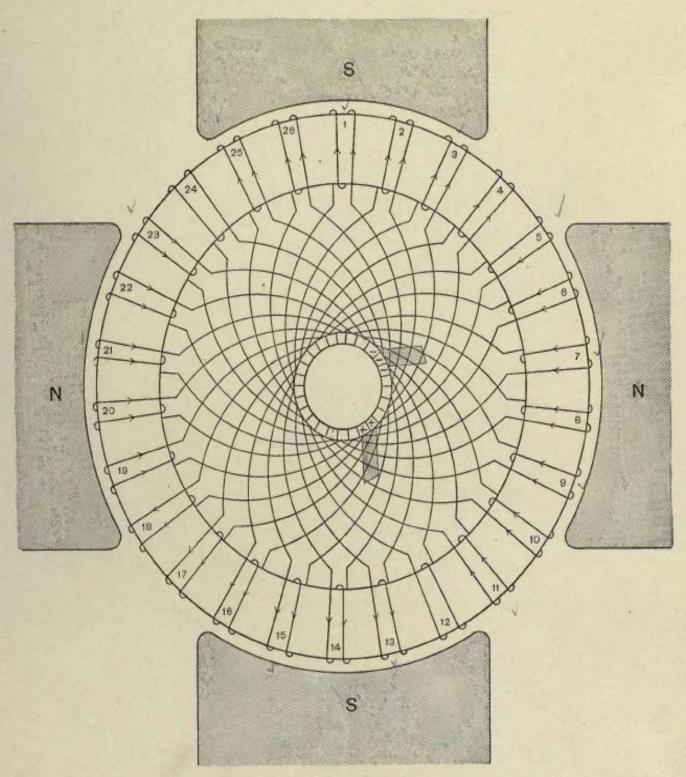


Fig. 20
TWO CIRCUIT, DOUBLE WINDING,

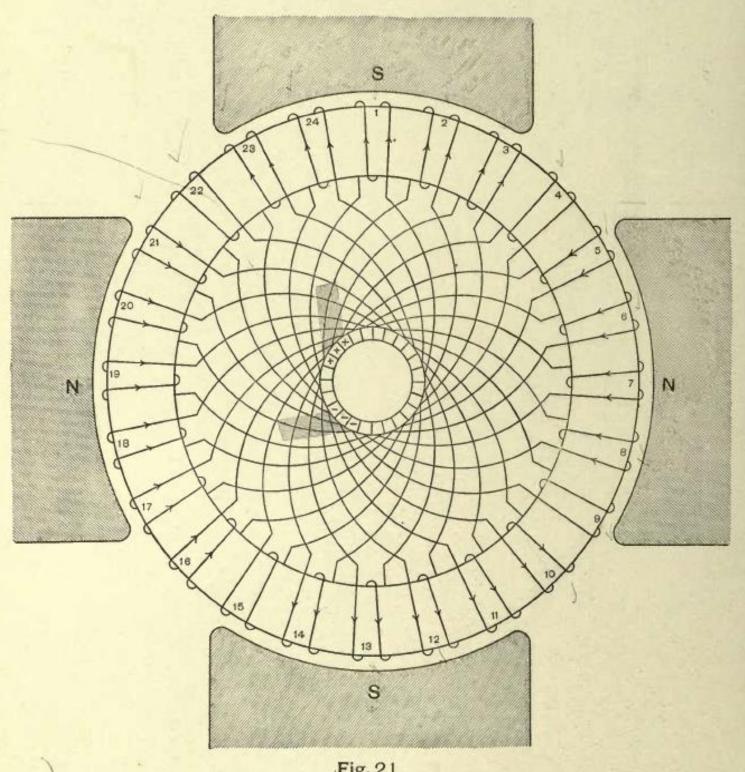


Fig. 21
TWO CIRCUIT, DOUBLE WINDING,

Figure 21 represents a two-circuit, singly re-entrant, double-wound ring armature.

In this case y=11, n=4, and m=2.  $s=\frac{4}{2}\times 11\pm 2=20$  or 24. 24 coils are taken. G.C.F. of "y" and "m" being 1, the winding is singly re-entrant.

In the position given, coils 9 and 22 are short-circuited at the negative brush, and 4 and 15 at the positive. The circuits through the armature are,—

$$\longrightarrow \left\{ \begin{array}{l} -\left\{ \begin{array}{l} 20-7-18-5-16---\\ 21-8-19-6-17--- \end{array} \right\} + \\ -\left\{ \begin{array}{l} 11-24-13-2---\\ 10-23-12-1-14-3 \end{array} \right\} + \end{array} \right\}$$



Figure 22 represents another two-circuit, singly reentrant, double-wound ring armature.

$$m=2, n=6, y=7, s=\frac{n}{2}y\pm 2=\frac{6}{2}\times 7\pm 2=19 \text{ or } 23.$$

"y" and "m" being prime, the winding is singly reentrant.

At the position shown, coils 4, 11, and 18 are short-circuited at the positive brush, and the circuits through the armature are: —

Two two-circuit, singly re-entrant, triple windings for gramme rings are given below without diagrams:—

$$m=3, n=6, y=7, s=\frac{n}{2} \times y \pm 3 = \frac{6}{2} \times 7 + 3 = 24.$$

The connections would be, -

 $1 - 8 - 15 - 22 - 5 - 12 - 19 - 2 - 9 - 16 - 23 - 6 - 13 - 20 - 3 - 10 - 17 - 24 - 7 - 14 - 21 \\ - 4 - 11 - 18 - 1$ 

$$m=3$$
,  $n=10$ ,  $y=10$ ,  $s=\frac{10}{2}\times 10-3=47$ .

1-11-21-31-41-4-14-24-34-44-7-17-27-37-47-10-20-30-40-3 -13-23-33-43-6-16-26-36-46-9-19-29-39-2-12-22-32-42 -5-15-25-35-45-8-18-28-38-1

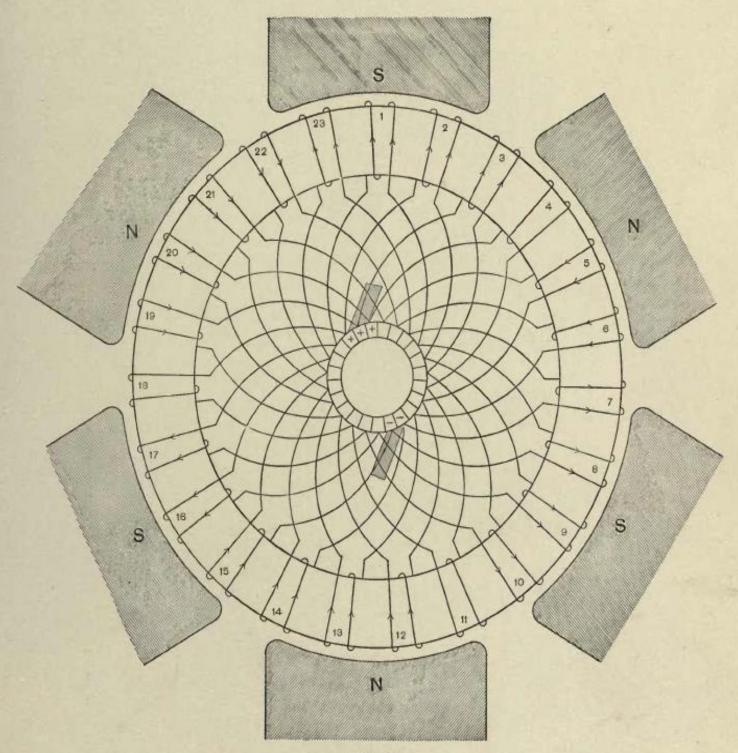


Fig. 22
TWO CIRCUIT, DOUBLE WINDING.

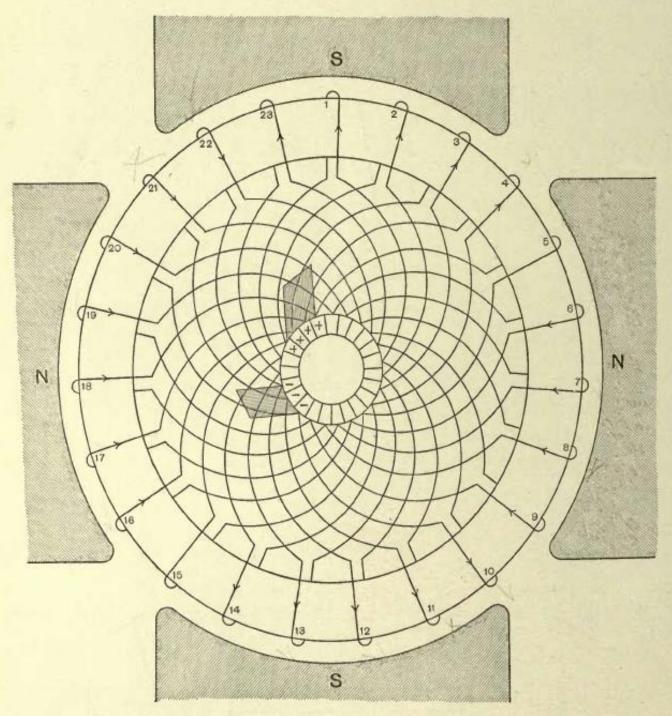


Fig. 23
TWO CIRCUIT, TRIPLE WINDING.

Figure 23 represents a two-circuit, singly re-entrant, triple winding.

$$m=3$$
,  $n=4$ ,  $y=10$ ,  $s=\frac{4}{3}\times 10\pm 3=23$ .

"m" and "y" being prime, the winding is singly re-entrant.

In the position shown, coils 5 and 15, in series, are short-circuited by the positive brush. The circuits through the armature are, —

The extreme irregularity of the various circuits in multiple is not characteristic of the winding, but is merely due to the very small number of coils chosen. In practical cases it would be negligible.

From the formula and conditions of page 40, and from the examples just given, it will be seen that two-circuit, multiple-wound, ring windings may be divided into the three following cases:—

Case I.—"y" and "m" are mutually prime. This gives a singly re-entrant winding of "m" multiple windings.

Illustration: -  $n=4, y=7, m=4, s=\frac{4}{5} \times 7 + 4 = 18.$ 

Connections are, — 1-8-15-4-11-18-7-14-3-10-17-6-13-2-9-16-5-12-1.

May be expressed symbolically as (000)

Case II. - "y" a multiple of "m." This gives "m" independently re-entrant windings.

Illustration: —  $n=4, y=8, m=4, s=\frac{4}{2} \times 8+4=20.$ 

1- 9-17-5-13-1 2-10-18-6-14-2

> 3-11-19-7-15-3 4-12-20-8-16-4

May be expressed symbolically as OOOO.

CASE III. — "y" and "m" have common factors. This gives a number of independently re-entrant windings, equal to the greatest common factor of "y" and "m."

Illustration: -

Connections are, -

$$n=4$$
,  $y=6$ ,  $m=4$ ,  $s=\frac{4}{2}\times 6+4=16$ .

The result is a two-circuit, quadruple winding with two independently re-entrant windings, because 2 is the greatest common factor of "y" and "m."

The connections are, -

May be expressed symbolically as @ @.

Case II. is really a special instance of Case III.

The above formula and controlling conditions will be found to hold for all numbers of poles, coils, pitches, and windings of the two-circuit, long-connection type of gramme-ring armature windings.



Figure 24 is a two-circuit, singly re-entrant triple winding of the type described in connection with Figs. 15 and 16, which, it should be remembered, is only a modification of the long-connection type.

$$n=4, y=10, m=3, s=\frac{n}{2} \times y \pm m = \frac{4}{2} \times 10 + 3 = 23.$$

At the position shown, coil 21 is short-circuited at the negative brush, and coils 3 and 4 at the positive brush. The circuits through the armature are,—

$$\rightarrow \begin{cases}
-\left\{\begin{array}{c}
8-18-5-\\
9-19-6-16\\
-20-7-17
\end{array}\right\} + \\
-\left\{\begin{array}{c}
22-12-2-15\\
-11-1-14\\
-10-23-13
\end{array}\right\} + \\
+
\end{cases}$$

Figure 24 should be compared with Figs. 15 and 16.

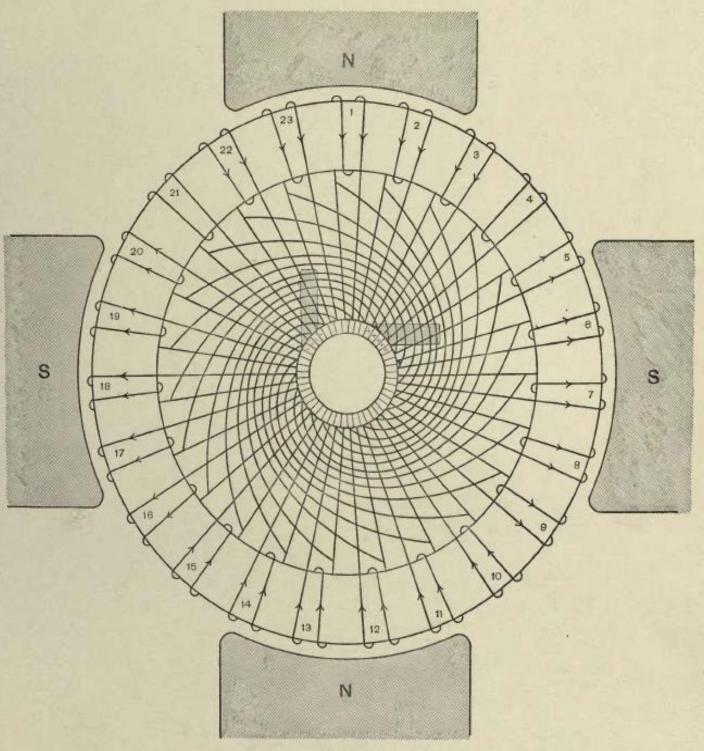
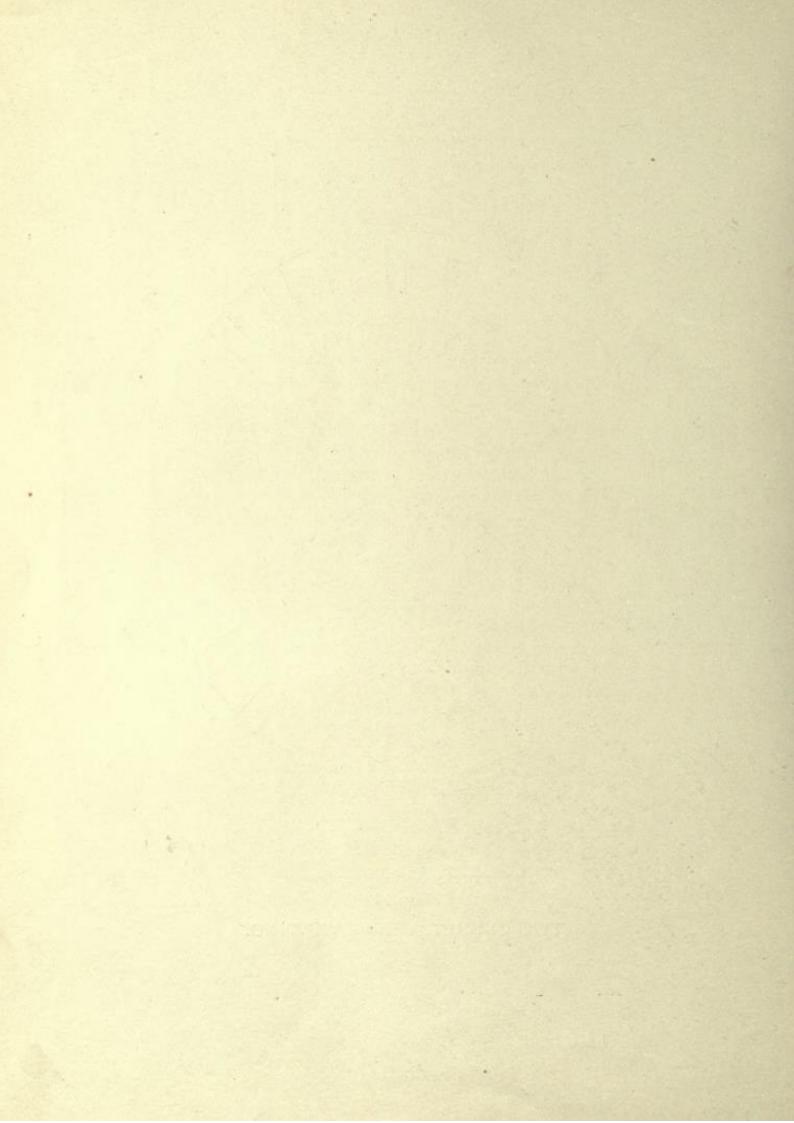


Fig. 24
TWO CIRCUIT, TRIPLE WINDING.





## CHAPTER V.

#### DRUM ARMATURE WINDINGS.

In drum windings, all connections from bar to bar must be made upon the rear and front ends exclusively, it not being practicable to bring connections through inside from back to front as is the case with rings. Consideration of this limitation will show that the two sides of any one coil must be situated in fields of opposite polarity, so that the electromotive forces, generated in the active conductors of a coil by their passage through their respective fields, shall be in the same direction.

In the case of a drum, it should also be noted that a coil is linked with the whole or nearly the whole flux from one pole piece, instead of, as in the ring armature, with only onehalf of the flux.

### BIPOLAR DRUM WINDINGS.

The winding of bipolar armatures is much less simple in the case of drums than in that of rings, and it will therefore be necessary to give considerable attention to the various methods in which such windings may be carried out. Figure 25 represents essentially the winding devised by von Hefner-Alteneck. It is used chiefly for small, smooth-core, wire-wound armatures, and the element of the winding, represented in the diagram by a pair of face conductors, and a back connection consists usually, in practice, of a coil of several turns, comparable in some respects to the coil of the ring windings; but in the diagram only one turn per coil will be shown. This will also be advantageous, inasmuch as large, iron-clad, bar-wound, multipolar drum armatures are derived from, and diagrammatically are very analogous to, the wire-wound, smooth-core armatures now under consideration.

An examination of Fig. 25 shows that, starting from a commutator segment, the winding proceeds over the front end to conductor No. 1; down No. 1 over the back to conductor No. 8, which, it should be noted, is adjacent to the conductor diametrically opposite No. 1. From No. 8 the winding returns to the next commutator segment, and is then carried to conductor No. 3 (skipping No. 2, which will later be joined over the back to a conductor almost diametrically opposite to it), down No. 3, over the back to No. 10, etc. From this it is seen that the "pitch" on the back end is 7 and on the front end is -5.

In the position shown, the circuits through the armature are, -

The coil represented by the conductors 13 and 4 is short-circuited at the positive brush, and coil 12-5 at the negative brush.

The customary convention is adopted in the diagram,  $\otimes$  indicating a current from the observer into the paper, and  $\otimes$  a current up out of the surface of the paper toward the observer.

A serious fault of this winding is that large differences of potential exist between adjacent conductors (or, usually, groups of conductors). This would be of no importance with the small numbers of conductors represented in these diagrams, but in actual cases, large numbers of conductors are used, and are placed close together in order to waste no available space.

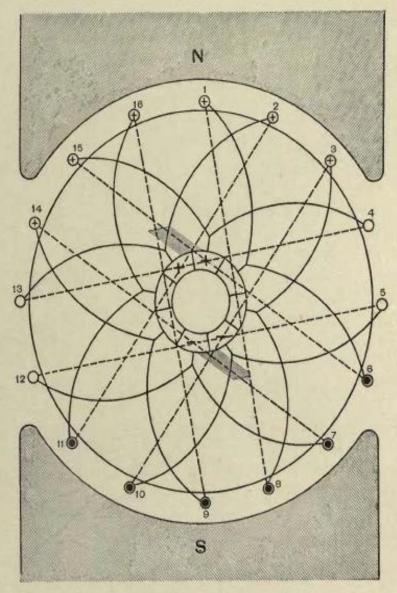


Fig. 25
TWO CIRCUIT, SINGLE WINDING.

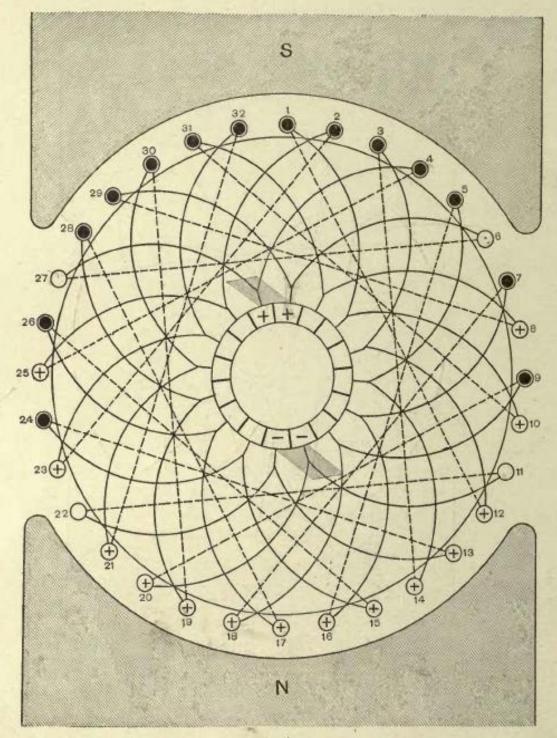


Fig. 26
TWO CIRCUIT, SINGLE WINDING.

Figure 26 gives the diagram of a winding discussed by Swinburne. Its characteristic feature is the use of a small pitch (in the figure the pitch at the back end is 11, and at the front end it is -9), whereby the turns consist of conductors separated by a much smaller angular distance than in the von Hefner-Alteneck winding.

An advantage of this winding is that there is much less crossing of the end connections than is the case where the pitch is taken larger. Thus the difficult question of insulation at the ends of the armature is greatly simplified.

Still further, it has been pointed out by Swinburne that the demagnetizing effect of the armature on the field is reduced, as may be seen from the fact that the currents in the conductors in the demagnetizing belt between the pole tips, namely, 23, 24, 25, and 26, and in 7, 8, 9, and 10, are alternately in opposite directions, and thus neutralize each other.

A serious disadvantage is that the short-circuited coils, 6-27 and 11-22, are considerably removed from the neutral line. This, together with the fact that the counter-electromotive forces present in several conductors of the circuit between brushes detract from the volts per unit of length of armature wire, reduces to rather small limits the extent to which such connecting over short chords should be carried.

In the position shown, the circuits through the armature are,—

$$\longrightarrow \ - \left\{ \begin{smallmatrix} 20 - & 9 - 18 - & 7 - 16 - & 5 - 14 - & 3 - 12 - & 1 - 10 - 31 - & 8 - 29 \\ 13 - 24 - 15 - 26 - 17 - 28 - 19 - 30 - 21 - 32 - 23 - & 2 - 25 - & 4 \end{smallmatrix} \right\} + \ \longrightarrow$$



In Fig. 27 it will be seen that the number of coils is odd (in the two preceding diagrams it was even), with the result that the two active sides of such coils may now be diametrically opposite.

This would not, however, usually be advisable, as it makes many more crossings at the ends, and therefore increases the difficulty of insulating.

Some advantage results from bringing the short-circuited coil (in the figure, coil 24-9 is short-circuited by the negative brush), exactly in the neutral line, this being, of course, only possible when the conductors forming its active sides are diametrically opposite.

The circuits through the armature in the position shown are. —

The pitch on the back end is 15, and on the front end it is -13.

Owing to the number of segments being odd, only one coil is short-circuited at once, unless wide brushes are used.

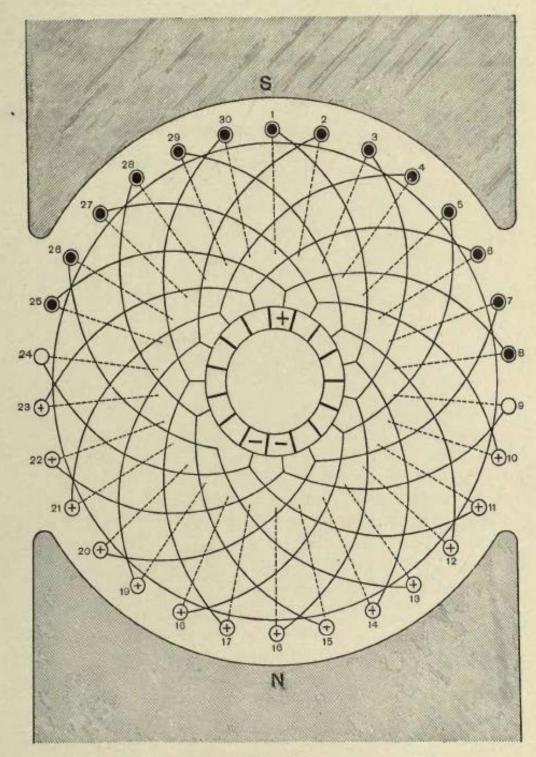


Fig. 27
TWO CIRCUIT, SINGLE WINDING.

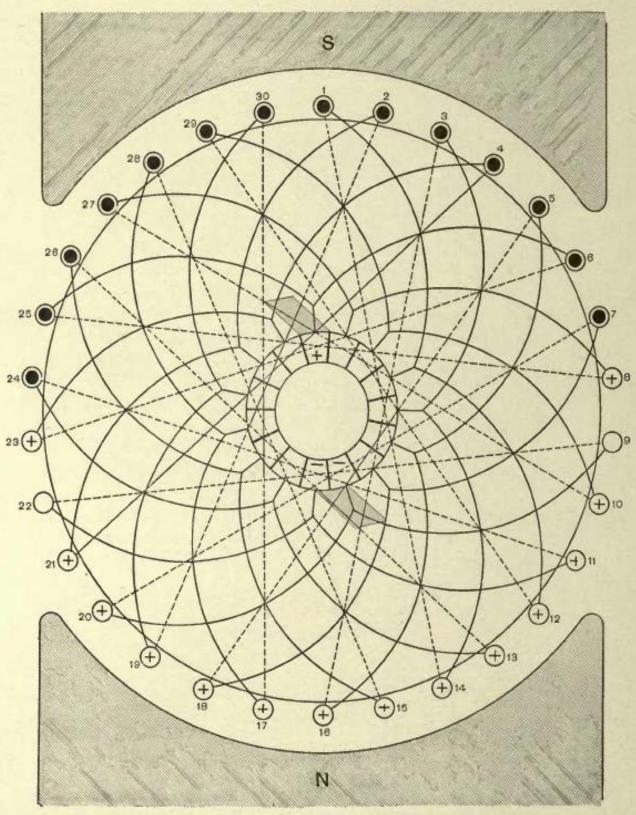


Fig. 28
TWO CIRCUIT, SINGLE WINDING.

In Fig. 28 there is also an odd number of coils (and therefore an odd number of commutator segments). But instead of connecting over the back from No. 1 to No. 16 (the conductor diametrically opposite No. 1) as in Fig. 17, connection is made over the back from No. 1 to No. 14, then over the front to No 3, etc., the pitch at the back end being 13, and on the front end -11. It is, therefore, a mild form of the Swinburne chord winding, as described in connection with Fig. 26. The end connections are better distributed and have fewer crossings than was the case in Fig. 27, where diametrically opposite conductors were connected into coils.

In the position shown, coil 22-9 is short-circuited at the negative brush, and the circuits through the armature are,—

$$\longrightarrow -\left\{ \begin{smallmatrix} 11-24-13-26-15-28-17-30-19-&2-21-&4-23-&6\\ 20-&7-18-&5-16-&3-14-&1-12-29-10-27-&8-25 \end{smallmatrix} \right\} + \longrightarrow$$



In Fig. 29 the winding is carried on over a still shorter chord, the pitch at the back end being 11 and at the front end −9.

It is very instructive to compare Figs. 27, 28, and 29, all of which have 30 face conductors (15 coils). But in Fig. 27 diametrically opposite conductors are connected over the back, the back pitch being 15. Figure 28 is a weak chord winding, the back pitch being 13. Figure 29 is a decided chord winding, the back pitch being 11. The points to be compared are the positions of the short-circuited conductors with reference to the neutral line; the amount of neutralizing of the effect of the demagnetizing belt between pole tips, and the comparative amount of crossing of connectors at the ends.

In Fig. 27 it was shown that diametrically opposite conductors could be connected into coils if the number of coils were chosen odd.

The same object may be attained with an even number of coils by winding them in two layers instead of in one layer, as has been the case in all the heretofore described bipolar drum armatures.

It should be again noted that the term "conductors" is used in these explanations, although "groups of conductors" could often be substituted therefor in small, smooth-core, wire-wound armatures.

Thus the set of "one-layer windings," just described, are those in which "conductors" or "groups of conductors" are, in the completed winding, arranged in one layer, although the individual wires of such a group may optionally occupy one or several layers. In the same way, the two-layer windings now to be described are those in which the completed winding consists of "conductors" or "groups of conductors" arranged in two layers, although the actual depth of individual wires may, when desirable, be greater than two.

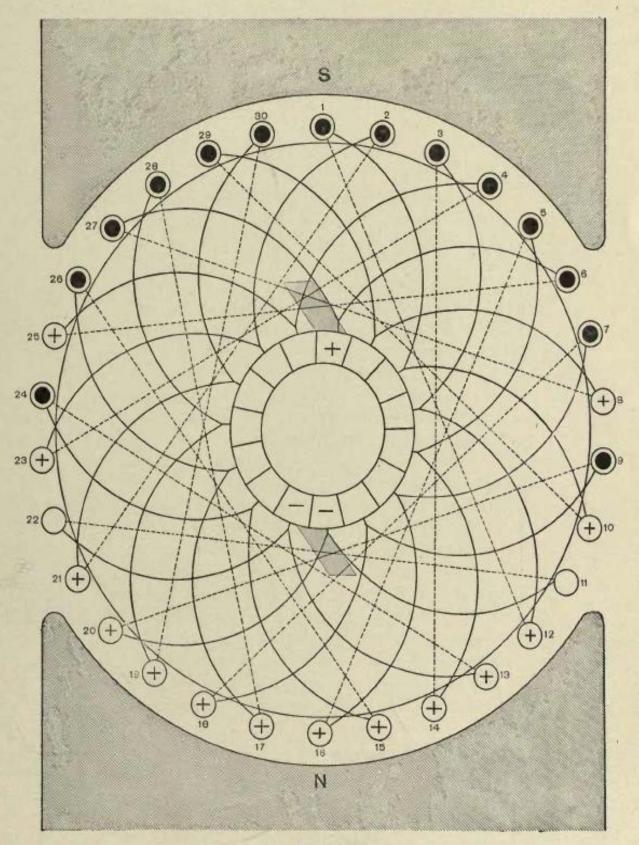
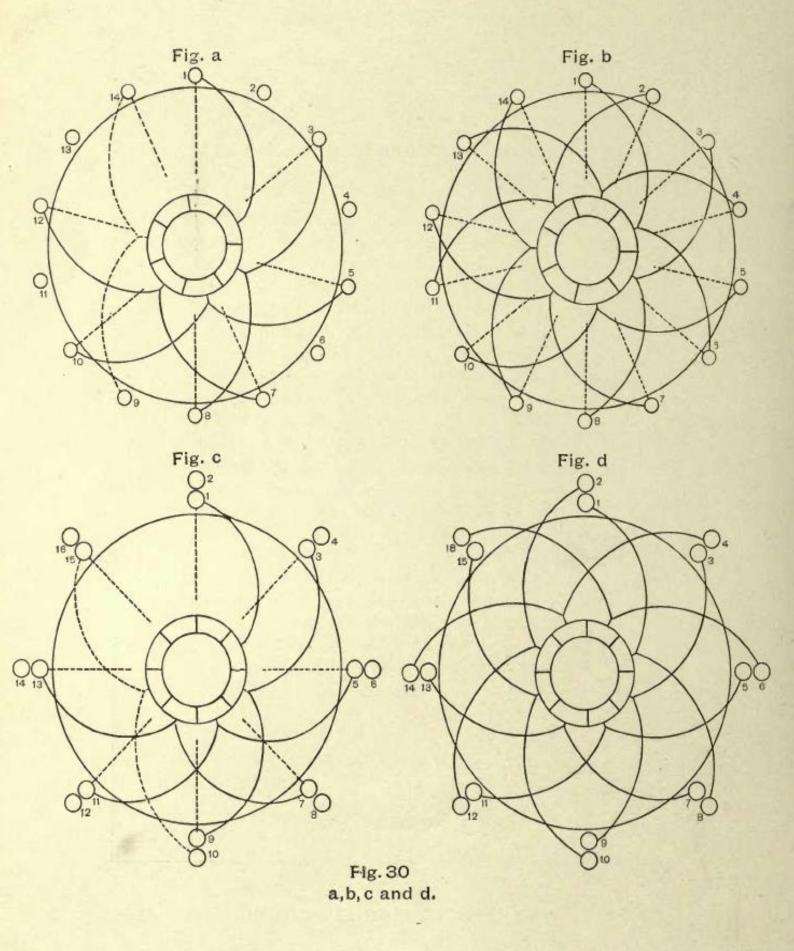


Fig. 29
TWO CIRCUIT, SINGLE WINDING.





In Fig. 30, diagrams a and b represent a single-layer bipolar drum winding with an odd number of coils, in which diametrically opposite conductors are connected together into coils. In diagram a the first half of the winding is carried out and proceeds from a commutator bar to conductor No. 1, to 8, to 3, to 10, to 5, to 12, to 7, to 14, and is then ready for the second half. It will be seen that at this stage only every other coil is connected up, and that only one-half of the commutator segments are utilized. Diagram b shows the winding completed. This winding, which is of the type shown in Fig. 27, is given here for comparison with the two-layer winding shown in diagrams c and d. In Fig. c it will be seen that the first half is exactly the same as the first half of the one-layer winding (except that it contains eight conductors instead of seven), and at the completion of the first half all the conductors of the lower layer are connected up in the order 1-9-3-11-5-13-7-15, and only one-half of the commutator segments are connected in. The coils remaining for the second half, instead of lying between those of the first half, occupy an outer layer. Diagram d shows the completed winding, with all the coils and commutator segments utilized.



Figure 31 represents a two-layer winding with thirty-two conductors, with diametrically opposite conductors connected into coils over the back end.

These back-end connections are not shown, because they would interfere with the clearness of the diagram. The connections are 1-17-3-19-5-21, etc. In the position shown, coil 25-9 is short-circuited at the negative brush and 26-10 at the positive brush, and the circuits through the armature are,—

$$\longrightarrow -\left\{ \begin{smallmatrix} 23-&7-21-&5-19-&3-17-1-16-32-14-30-12-28\\11-27-13-29-15-31-18-2-20-&4-22-&6-24-&8 \end{smallmatrix} \right\} + \longrightarrow$$

It will be seen from this table that maximum difference of potential exists between conductors lying directly over each other in different layers, such as 27 and 28, or 7 and 8. But adjacent conductors have only small differences of potential; therefore, the two layers should be carefully insulated from each other.

It is an advantage to have the conductors 25-9 and 26-10 of the two short-circuited coils all situated on one diameter, as they may therefore be brought diametrical, and therefore are capable of being short-circuited more nearly in the neutral position.

A disadvantage of the winding is that, one-half being wound exclusively in the lower layer and the other half in the upper, they have unequal lengths and different peripheral speeds, and in those recurring positions in which the two circuits through the armature consist respectively of the lower and the upper layer, the condition will be unbalanced.

In practice, however, it is frequently found expedient to use this connection because of the ease of winding, the inequality being made as small as possible. It will be shown later how this inequality may be obviated; the winding will be, however, less easy to execute.

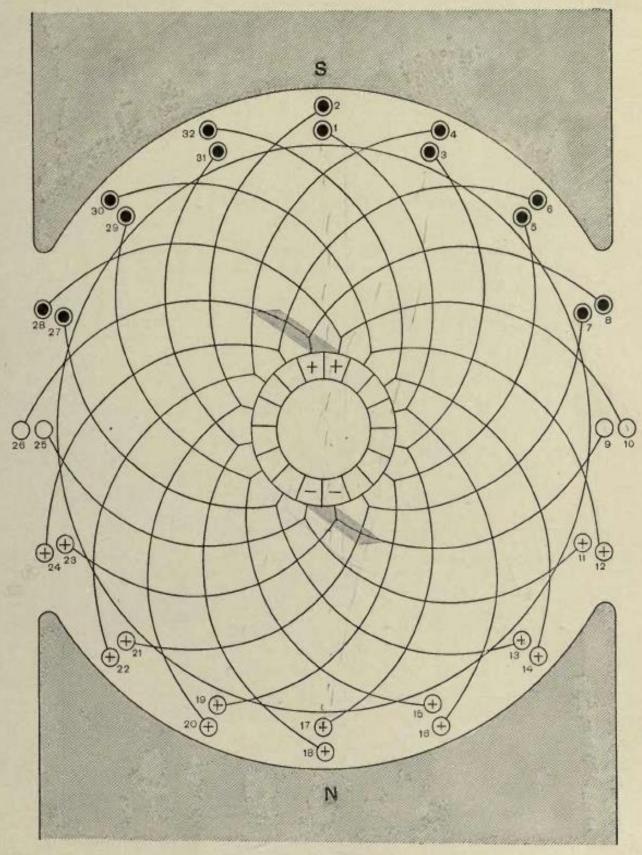


Fig. 31
TWO CIRCUIT, SINGLE WINDING.

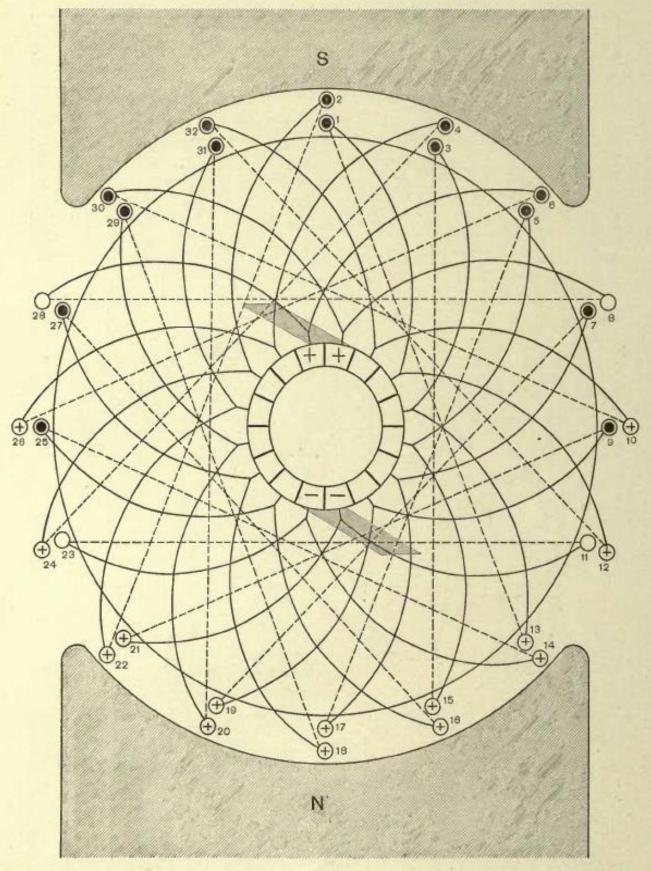


Fig. 32
TWO CIRCUIT, SINGLE WINDING

In Fig. 32 the winding is of the Swinburne type, being connected over the ends along a short chord. Thus, starting from a commutator segment, it passes down No. 1, over the back to No. 13, over the front to No. 3, and so on through 3, 15, 5, 17, 7, 19, 9, 21, 11, 23; but coming over the front from 23 it would naturally go to 13 of the lower layer. This, however, is already used, so the winding continues by No. 14, which is directly over No. 13 in the top layer, and then on through 25–16–27–18–29–20–31–22. From 22 it would naturally go to No. 1, but, as the winding is not yet completed, it must go instead to No. 2, which is directly over No. 1, and then proceed from 2 through 24–4–26–6–28–8–30–10–32–12, and then it closes on itself at No. 1. This winding is not at all difficult, because, although the lower layer is not entirely completed before beginning to wind the upper layer, yet in that part of the armature on which it is desired to wind the upper layer, the lower layer is entirely completed, and for quite a distance beyond, so that there would be no trouble in inserting the necessary insulation, etc.

In the position shown, coil 28-8 is short-circuited at the positive brush, and coil 23-11 at the negative brush. It is a disadvantage to have the short-circuited coils so far from the neutral line.

The circuits through the armature in the given position are, -

It will be seen that in this armature there can be no position in which one layer belongs exclusively to one circuit and the other to the other circuit. Therefore the discrepancy in lengths and peripheral speeds of the two circuits through the armature will, at the most unfavorable moment, be less than when diametrically opposite conductors are connected into coils. The winding has, in common with all chord windings, the advantage of less crossings of the end connections. The diagram shows particularly well the absence of demagnetizing action in the zone of conductors between pole tips.

If, in Fig. 32, page 66, conductor No. 1 had been connected over the back to No. 15 instead of to No. 13, it would still have been a chord winding, but with somewhat less marked characteristics than that of Fig. 32. All the advantages and disadvantages would have been on a smaller scale.

Figure 33 represents a winding in which coils of the outer and inner layer are alternately connected. The rearend connections are not drawn, but are diametrical. Thus the series is 1-15-4-18-5-19-8-22-9-23-12-26-13-27-16-2-17-3-20-6-21-7-24-10-25-11-28-14-1. This makes both circuits through the armature of very nearly equal length and of very nearly equal average peripheral speed.

In the position shown, coil 21-7 is short-circuited by the positive, and 22-8 by the negative brush. The circuits through the armature are,—

$$\longrightarrow \ -\left\{ \begin{smallmatrix} 19-& 5-18-& 4-15-& 1-14-28-11-25-10-24\\ 9-23-12-26-13-27-16-& 2-17-& 3-20-& 6 \end{smallmatrix} \right\} + \ \longrightarrow$$

For this winding to be regular, the number of conductors must be an odd multiple of 4.

Other bipolar drum windings have been proposed by Hering, Western Electric Company, and others, each of which possesses certain special advantages. It might be well especially to consult an article by Hering in "Electrician and Electrical Engineer," Vol. 4, 1885, p. 423, and Vol. 5, 1886, p. 84.

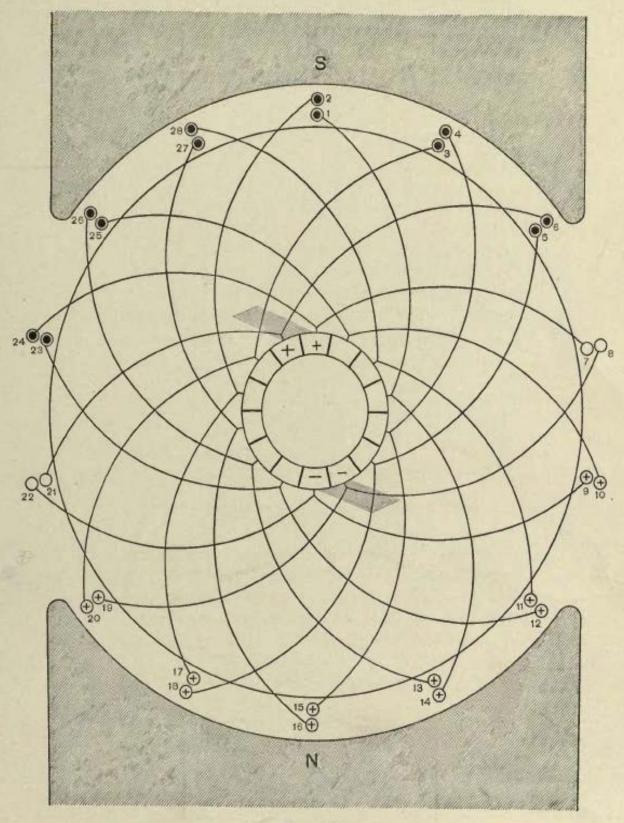
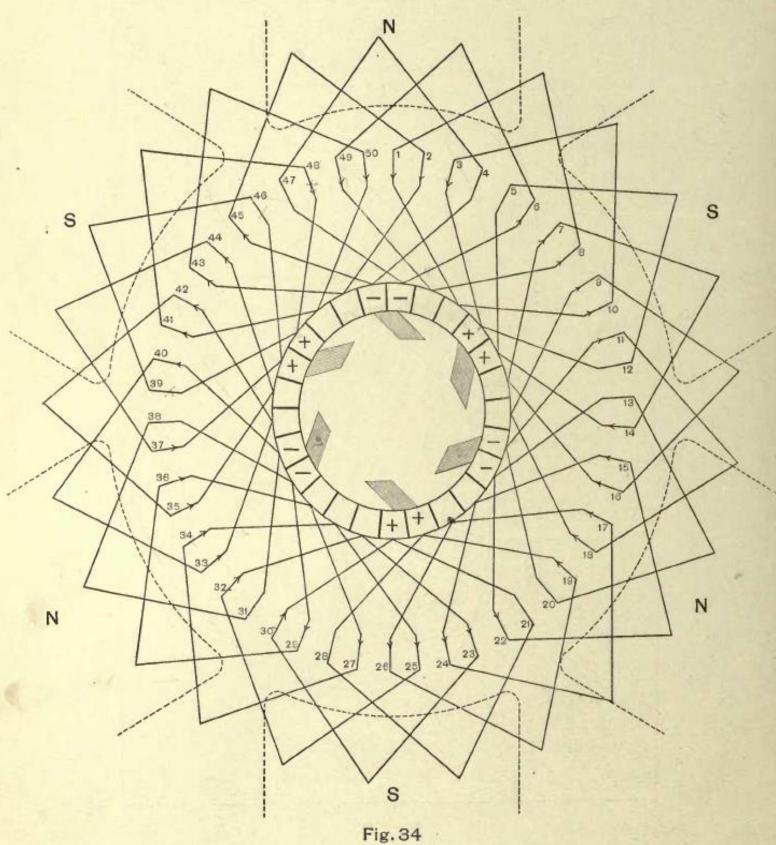


Fig. 33
TWO CIRCUIT, SINGLE WINDING.



SIX CIRCUIT, SINGLE WINDING.

## CHAPTER VI.

## MULTIPLE-CIRCUIT, SINGLE-WOUND, MULTIPOLAR DRUMS.

For multiple-circuit, multipolar drums, the condition to be fulfilled to make the winding re-entrant is that there shall be an even number of bars. The pitch at one end of the armature must exceed that at the other end by 2 (for single windings), each of these pitches being odd. If n is the number of poles and C the number of face conductors, the average pitch should not differ much from  $\frac{C}{n}$ ; for if it is much less, two successive conductors will often lie under the same pole piece, and their induced electromotive forces will be in opposition to each other, whereas they should be additive. If the average pitch is much greater than  $\frac{C}{n}$ , the cross-connections will be unnecessarily long, and the armature resistance and cost of copper unnecessarily high. Suppose a preliminary calculation for a single-layer, six-pole machine shows that about 49 conductors are required, it will be seen that  $\frac{C}{n} = \frac{49}{6} = 8.17$ . The two-end pitches must both be odd numbers, and must differ by 2. Therefore, take 7 and 9. The mean pitch is 8. The condition to be fulfilled by the total number of conductors is that it shall be an even number. Let it be 50.

This case is shown in Fig. 34. In this diagram the radial lines represent the face conductors. The connecting lines on the inside represent the end connections at the commutator end, and those on the outside represent the end connections at the pulley end. The brushes are placed inside the commutator for convenience.

At the position shown, the conductors without arrow-heads are short-circuited. The circuits through the armature are, -

$$\rightarrow \begin{cases} -\begin{cases} 6-49-8-1-10-3\\ 45-2-43-50-41-48 \\ -\begin{cases} 22-15-24-17-26-19\\ 11-18-9-16-7-14 \\ -\begin{cases} 40-33-42-35-44-37\\ 29-36-27-34-25-32-23-30 \end{cases} + \end{cases}$$

The front-end pitch is y=9, and the back-end pitch is y=-7.



If the pitches had been taken 7 and -5 instead of 9 and -7, retaining the same number (50) of face conductors, the diagram given in Fig. 35 would have been the result. This, it will be seen, is an application of the chord winding to a multipolar armature. The current in the conductors in the neutral zone is alternately in opposite directions, so that the demagnetizing action of the armature is small. The end connections are shorter, occupying less room and reducing the armature resistance and cost of copper. The short-circuited conductors are, however, at some distance from the neutral lines, and, although the electromotive forces in each pair will partly neutralize each other, it would be advisable, in cases where such chord windings are adopted, to have as great distances between pole tips as other circumstances permit.

In the given position, the short-circuited conductors are 4-49, 12-7, 20-15 28-23, 38-33, 46-41. The armature circuits are,—

The front-end pitch is y=7, and the back-end pitch y=-5.

If it should be considered desirable to have all the paths through the armature contain exactly the same number of conductors, then the number of face conductors should be chosen a multiple of the number of poles. But with a large number of conductors this would generally not be an important consideration.

In modern practice the conductors in large multipolar machines frequently consist of bars arranged in slots. The end connections then become strips arranged in two or more spiral layers at each end. If there were only one conductor per slot, two layers at each end would still be necessary, as it would be the same as if the lower conductors were brought up side of the upper conductors, and every other conductor would, therefore, as before, be connected over in an opposite direction from its neighbor.

For multiple-circuit, single-wound armatures there may be any even number of conductors per slot, and any number of slots. No new diagrams are necessary to show the cases of two or more conductors per slot, as Figs. 34 and 35 may be interpreted as having twenty-five slots and two conductors per slot, in which case odd-numbered conductors may be considered to belong to the upper layer, and even-numbered conductors to the lower layer. Connection is always made between odd and even numbered conductors, the pitch being always odd. The front-end and back-end pitches must differ by 2, and must have opposite signs.

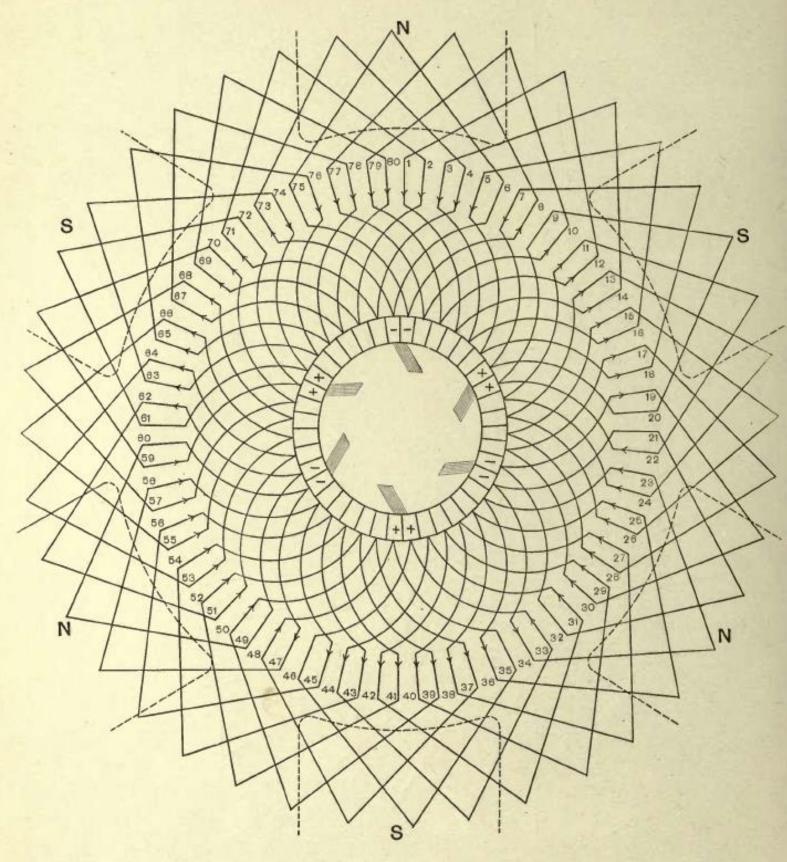


Fig. 36 SIX CIRCUIT, SINGLE WINDING.

Figure 36 represents a six-circuit, single-wound, drum winding with eighty conductors. The number of conductors is purposely taken large, so that a study of the diagram and winding table may show the magnitude of the differences of potential in neighboring conductors.

At the given position, conductors 75-6, 9-20, 21-32, 35-46, 49-60, and 61-72 are short-circuited at the brushes. The circuits through the armature are,—

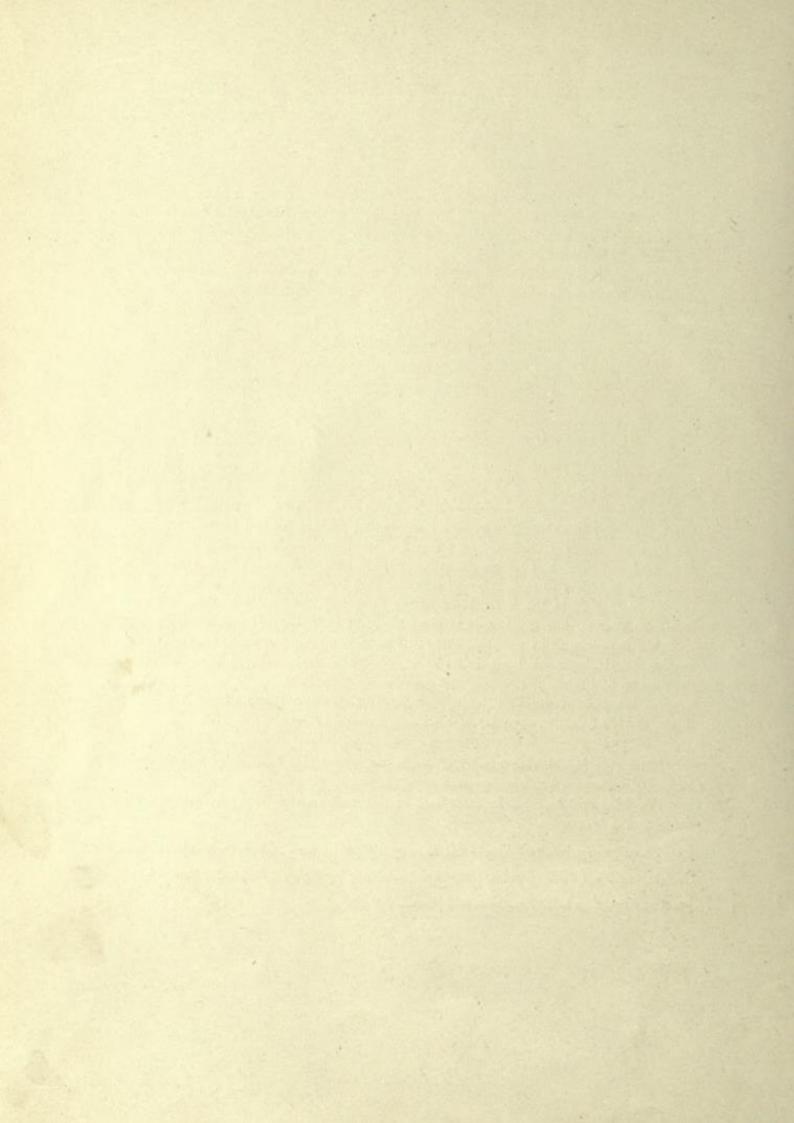
An inspection of the above table will show that the full difference of potential exists at recurring intervals between each pair of sequential conductors, such as 7 and 8, or 47 and 48. In practice, such conductors will often consist of two bars lying one above the other in the same slot. This shows that such upper and lower layers in a slot should be carefully insulated. On the other hand, alternately sequential conductors, as 5 and 7, or 47 and 45, have between them only the small difference of potential of two conductors in series; so that, in practice, where such conductors usually belong both to the upper or both to the lower layer of the same slot, comparatively thin layers of insulation suffice. For instance, it is often the case in multiple-circuit windings that there are four conductors per slot, arranged two wide and two deep. This case would require that the horizontal layer of insulation between conductors should be much thicker than the vertical layer.

For this class of windings (multiple-circuit, single-wound drums) a formula is superfluous, and the following summary of conditions will suffice:—

There may be any even number of conductors, except that in ironclad windings the number of conductors must also be a multiple of the number of conductors per slot.

The front and back pitches must both be odd, and must differ by 2; therefore the average pitch is even.

The average pitch "y" should not be very different from  $\frac{c}{n}$ , where c=number of conductors, and n=number of poles. For chord windings, "y" should be smaller than  $\frac{c}{n}$  by as great an amount as other conditions will permit.



### CHAPTER VII.

#### MULTIPLE-CIRCUIT, MULTIPLE-WOUND, MULTIPOLAR DRUMS.

THE next windings to be considered are multiple-circuit, multiple-wound, multipolar drums.

The following rules control these windings: -

The number of conductors, C, must be an even number. The pitches must be odd. If y=frontend pitch, then -(y-2m)=back-end pitch, where m=number of windings (double, triple, quadruple, etc.).

These "m" windings may form one re-entrant winding, "m" independent re-entrant windings, or a number of re-entrant windings equal to some factor of "m," each of which re-entrant windings is composed of two or more components.

To determine the proper number of conductors for any of the above cases, the following rule should be observed: —

If "m" equals the number of windings, and "C" equals the number of face conductors, then the number of independently re-entrant windings will be equal to the greatest common factor of  $\frac{C}{2}$  and m.

For instance, if a quadruple winding has 28 conductors, then the greatest common factor of (m=4) and  $\left(\frac{C}{2} = \frac{28}{2} = 14\right)$  is 2, and the quadruple winding will consist of two independent double windings, each of the two being re-entrant. This may be represented symbolically as  $\bigcirc$   $\bigcirc$ .

If C=24, and m=4, the greatest common factor of  $\left(\frac{C}{2}=\frac{24}{2}=12\right)$  and (m=4) is 4, and the quadruple winding will be made up of *four* independent single windings. This may be represented symbolically as  $\bigcirc\bigcirc\bigcirc\bigcirc\bigcirc$ .

If C=26 and m=4, the greatest common factor of  $\left(\frac{C}{2} = \frac{26}{2} = 13\right)$  and (m=4) is 1, and the quadruple winding will consist of *one* singly re-entrant quadruple winding. This may be represented symbolically as (m=4)

The above rule applies to any winding (double, triple, quadruple, etc.).

It is interesting to note that, for "multiple-circuit" windings, the rule for the number of multiple windings is independent of the number of poles and of the pitch.

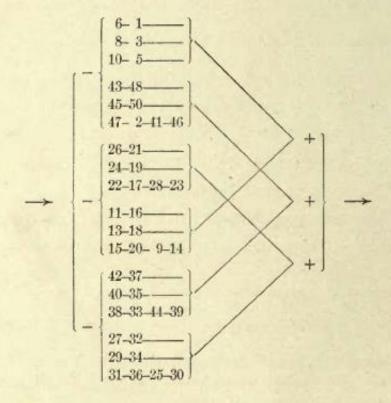
The number of conductors, "C," the average pitch, "y," and the number of poles, "n," should be so chosen that  $n \times y$  shall be somewhere nearly equal to C, being preferably a little smaller than C.

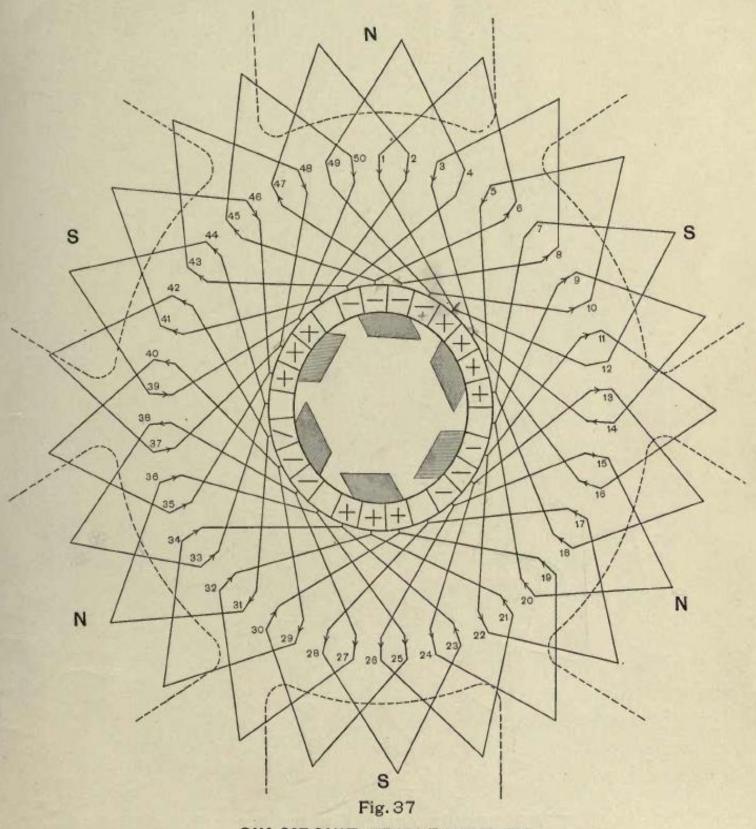


Figure 37 which, like Figs. 34 and 35, has six poles and fifty conductors, is a singly re-entrant triple winding. C=50; m=3. Greatest common factor of  $\frac{C}{2}$  and m is 1. Therefore, by the preceding rule, the result is one singly re-entrant triple winding. The winding may be represented symbolically as  $\bigcirc$ .

The average pitch should be a little less than  $\frac{C}{n} = \frac{50}{6} = 8.33$ , and the forward and backward pitches must differ by (2m=6). Therefore the front end pitch is taken y=11, and the back-end pitch y=-5.

In the given position, conductors 49 and 4 are short-circuited at a negative brush, and 12 and 7 at a positive brush. The circuits through the armature are,—





SIX CIRCUIT, TRIPLE WINDING.

8 poles -276 bors 138 alots



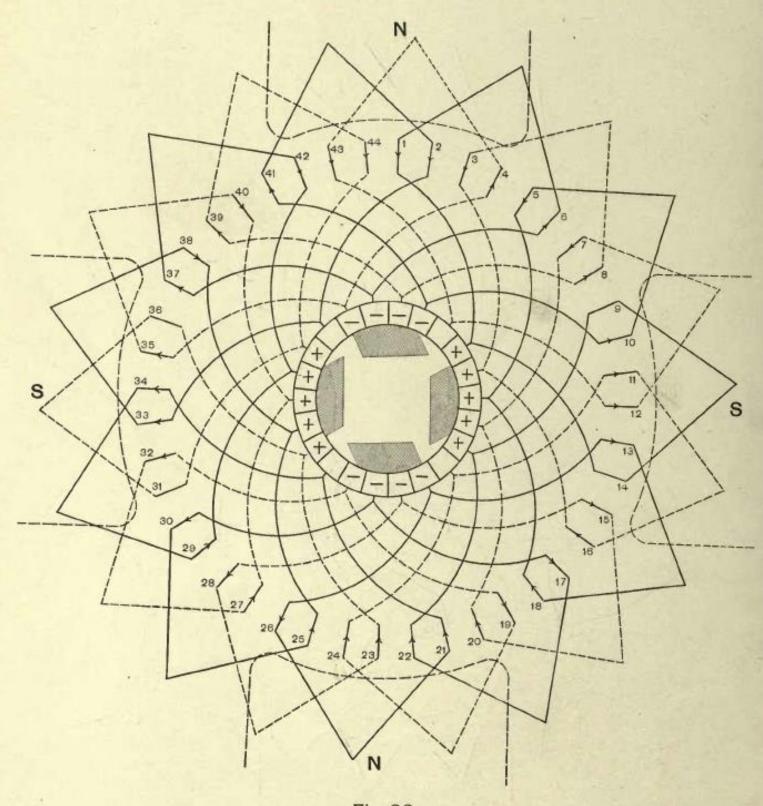
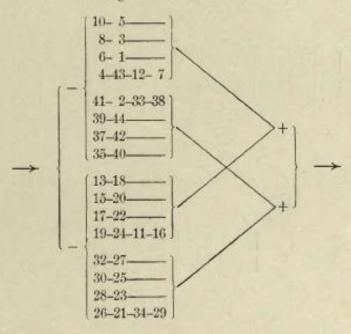


Fig. 38
FOUR CIRCUIT, QUADRUPLE WINDING.

Figure 38 is a four-circuit, doubly re-entrant quadruple winding in which n=4, C=44, and m=4. The greatest common factor of  $\frac{C}{2}$  and "m," i.e., of 22 and 4, is 2; therefore there are two independent, singly re-entrant, double windings. The winding may be represented symbolically by  $\bigcirc$   $\bigcirc$ . These two windings are represented on the diagram by full and dotted lines. The front-end pitch has been taken 13, and the back-end pitch -5, the difference being necessarily 2m=8. Inspection will show that the two windings are, —

In the given position, 9-14 and 31-36 are short-circuited at the positive brushes. The circuits through the armature are,—



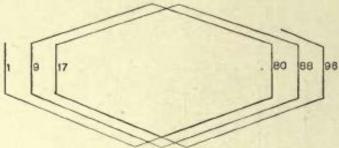
The extreme irregularity exhibited in the diagrams and tables of the multiple windings is due to the necessarily small numbers of conductors chosen. With the magnitudes taken in practical work, everything will be sufficiently regular.



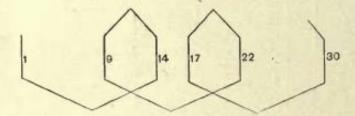
Figure 39 is the same quadruple winding as Fig. 38, except that the pitches are taken 15 and -7 instead of 13 and -5. This was drawn to emphasize the fact that there is nothing absolute in the choice of the pitch in these multiple circuit armatures, except that in the case of the multiple windings, the numerical differences between the forward and backward pitches must be equal to 2 m, where "m" is the number of windings. As before stated, the average pitch should not differ much from  $\frac{C}{n}$ , and should be somewhat less, rather than greater.

Figure 38, which partakes in a small degree of the nature of the short chord windings (as compared with Fig. 39), has a very much larger percentage of the conductors subjected to counter-induction than would be the case in actual practice with large numbers of conductors.

For instance, the average pitch might often be represented by some such number as 75. If it were to be a quadruple winding, the two pitches should differ by 2m or 8. Therefore the forward pitch would be taken 79, and the backward pitch -71, so that the order of the winding would be 1-80-9-88, etc., whereas in the case of small numbers of conductors, such as in Fig. 38, the order of the winding was 1-14-9-22-17-30, etc. It will be evident that the distinction between these two cases is, that with the larger number of conductors there are many forward and backward steps before the original loop is crossed, thus:—



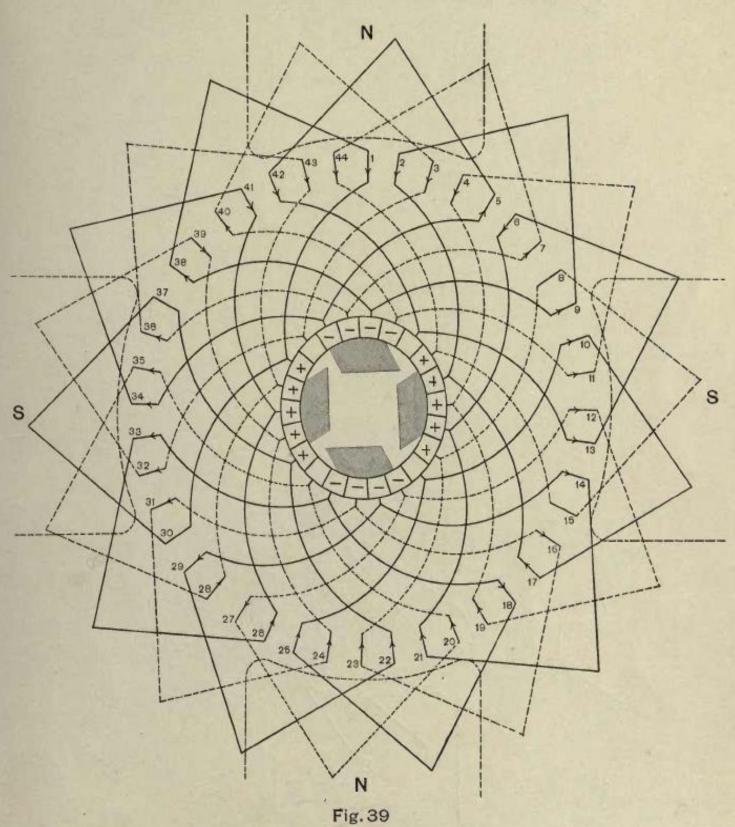
But in the ease of the small number of conductors the loop is crossed almost at once, thus: -



In other words, with multiple windings and small numbers of conductors, the numerical differences between the forward and backward pitches is a large percentage of the average pitch, whereas with the large numbers of conductors used in practice, it is a very small percentage of the average pitch.

The fact that irregularities are much exaggerated by the necessary choice of rather small numbers of conductors should be borne in mind in the study of these diagrams, particularly those of multiple windings.

If, instead of the quadruple windings consisting of two independent doubly re-entrant windings of Figs. 38 and 39, one singly re-entrant quadruple winding is desired, a number of conductors must be



FOUR CIRCUIT, QUADRUPLE WINDING.



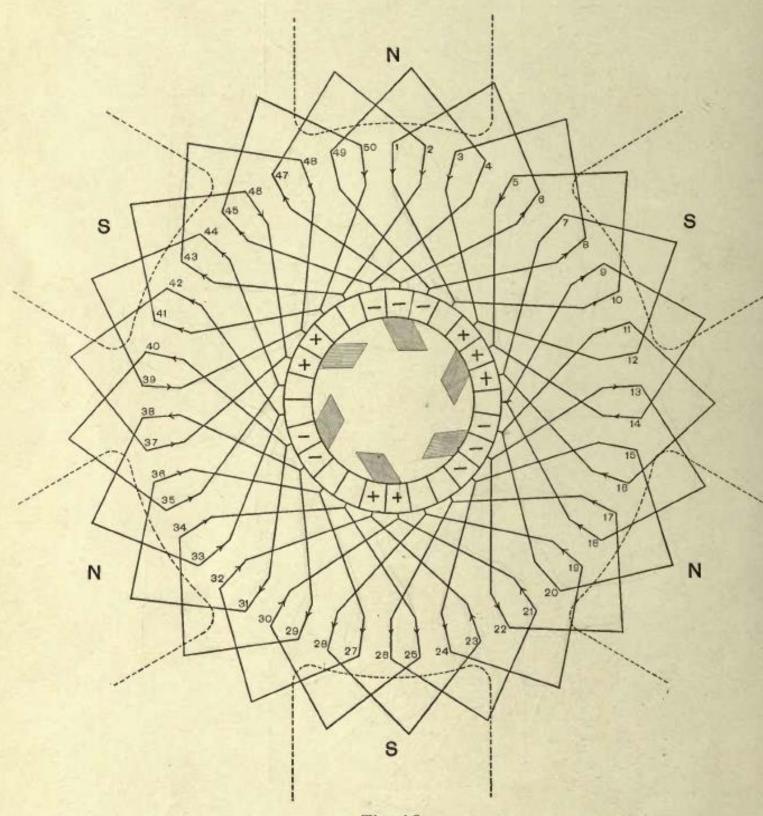


Fig. 40 SIX CIRCUIT, DOUBLE WINDING.

chosen such that  $\frac{C}{2}$  and "m" (4) shall be mutually prime. Take C=42. Then  $\frac{C}{2}=21$ , and m=4, which are mutually prime. If the forward pitch is taken y=13, and the backward pitch y=-5, the winding will be,—

$$1-14-9-22-17-30-25-38-33-4-41-12-7-20-15-28-23-36-31-2-39-10-5-18-13-2\ell-21-34-29-42-37-8-3-16-11-24-19-32-27-40-35-6-1$$

This would be represented symbolically as (OOO) and would be a singly re-entrant quadruple winding.

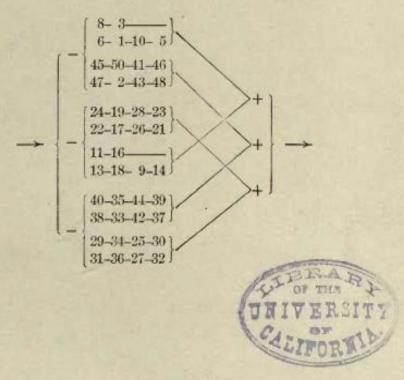
If four entirely independent windings are desired,  $\frac{C}{2}$  and "m" must have 4 for their greatest common factor. Taking C=40, and making the front and back pitches respectively y=13 and y=-5, the winding would be,—

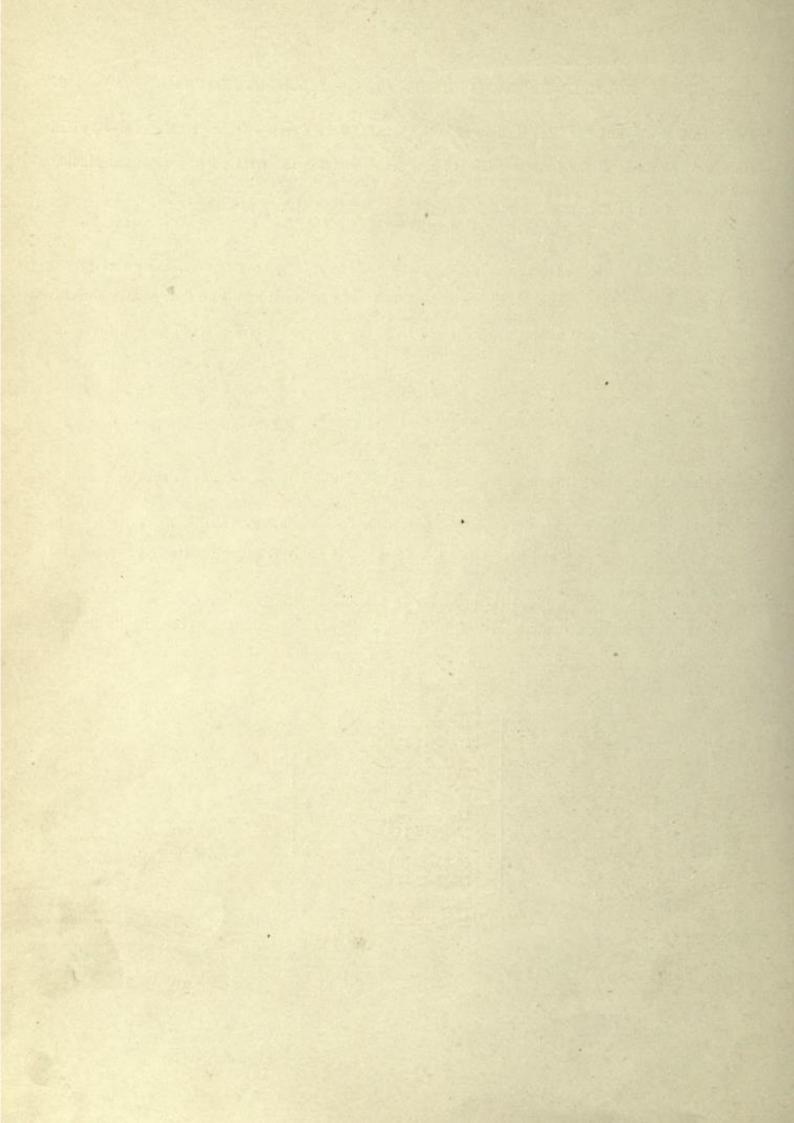
This could be represented symbolically as \( \rightarrow \rightarr

In Fig. 40 is shown a six-circuit, singly re-entrant, double winding. C=50, n=6, m=2. The greatest common factor of  $\frac{C}{2}$  and "m" being 1, the winding is singly re-entrant, and may be represented symbolically as  $\odot$ .

The forward pitch is y=9, and the backward pitch is y=-5.

In the given position, conductors 49-4, 7-12, and 15-20 are short-circuited. The circuits through the armature are,—







# CHAPTER VIII.

#### TWO-CIRCUIT, SINGLE-WOUND, DRUM ARMATURES.

The "two-circuit" windings now to be considered are distinguished by the fact that the pitch is always forward, instead of alternately forward and backward, as in the "multiple-circuit" windings, just described.

The sequence of connections leads the winding from a certain bar opposite one pole piece to a bar similarly situated opposite the next pole piece, and so on, so that as many bars as pole pieces are passed through before another bar in the original field is reached. Such progression around the armature is continued until all the bars are connected in, and the winding returns on itself.

Two-circuit, drum windings, like the two-circuit, gramme-ring windings, have for a given voltage the fraction  $\frac{2}{n}$  as many conductors as multiple-circuit windings, with the attendant advantages, stated for the two-circuit, gramme-ring windings. The advantages, that the circuits from brush to brush consist of conductors influenced by all the poles, are — when there is but one turn in each coil — the same as in the two-circuit, short-connection ring winding. When there are several turns in the coil, the advantages are subject to the same reservations as in the two-circuit, long-connection, ring winding. The advantages, due to such arrangements of the conductors, have been confined to machines of small electrical output. In machines of large electrical output, in which there are a number of sets of brushes of the same sign (otherwise the cost of the commutator is excessive), the advantages possible from equal currents in the circuits have been overbalanced by the increased sparking due to unequal division of the current between the different sets of brushes of the same sign.

An examination of the diagrams will show that in the two-circuit windings the drop in the armature, likewise the armature reaction, is independent of any manner in which the current may be subdivided among the different sets of brushes, but depends only upon the sum of the currents at all the sets of brushes of the same sign. There are, in the two-circuit windings, no features that tend to cause the current to subdivide equally between the different sets of brushes of the same sign, and, in consequence, if there is any difference in contact resistance between the different sets of brushes, or if the brushes are not set with the proper lead with respect to each other, there will be an unequal division of the current.

When there are as many sets of brushes as poles, the density at each pole must be the same, otherwise the position of the different sets of brushes must be shifted with respect to each other to correspond to the different intensities, the same as in the multiple-circuit windings.

In practice it has been found difficult to prevent the shifting of the current from one set of brushes to another. The possible excess of current at any one set of brushes increases with the number of sets; likewise the possibility of excessive sparking. For this reason the statement has been sometimes made that the disadvantages of the two-circuit windings increase with the number of poles.

From the above, it may be concluded that any change of the armature with respect to the poles will in the case of two-circuit windings be accompanied by shifting of the current between the different sets of brushes; therefore to maintain a proper subdivision of the current the armature must be maintained in one position, with respect to the poles, and with exactness, since there is no counter action in the armature to prevent the unequal division of the current.

In the case of multiple-circuit windings, it will be noted that the drop in any circuit, likewise the armature reaction in the field in which the current is generated, tends to prevent the excessive flow of current from the corresponding set of brushes. On account of these features, together with the consideration that when there are as many brushes as poles the two-circuit armatures require the same nicety of adjustment with respect to the poles as the multiple-circuit windings, the multiple-circuit windings are generally preferable, even when the additional cost is taken into consideration.

Denoting the number of face conductors by "C," the number of poles by "n," and the average pitch by "y," the formula controlling the two-circuit, single-wound, multipolar drum, is,—

$$C = ny \pm 2$$
.

It is preferable to have the pitch "y" the same at the two ends, because the two sets of end connections will then be of the same length, but the choice of the number of conductors "C" for any particular case is less restricted (when the number of poles is greater than four) if the front and back pitches are permitted to differ by 2. Each pitch, must, moreover, be an odd number, as, in order that the winding may pass through all the conductors before returning upon itself, it must pass alternately through odd and even numbered conductors. Also when, as is usually the case, the bars occupy two layers, it is necessary to connect from a conductor of the upper to one of the lower layer so as to obviate interference in the positions of the spiral end connections. Where different pitches are used at the front and back ends, each being odd, the average "y" appearing in the formula will be even.

In Fig. 41 is given a two-circuit, single winding for a four-pole drum. The pitch is y=9 at both ends.

$$C = ny \pm 2 = 4 \times 9 \pm 2 = 34$$
 or 38.

Thirty-four conductors were taken. If it is necessary to have thirty-four conductors, it would be better to take the average "y" equal to eight, and then to use y=9 at one end and y=7 at the other. It is thus possible to shorten the end connections at the end at which the shorter pitch is used, and thus avoid using an unnecessary amount of copper. This will also make the armature resistance less, and will give more room for the end connections.

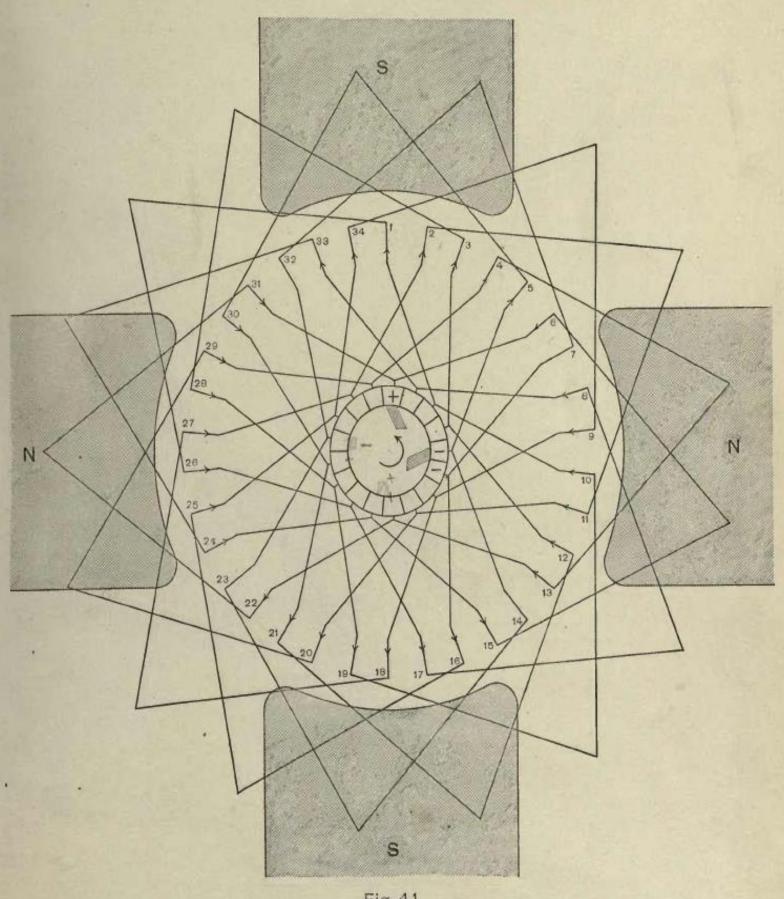


Fig. 41,
TWO CIRCUIT, SINGLE WINDING.

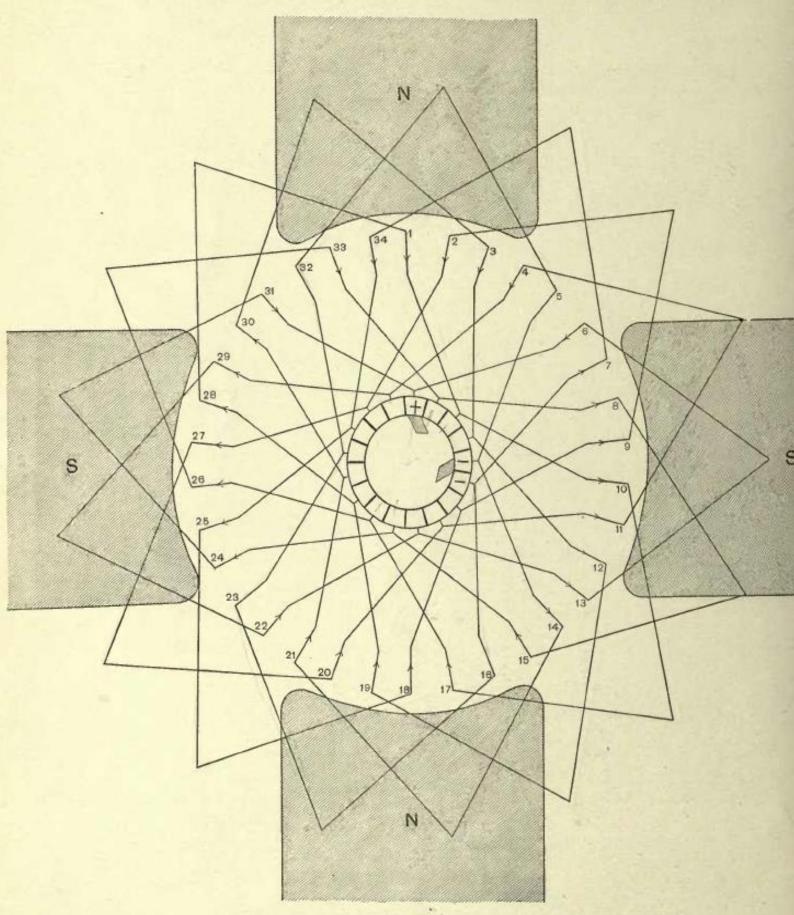


Fig. 42
TWO CIRCUIT, SINGLE WINDING,

In Fig. 42 this has been done, the front-end pitch being y=9 as before, but the back-end pitch being y=7. The average pitch is y=8.

$$C = ny \pm 2 = 4 \times 8 \pm 2 = 30$$
 or 34.

Thirty-four conductors have been taken.

If thirty-eight conductors should be preferable to thirty-four, then the best arrangement would be to use y=9 at both ends.

$$C = ny \pm 2 = 4 \times 9 \pm 2 = 34$$
 or 38.

This case has not been drawn, but it would be the proper method for thirty-eight conductors, as the only other way would be to have a front-end pitch y=11 and a back-end pitch y=9, giving an average pitch y=10.

$$C = ny \pm 2 = 4 \times 10 \pm 2 = 38$$
 or 42.

This last choice, i.e. pitches of 9 and 11, would be undesirable, as the connections at the end with a pitch of 11 would be unnecessarily long. Therefore, as a general rule, the pitch should be chosen a little less than  $\frac{C}{n}$ , and when this would result in an even pitch, the pitch at one end may be made (y+1) and at the other end (y-1). Of course, the advantage of having both sets of end connections exactly equal might offset the small saving in material. This would have to be determined for the case in hand. Often, however, even where the same pitch is used at both ends, other considerations make it necessary to use two differently proportioned sets of connecting strips.

This matter of the possibility of using two different pitches, so that the "y" of the equation  $C=ny\pm 2$  may be any integer, odd or even, is not so very important in the case of four-pole armatures, as it does not increase the range of choice of conductors. But for six, eight, and higher numbers of poles the introduction of even integers for "y" gives many more possible numbers of conductors than if it were necessary to be

confined to odd integers.

Thus, for the case of six-pole windings, the formula  $C = ny \pm 2$  becomes  $C = 6y \pm 2$ . If "y" is put successively equal to 10, 11, 12, 13, 14, and 15, the possible numbers of bars will become as follows:—

$$y=10$$
  $C=60\pm2=58$  or  $62$   
 $y=11$   $C=66\pm2=64$  or  $68$   
 $y=12$   $C=72\pm2=70$  or  $74$   
 $y=13$   $C=78\pm2=76$  or  $80$   
 $y=14$   $C=84\pm2=82$  or  $86$   
 $y=15$   $C=90\pm2=88$  or  $92$ .

Thus it may be seen that if it were only permissible to use odd integers for "y," the possible conductors for this range would be limited to 64, 68, 76, 80, 88, and 92; but by using unequal pitches at the two ends, the average "y" becomes even, and the possible numbers of conductors to which the choice is limited is doubled. It is very important that this point should be borne in mind, as the rule often used for four-pole machines that C must equal number of poles times an odd number, plus or minus two, is sometimes mistakenly extended to larger numbers of poles, and a number of conductors is chosen either larger or smaller than is desired; whereas, if different pitches at the two ends had been used, a much more suitable choice might have been made.

Another limiting consideration is, that the numbers of conductors <u>per slot</u> is governed largely by the capacity and voltage of the machine, so that sometimes two, sometimes four, and in exceptional cases even six or eight, bars might be desired per slot, therefore, the total number of conductors "C" must be a multiple of 2, 4, 6, or 8, as the case may be. If, in the case of a six-pole armature, only two conductors per slot are desired, the pitch may be either odd or even; but it will be found that where four conductors per slot are wanted, and where, therefore, "C" must be a multiple of 4, that only the numbers of conductors obtained by making "y" an odd integer meet the requirement. And if six conductors per slot should be wanted (and it seldom would be, because the mechanical fitting of the connections would be so troublesome), neither the use of an odd nor of an even integer would (in the case of a six-pole armature) give a possible number for "C."

In the following illustrative diagrams it will not be necessary to take pains to show how many conductors there are per slot. They will be drawn with the conductors spaced at equal intervals, and one, two, four, or more, as desired, may be supposed to be brought together in a slot.

In Fig. 43 is given a diagram for a six-pole, two-circuit, single-wound, drum armature. The pitch is y=11 at both ends.

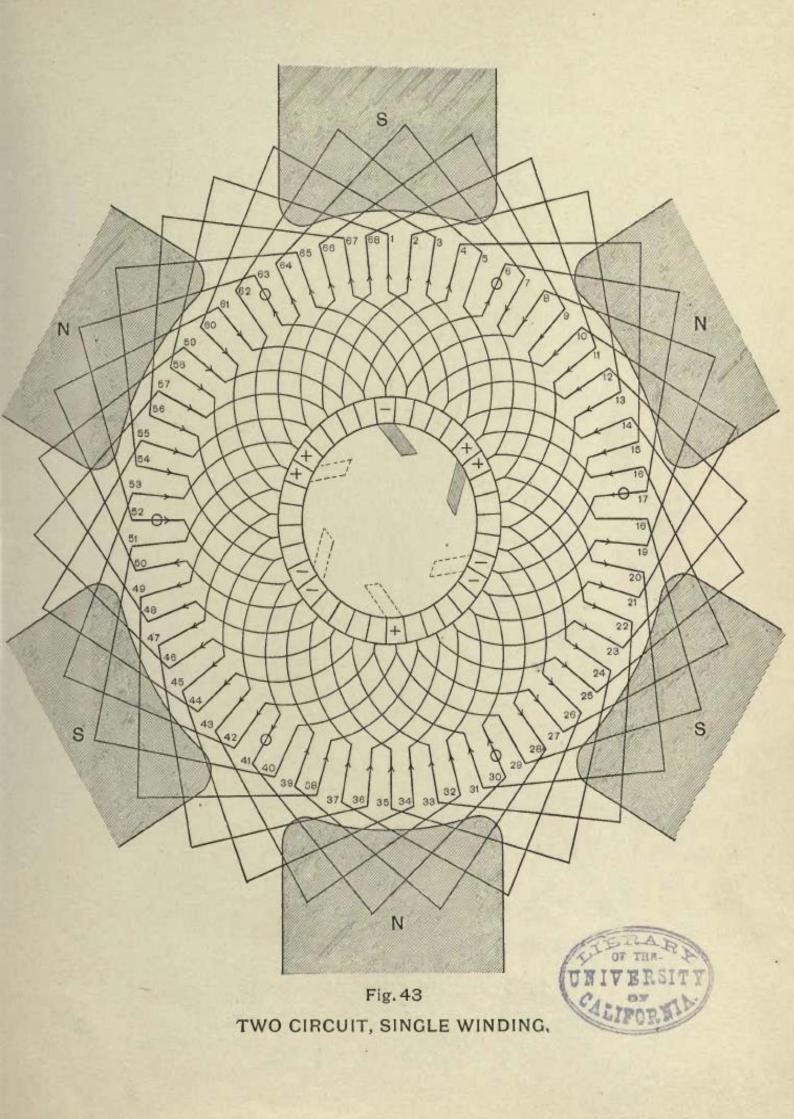
$$C = ny \pm 2 = 6 \times 11 \pm 2 = 66 \pm 2 = 64$$
 or 68.

Sixty-eight conductors were taken, and they could be arranged one, two, or four per slot, as other conditions might determine.

In the position shown, the positive brush short-circuits the group of conductors 5-62-51-40-29-18, all in series. The circuits through the armature are,—

An examination of the preceding table will show that immediately sequential conductors, such as 6 and 7, have between them, at recurring periods, the full difference of potential of the winding. But alternately sequential pairs of conductors, as 6 and 4, or 63 and 65, have between them only the difference of potential of "n" bars.

For the above analysis, only the two full-lined brushes were supposed to be in service. If, however, the four brushes shown by the dotted lines were added, the short-circuited bars would consist of groups of two each, in series between different brushes of like sign. In the given position, these groups would be 5-62, 51-40, and 29-18 at the positive brushes, and 63-52, 41-30, and 17-6 at the negative brushes. The circuits through the armature would be the same, with the exception that the brus short-circuited by the negative brushes would now disappear from the list. These six conductors, 6, 17, 63, 52, 41, 30, have been underlined in the table, and are marked on the diagram by small circles.



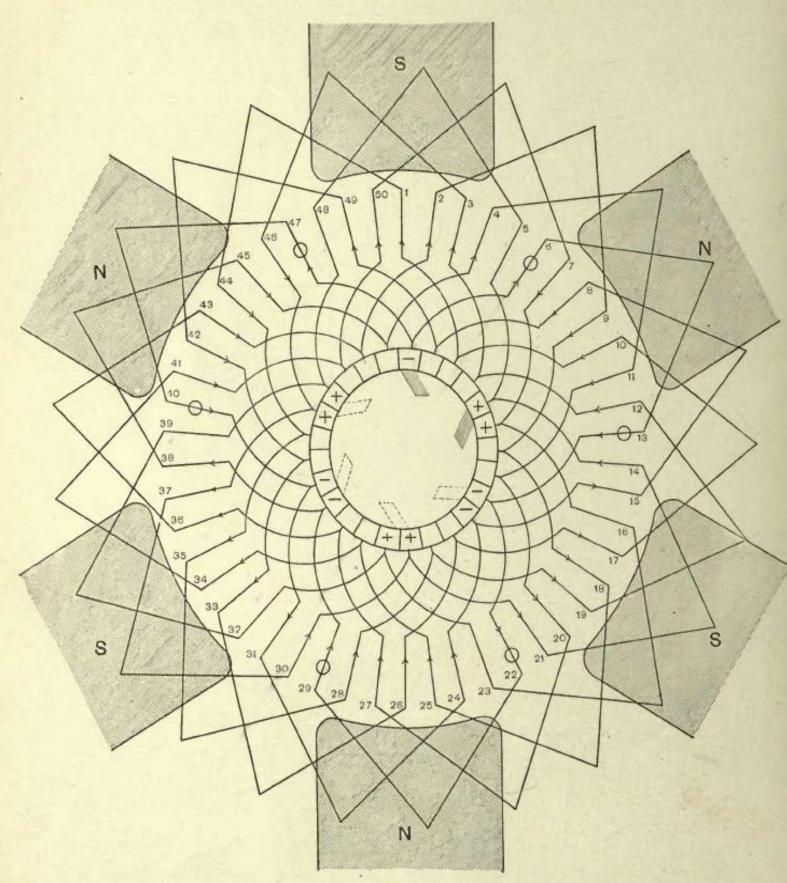


Fig. 44
TWO CIRCUIT, SINGLE WINDING.

In Fig. 44 is given a diagram for a two-circuit, six-pole armature. The back-end pitch is y=7, and the front-end pitch is y=9. Therefore the average pitch is y=8.

$$C = ny \pm 2 = 6 \times 8 \pm 2 = 46$$
 or 50.

Fifty conductors are taken. As in the preceding diagram, only the six conductors without arrow-heads are short-circuited when the two full-line brushes alone are active. But when all six brushes bear on the commutator, the conductors designated by small circles are also short-circuited.



### TWO-CIRCUIT WINDINGS WITH CROSS-CONNECTED COMMUTATORS.

Figures 45, 46, 47, and 48 are illustrative of a class of two-circuit windings that possess the distinctive feature that the number of coils may bear a relation to the number of poles not possible with the other two-circuit windings described. An examination of the diagrams will show that the different coils of a winding may be subdivided in groups, each group having either as many coils as there are pairs of poles, or half as many, these different groups being connected in series by a cross-connected commutator.

Figure 45 is an example of this class. As will be seen, it consists of an eight-pole drum armature, with fifty-six conductors connected up as a two-circuit, single winding.

The underlying principle is best understood by noting one "element" of the winding, such as the eight polar conductors drawn with very heavy lines. It starts from a certain commutator segment, and after proceeding under each of the eight pole pieces, it returns to the adjacent segment. It should be further observed that, unlike the heretofore described two-circuit drum armatures, the conductors of this element are separated from each other by an angular distance equal exactly to  $\frac{360}{8} = 45^{\circ}$ , instead of, as in the ordinary two-circuit drum windings, being separated by an angular distance a little greater or less than this.

$$C=56$$
,  $n=8$ , y (the "pitch")= $\frac{56}{8}=7$ .

It should be particularly noted that, with this winding, a number of conductors is used which is an exact multiple of the number of poles. This, of course, is not possible with the ordinary two-circuit drum windings, which are controlled by the formula —

$$C = ny \pm 2$$
.

As will be seen from the diagram, this winding requires cross-connection of the commutator, but in many machines this disadvantage might be offset by the fact that, owing to the symmetrical arrangement of the conductors with reference to the pole pieces, the objectionable "selective commutation" of the ordinary type would probably be avoided.

To return to a study of the diagram, it will be seen that there are  $\frac{C}{n} = \frac{56}{8} = 7$  sets of "elements" exactly the same as that above described, except that each is located at an angular distance of  $\frac{360}{7}$  from the preceding one. To facilitate comprehension of the diagram, these seven "elements" have been drawn in with different styles of lines, and are readily distinguishable.

It is therefore obvious that, if it were not for the commutator cross-connections, the winding would consist of seven sets of eight conductors each, and that each such set has its two terminals at a pair of adjacent segments. These individual coils are put in the proper series relation between brushes by the commutator cross-connection. The resultant design is perfectly symmetrical.

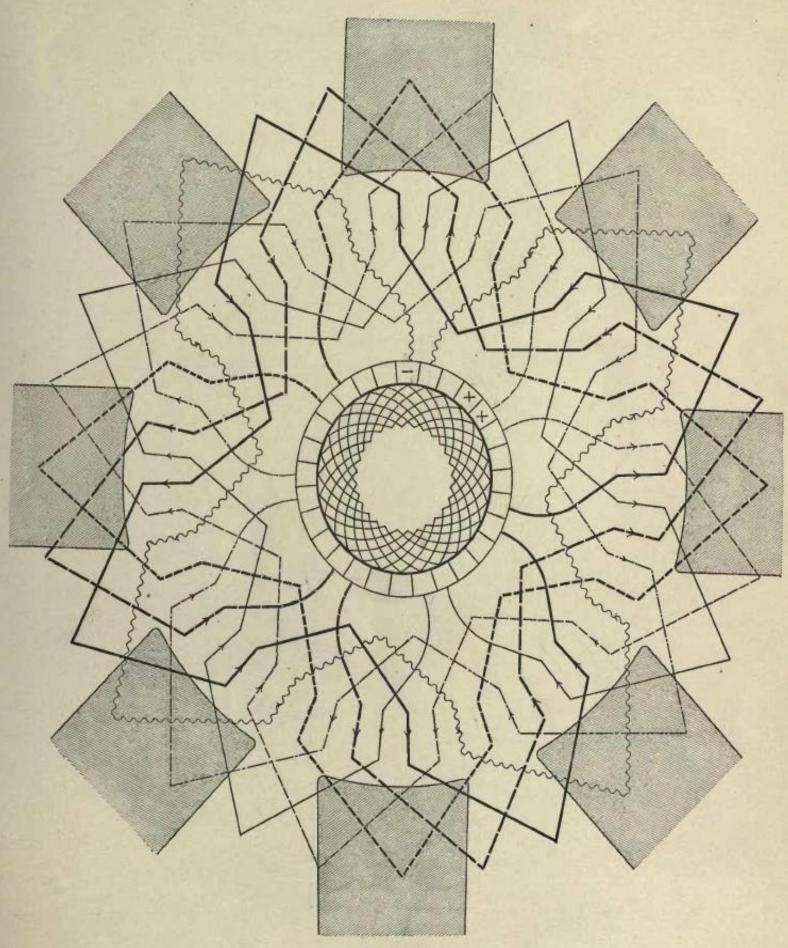


Fig. 45

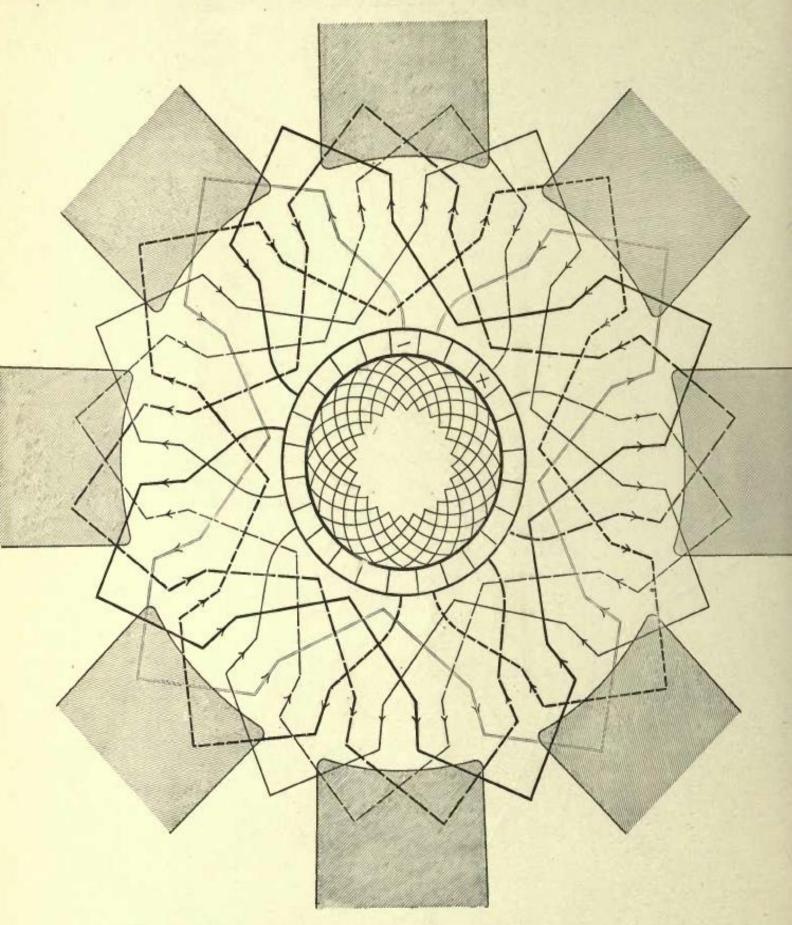


Fig. 46

Figure 46 differs only in having forty-eight conductors, with the necessary consequence that, the pitch being even  $\binom{48}{8}=6$ ), it has to be different at the front and back. It is seven at the commutator end, and five at the other end. This slight irregularity makes the wording of the description of Fig. 45 not absolutely applicable to this diagram, the chief difference being that, although every pair of successive conductors are exactly similarly located with respect to a pair of poles as every other pair, the same cannot be said of every individual conductor of an element, the distance between them being successively greater and less than  $\binom{360}{8}$ .

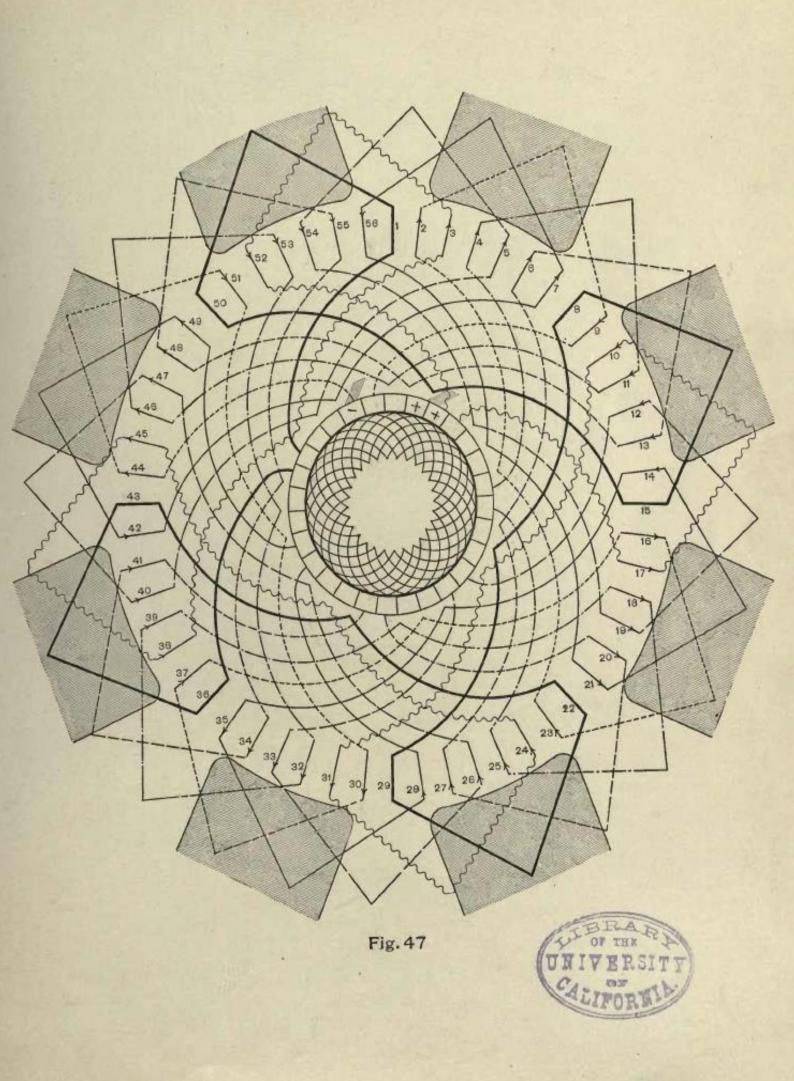


Figure 47 represents a two-circuit single-winding, identical with Fig. 45, except that the connecting leads at the front end are twice as long.

This is used in some "form" windings, where the two ends of a coil are brought out in front at a point half-way between the two slots holding the wires of a coil. The long front connections would never be used in bar windings, where each face conductor of the diagram represents only one conductor, for it would be a waste of copper. Short leads such as those of Fig. 45 would, for such bar windings, always be used.

An "element" of the winding may be readily seen from the heavy lining in the diagram.

Windings of same type as Fig. 47 could be made corresponding to Fig. 46, as well as to Fig. 45. In fact, the underlying principle of this winding is identical with that of the type illustrated by Figs. 45 and 46.



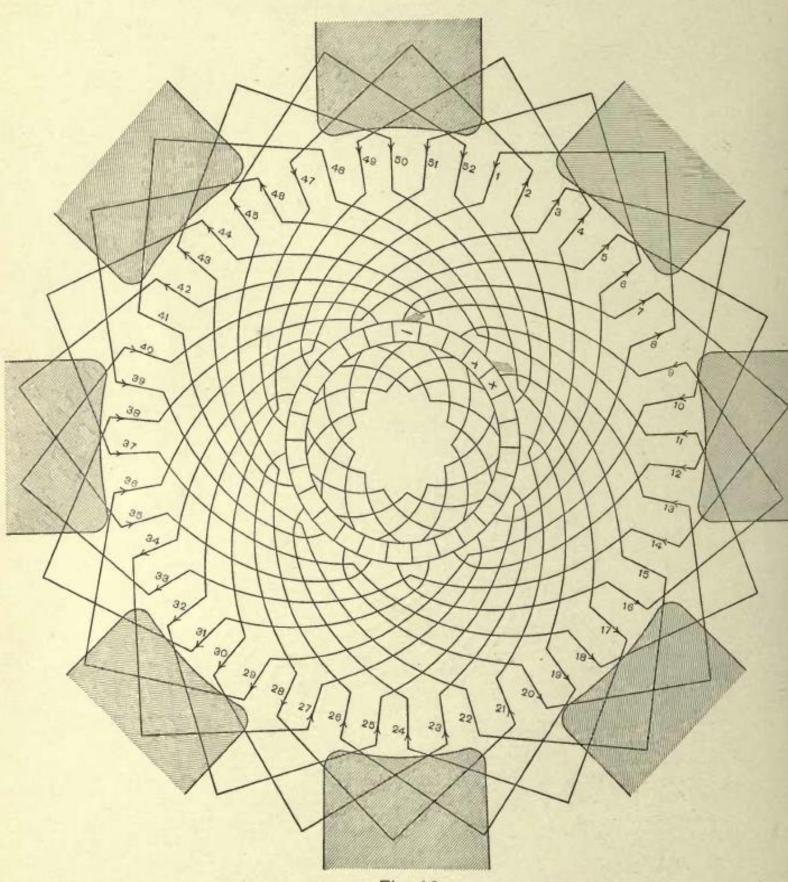


Fig. 48

Figure 48 represents a two-circuit single winding for an eight-pole machine, in which four conductors constitute an element. The number of conductors is here taken to be fifty-two. There are therefore  $\frac{5}{4}^2 = 13$  elements. It is a condition of this winding that the number of elements must be an odd number. From this it follows that the total number of conductors cannot be a multiple of the number of poles.

It serves, therefore, for numbers of conductors with which the previously described winding (where C is a multiple of n) could not be used. It probably, however, would not be so well balanced as in the case where C is a multiple of n. The commutator requires cross-connecting, as shown in the diagram. The cross-connections at the front end are of twice the usual length.



# WENSTRÖM TWO-CIRCUIT, WIRE-WOUND ARMATURE.

Figure 49 represents a winding devised by Wenström to lessen the depth of the end windings of wire wound armatures.

The particular case represented by the diagram had thirty-five lozenge-shaped slots, each containing four conductors. For the sake of clearness only the connections of the wires between two adjacent commutator segments are shown, and no difficulty will be found in completing the winding, by continuing on through the remaining segments.

This method is, of course, only suitable for wire-wound armatures and like most such wire windings, it is difficult to repair.

It is to be noted that these armatures, which have been quite extensively used, were completely ironclad, there being no slot opening.

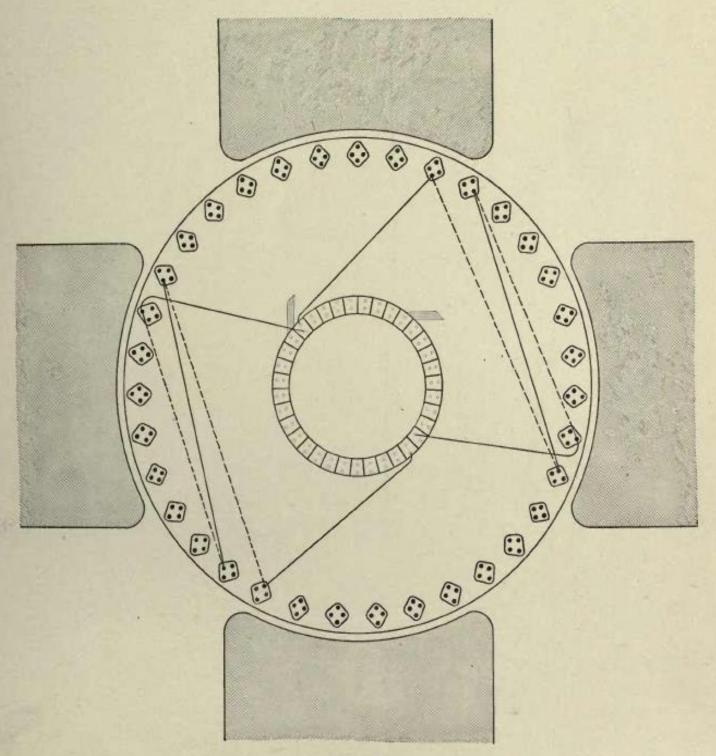


Fig. 49

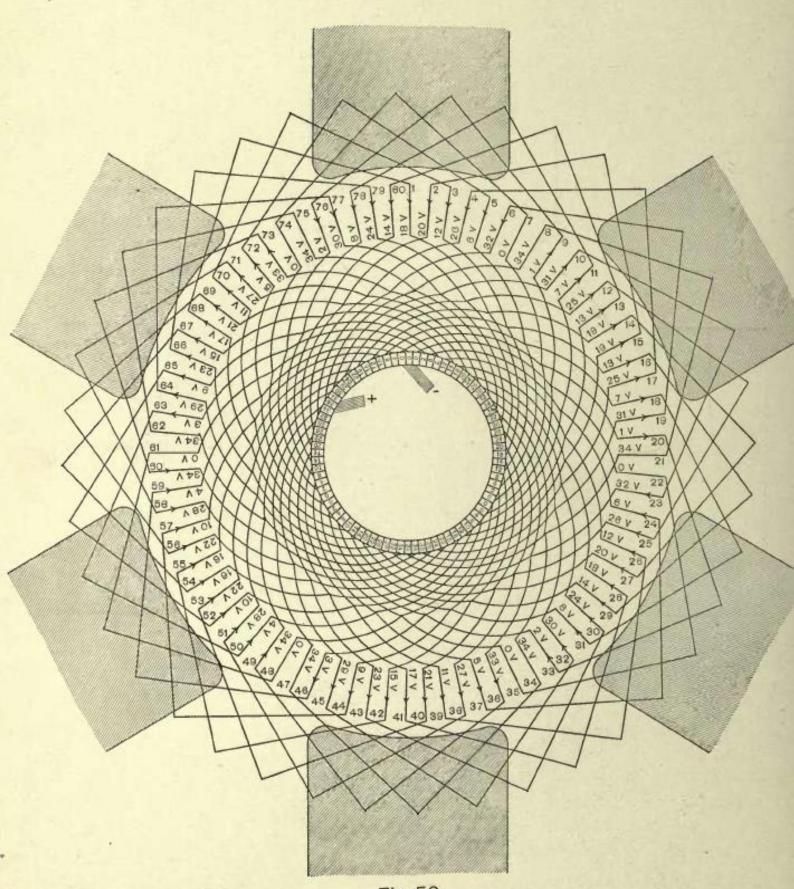


Fig. 50
TWO CIRCUIT, SINGLE WINDING.

## CHAPTER IX.

#### INTERPOLATED COMMUTATOR SEGMENTS.

In Fig. 50 is given a two-circuit single winding. n=6, y=13,  $C=ny\pm 2=6\times 13\pm 2=76$  or 80. Eighty conductors have been taken. This would naturally give forty commutator segments. Suppose speed, strength of field, and active length of conductors to be of such magnitudes as to generate one volt per conductor. Noting that, as shown in the figure, twelve conductors are short-circuited, there will be  $\frac{80-12}{2}=34$  active conductors in series between brushes. Therefore the total E.M.F. will be 34 volts. There would be (before interpolating)  $\frac{40-6}{6}=5.67$  segments between every two neutral points of the commutator. Therefore there would be  $\frac{34}{5.67}=6$  volts between every two adjacent segments.

Suppose this to be higher than is desired. It might then be proposed to double the number of segments by the method of cross-connecting shown in Fig. 50. This will increase the number of segments to eighty. Following the circuit through from the negative to the positive brush, the conductors have been labeled 1 volt, 2 volts, 3 volts, etc., adding one volt for each conductor. Taking the potential of the negative brush as zero, this gives the potential of each conductor. Following down from each conductor to its attached segments, they have been numbered in a corresponding manner; thus the four segments connected to the two bars at 20 volts potential have been marked 20, etc.

An examination of the figure will now make it apparent that proceeding from the neutral points (at zero potential) the voltage increases alternately by two and by four volts per segment, the average being three volts per segment. Therefore, although the average volts per segment have been decreased to one-half of what they were for forty segments, half of the segments have between them only one-third, and the remainder, two-thirds, of the original volts per segment. Therefore, for a six-pole armature, the volts per segment cannot be halved by interpolation. And in order to reduce them to one-third throughout, it is not sufficient to cross-connect as shown in the figure, but it is necessary to triple the natural number of segments and cross-connect every three corresponding segments. This would be far from simple.



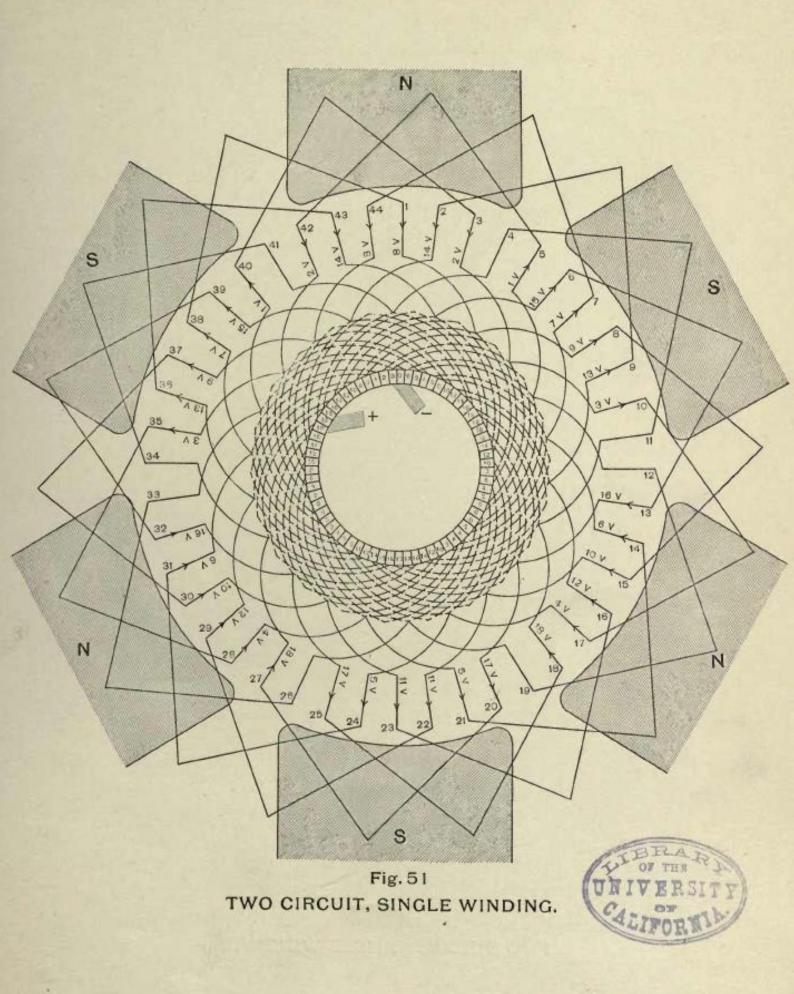
A fairly large number of conductors was taken in Fig. 50, in order to give a thorough explanation of the principles involved in interpolating segments. The further study of the subject can, however, be more satisfactorily carried on with small numbers of conductors.

In Fig. 51 is shown another two-circuit, single winding, with n=6, y=7,  $C=ny\pm 2=6\times 7\pm 2=40$  or 44. Forty-four conductors are taken. Without interpolation, twenty-two segments would be used. Here  $3\times 22=66$  segments are used. This is arrived at by connecting together every three corresponding commutator segments.

If, as in the preceding figure, only two segments had been cross-connected, the connections shown by the full lines would have sufficed. Cross-connecting every three corresponding segments involved the addition of the dotted line connections. This, as the diagram shows, doubles the total number of commutator cross-connections, and is therefore mechanically objectionable.

But the volts between bars are now everywhere equal instead of being alternately V and 2V as in Fig. 50. This may be seen by an examination of the numbers on the conductors and segments, which have been arranged according to the conventional method described.

Thus, proceeding from the segments under the negative brush, the voltage would increase regularly by two volts per segment up to the positive brush, so that whereas, in the former cases, the order was 2, 4, 8, 10, 14, 16, etc., it is now 2, 4, 6, 8, 10, 12, 14, 16, etc.



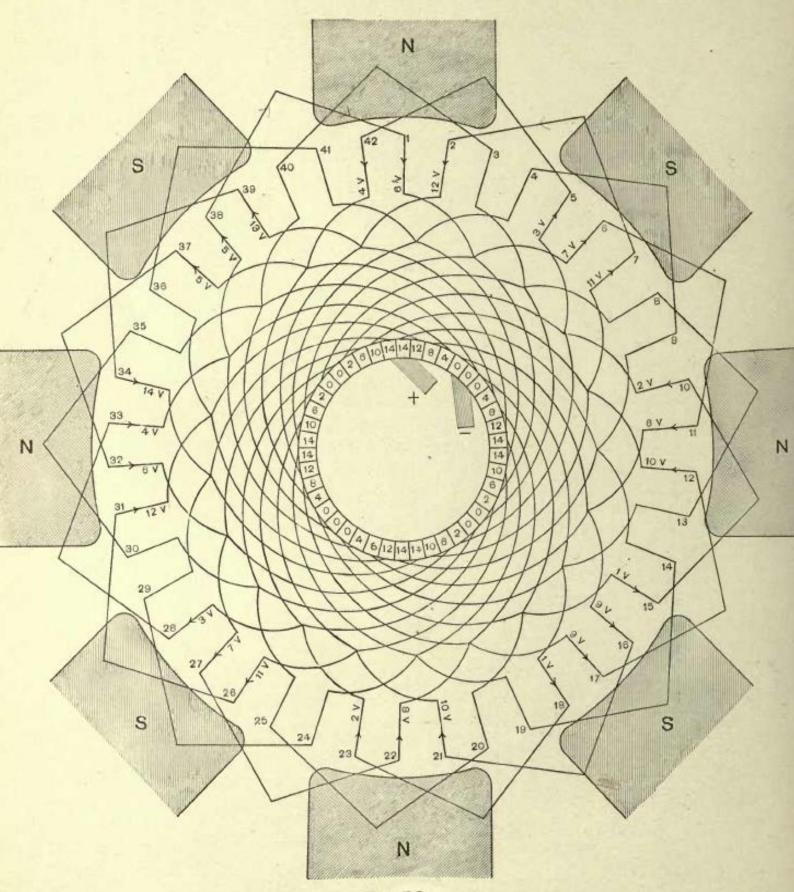


Fig. 52 TWO CIRCUIT, SINGLE WINDING.

In Fig. 52 is given the diagram of a two-circuit, single-wound, eight-pole armature with forty-two conductors.  $C=ny\pm 2$ ;  $8\times 5+2=42$ . It is given to show that, with even numbers of pairs of poles, the number of commutator bars may be doubled by interpolation, and that the result will be to halve the volts between every two segments instead of producing the unsymmetrical result observed in the case of an odd number of pairs of poles.

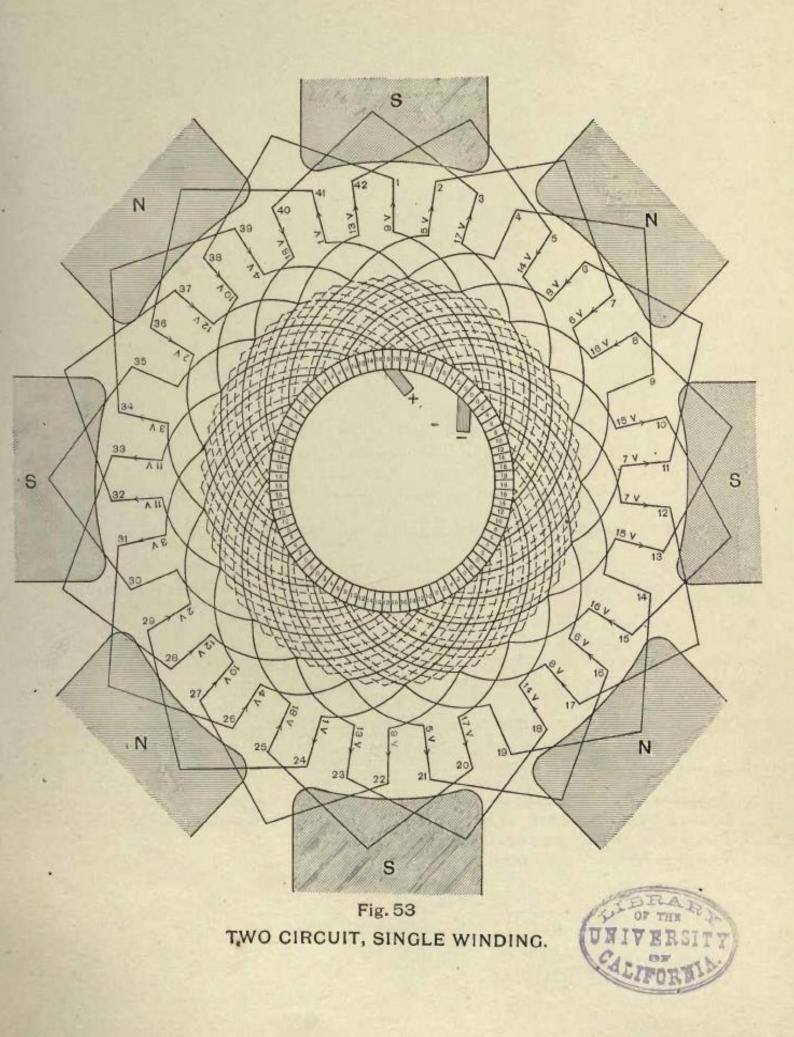
An examination of Fig. 52 will show that commutator segments 180° apart are cross-connected. The scheme of studying the relative potential of conductors and commutator segments is the same as that used in the case of the two preceding figures, and can be followed through without trouble. Some confusion may result from the fact that owing to the small number of conductors taken, the length of the two eircuits through the armature are quite unequal, one path consisting of twelve conductors, and the other of fourteen. As the positive neutral points where these two paths meet must be at the same potential, all the segments at these positions have been indicated as being at a potential of fourteen volts, so that the sequence of figures giving the potentials of the segments is, in four of the eight cases, 0, 4, 8, 12, 14; increasing regularly by four volts until the very end, where the increase is but two volts.

In the other four cases, for the same reason, the sequence is 0, 2, 6, 10, 14, showing the irregularity at the negative neutral points. . With the large number of conductors used in practice no misunderstanding would result.



With an even number of pairs of poles it is not necessary to be confined to using only twice the natural number of commutator segments. Thus in Fig. 53 is given the same eight-pole winding as in Fig. 52, with the exception that eighty-four segments are used instead of forty-two. The natural number of segments would be twenty-one.

As the conventions used in the previous descriptions are followed in mapping out the relative potentials of the various parts, no further explanations will be necessary.



# CHAPTER X.

### TWO-CIRCUIT, MULTIPLE-WOUND, DRUM ARMATURES.

THE next class is that of the two-circuit, multiple-wound, drum armature.

The general formula is: -

 $C = ny \pm 2 m$ ,

where

C = number of face conductors,

n = number of poles,

y = average pitch,

m =number of windings.

The "m" windings will consist of a number of independently re-entrant windings, equal to the greatest common factor of "y" and "m." Therefore, where it is desired that the "m" windings shall combine to form one re-entrant system, it will be necessary that the greatest common factor of "y" and "m" be made equal to 1.

Also, when "y" is an even integer, the pitch must be taken alternately as (y-1) and (y+1).

In Fig. 54 is reproduced a winding described by E. Arnold ("Die Ankerwicklungen der Gleichstrom-Dynamomaschinen," p. 70, Fig. 80), and by Dr. Kittler ("Handbuch der Elektrotechnik," 2d ed., p. 535, Fig. 403, b). It is classified by them as a four-circuit, single winding. They show four narrow brushes, and point out that the winding has the peculiarities that, in connecting up, the pitch is always taken forward, and that the short-circuiting of a coil occurs between opposite brushes of like polarity, instead of entirely at one brush, as is usually the case. They give no further instances of the application of this winding, except that Herr Arnold proposes for it the formula:—

 $C = n(y \pm 1),$ 

and adds that if  $\frac{C}{2}$  and "y" have a common factor, a singly re-entrant winding is not obtained, several independently re-entrant windings being the result. He follows this statement with a diagram having C=28, n=4, and y=6.

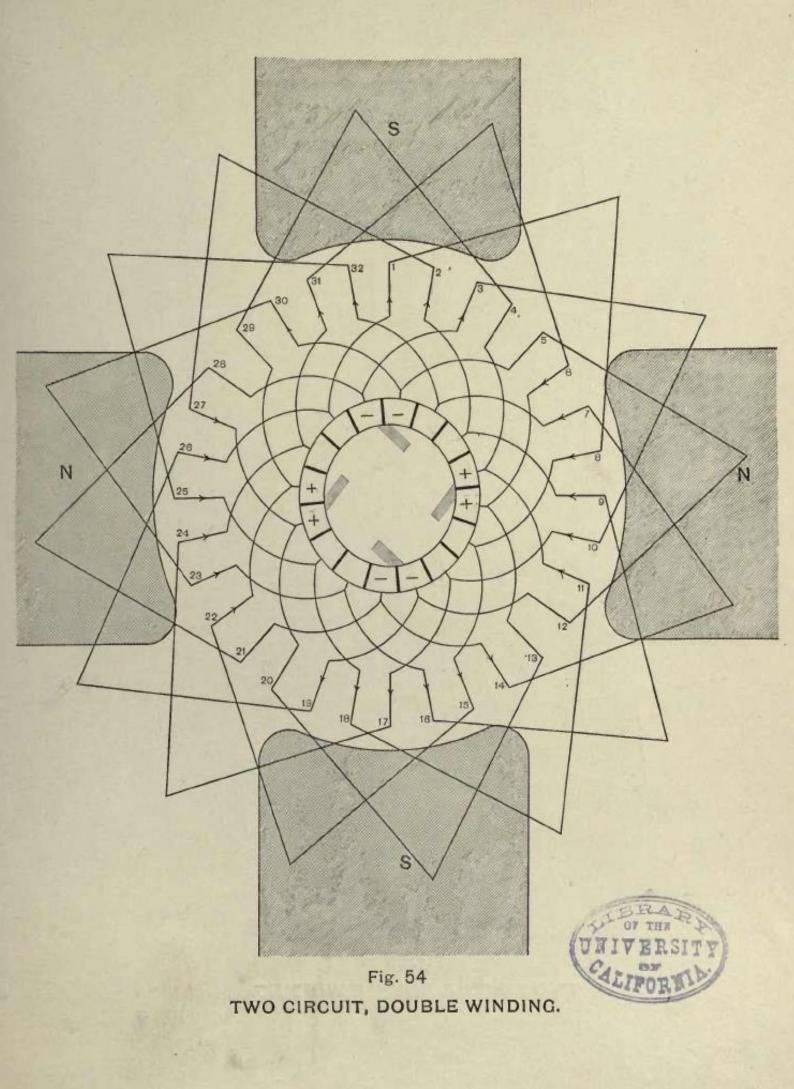
[28=4(6+1)],

which gives two independently re-entrant windings, and shows, as before, four points of commutation.

Returning to a consideration of Fig. 54, it may be seen that at the given position, conductors 5-12 and 21-28 are short circuited at the negative brushes, and 13-20 and 29-4 at the positive.

The circuits through the armature are, -

$$\rightarrow \left\{ 
\begin{array}{c}
 -\left\{ 
\begin{array}{c}
 3-10-17-24-31-6 \\
 30-23-16-9-2-27 \\
 -\left\{ 
\begin{array}{c}
 14-7-32-25-18-11 \\
 19-26-1-8-15-22
\end{array} \right. + 
\right\}$$



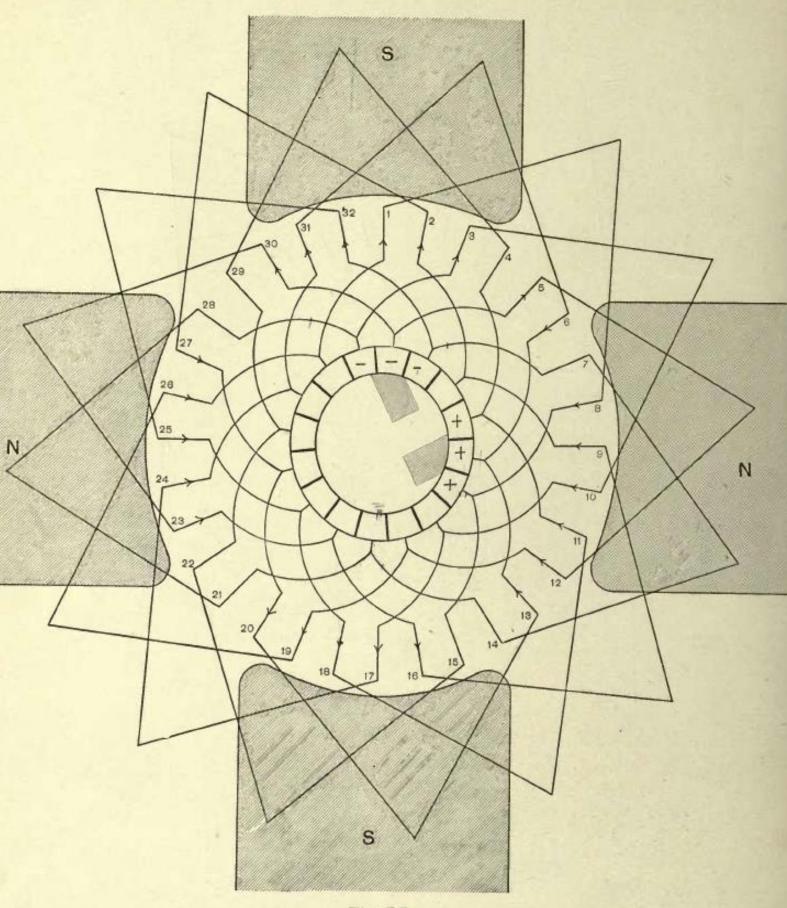


Fig. 55
TWO CIRCUIT, DOUBLE WINDING.

Now in Fig. 55 will be found the very same winding as in Fig. 54, with the exception that two wide brushes are shown instead of four narrow ones. Short-circuiting of a coil now necessarily occurs at one brush, and a study of the winding shows that it is one of the singly re-entrant multiple-wound type, this particular one being a two-circuit, singly re-entrant, double winding.

At the position shown, conductors 7-14-21-28 are short-circuited at the negative brush, and 15-22-29-4 at the positive. The circuits through the armature are:—

It will be seen that, owing to the very small number of conductors, the winding is extremely irregular, but it will not be difficult to perceive that the nature of the course taken by the current through the armature remains essentially unaltered from that of Fig. 54, consisting, as there, of four paths with an average of six conductors in series per path. The current, however, enters the armature from one wide brush, which always spans more than one segment, and departs from a similar wide brush  $\left(\frac{360}{n}\right)^{\circ}$  removed. But in the former case (Fig. 54), it entered two of the paths by one narrow negative brush, and the other two by another, situated  $\left[\frac{360}{n}\right]^{\circ}$  distant.

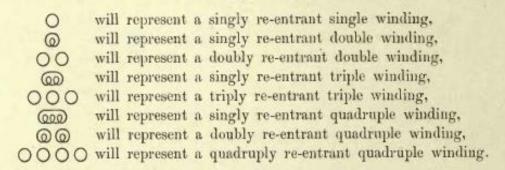
It appears, therefore, conclusive that Fig. 54 is in all essential respects identical with a two-circuit, singly re-entrant, double winding, but this was probably not perceived by the above-mentioned authors: otherwise they would undoubtedly have extended the principle to higher orders of multiples and other numbers of poles. An eight-pole, two-circuit, singly re-entrant, triple winding (which would, of course, follow six paths through the conductors of the armature) would probably not have been considered possible, their conception of the winding apparently being that it was a multiple winding with as many paths through the conductors of the armature as the machine had poles. The formula and rules enunciated in this investigation follow naturally from the true conception of this winding, whereas the formula and condition stated by Herr Arnold may be seen, by a few attempts to apply it, to be entirely inadequate for the purpose of obtaining the necessary data for constructing such windings.



The two preceding figures (54 and 55) were given for the purpose of showing in how far the two-circuit, multiple windings have been understood in the past. The numbers of conductors were, however, entirely inadequate to fully illustrate the nature of the windings.

As this class promises to have a somewhat wide application, it is proposed to give a good many examples, selecting for the purpose various values of "C," "n," "y" and "m," and briefly analyzing each case on the basis of the rules given on page 114.

The symbolical representations heretofore used will be continued, thus: -



According to the above nomenelature, Fig. 40 would be a six-circuit, singly re-entrant, double winding [@]; Fig. 37 would be a six-circuit, singly re-entrant, triple winding [@]; and Fig. 38 a four-circuit, doubly re-entrant, quadruple winding [@]. The use of the middle expression, "singly, doubly, etc., re-entrant," is unavoidable for absolute definiteness, but it will in most cases be sufficiently definite to speak, for example, of a "six-circuit, triple winding" and a "two-circuit, quadruple winding," where absolute exactness would require them to be spoken of respectively as a "six-circuit, singly re-entrant, triple winding" and a "two-circuit, doubly re-entrant, quadruple winding."

Figure 56 is a four-pole, two-circuit, singly re-entrant, triple winding. It is represented symbolically thus:  $\bigcirc$  n=4, and m=3. In order that it should be singly re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 1. Therefore "y" was taken equal to 16.

$$C = ny \pm 2 m = 4 \times 16 \pm 2 \times 3 = 58 \text{ or } 70.$$

Seventy conductors have been taken, and "y" is alternately 15 and 17, it being, of course, impossible to use 16.

In the position shown, the conductors without arrowheads are short-circuited, and the circuits through the armsture are:—

$$\begin{array}{c} \longrightarrow \\ - \\ \begin{pmatrix} 67-50-35-18-& 3-56-41-24 \\ 65-48-33-16-& 1-54-39-22 \\ 63-46-31-14-69-52-37-20-& 5-58-43-26 \\ 10-27-42-59-& 4-21-36-53-68-15 \\ 8-25-40-57-& 2-19-34-51-66-13 \\ 6-23-38-55-70-17-32-49-64-11 \\ \end{array} \right\} + \\ \begin{array}{c} \longrightarrow \\ - \\ \longrightarrow \\ \end{array}$$

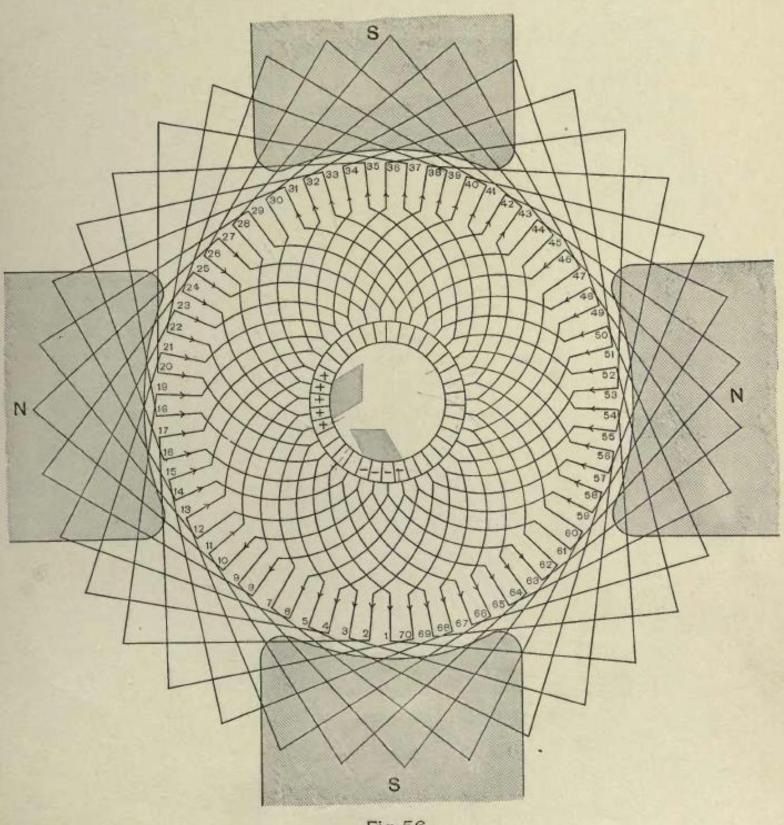
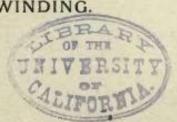
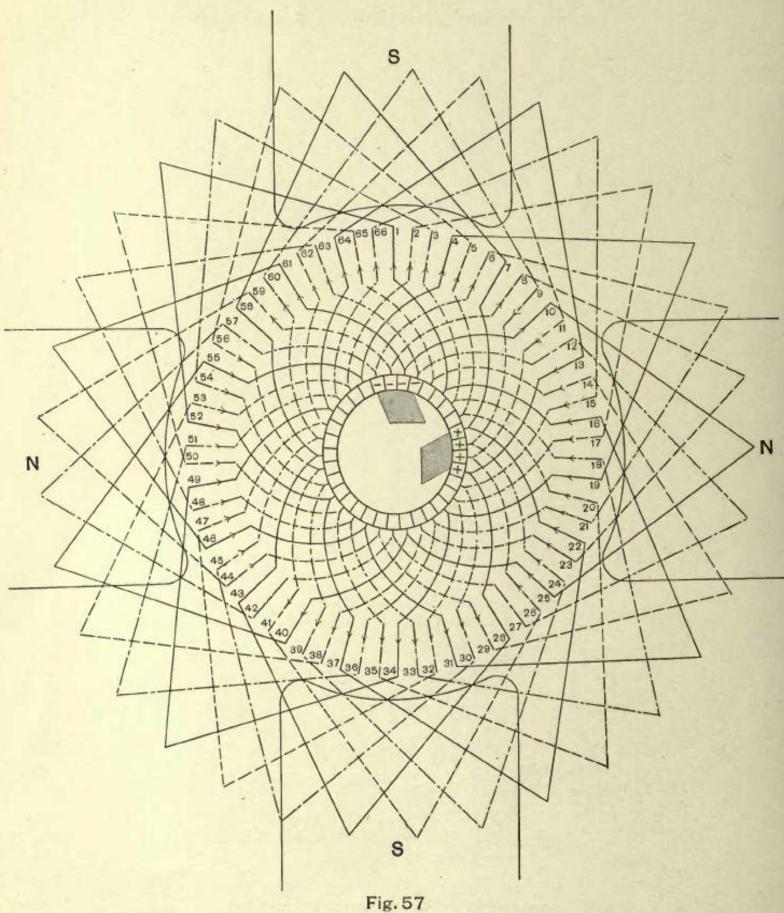


Fig. 56

TWO CIRCUIT, TRIPLE WINDING.





TWO CIRCUIT, TRIPLE WINDING.

Figure 57 is a four-pole, two-circuit, triply re-entrant, triple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc$  n=4, and m=3. In order that it should be triply re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 3. Therefore "y" was taken equal to 15.

$$C = ny \pm 2 m = 4 \times 15 \pm 2 \times 3 = 54$$
 or 66.

Sixty-six conductors have been taken. The three independently re-entrant windings have been shown by three different styles of lines.

In the position shown, the conductors without arrowheads are short-circuited, and the circuits through the armature are: —

$$\rightarrow \begin{bmatrix} 63-48-33-18-&3-54-39-24-&&&\\ 61-46-31-16-&1-52-37-22-&&&\\ 59-44-29-14-65-50-35-20-&5-56-41-26\\ 10-25-40-55-&4-19-34-49-64-13-&&\\ 8-23-38-53-&2-17-32-47-62-11-&&\\ 6-21-36-51-66-15-30-45-60-&9-&& \end{bmatrix} + \rightarrow$$

It is interesting to compare this winding and table with the preceding, and to notice how very slightly they differ.



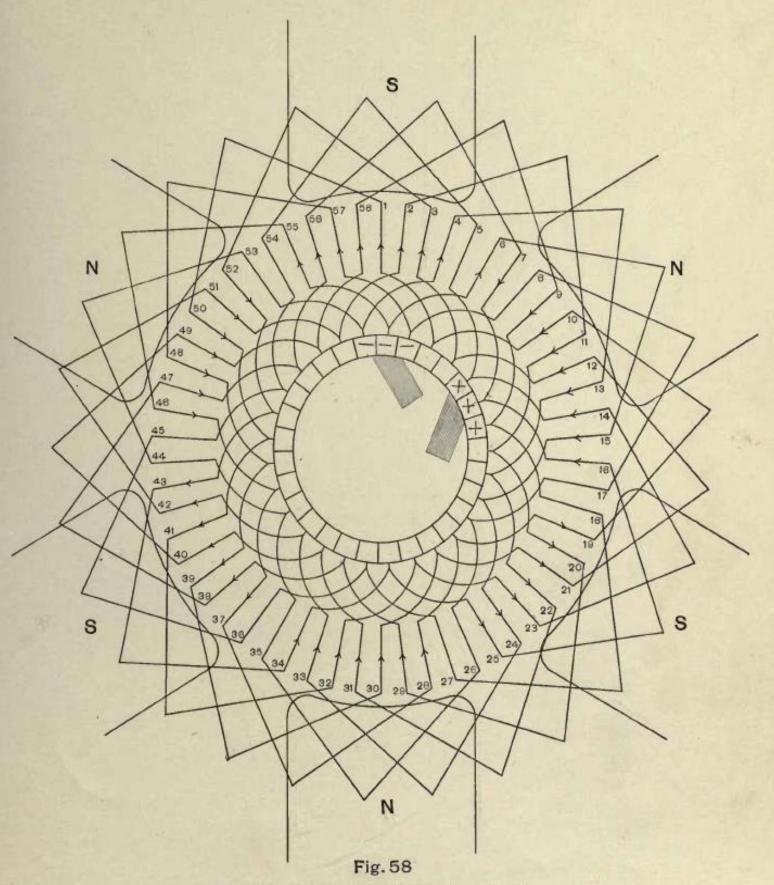
Figure 58 is a six-pole, two-circuit, singly re-entrant, double winding. It would be represented symbolically as  $\odot$ . n=6, and m=2.

In order that it should be singly re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 1. Therefore "y" was taken equal to 9.

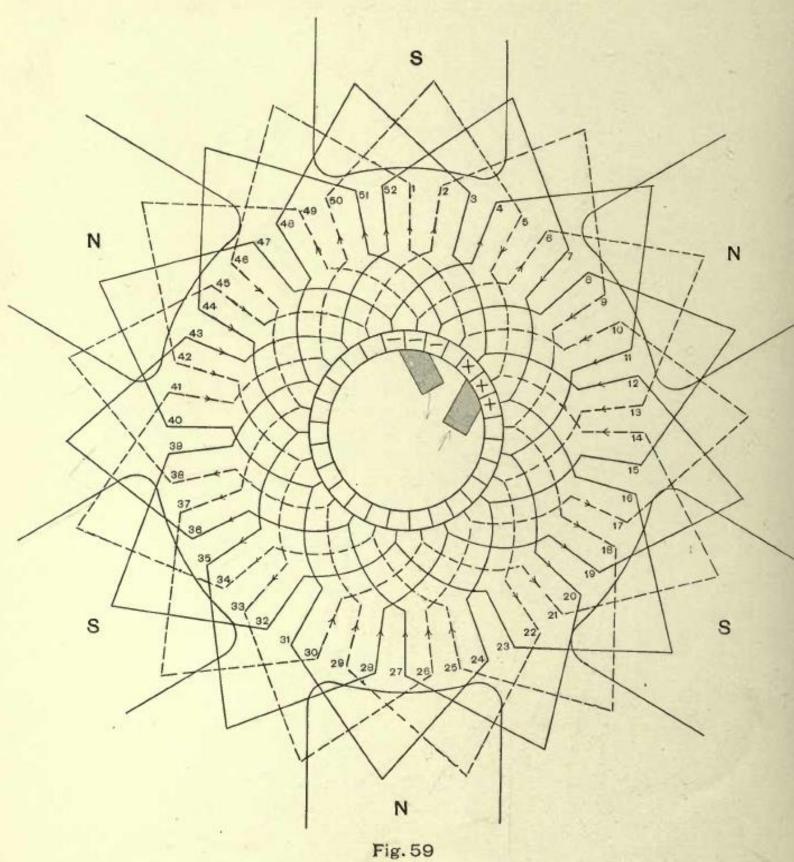
$$C = ny \pm 2 m = 6 \times 9 \pm 2 \times 2 = 50 \text{ or } 58.$$

Fifty-eight conductors have been taken.

In the position shown, the circuits through the armature are:—



TWO CIRCUIT, DOUBLE WINDING.



TWO CIRCUIT, DOUBLE WINDING.

Figure 59 is a six-pole, two-circuit, doubly re-entrant, double winding, the symbolical representation being  $\bigcirc\bigcirc$   $\bigcirc$  n=6, and m=2. In order that it should be doubly re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 2. Therefore "y" was taken equal to 8.

$$C = ny \pm 2 m = 6 \times 8 \pm 2 \times 2 = 44$$
 or 52.

Fifty-two conductors have been taken, and "y" is alternately 7 and 9, it being, of course, impossible to use y=8.

In the position shown, the conductors without arrowheads are short-circuited, and the circuits through the armature are:—

$$\longrightarrow - \begin{bmatrix} 51 - 44 - 35 - 28 - 19 - 12 - 49 - 42 - 33 - 26 - 17 - 10 - 1 - 46 - 37 - 30 - 21 - 14 - 6 - 13 - 22 - 29 - 38 - 45 - 2 - 9 - 18 - 25 - 34 - 41 - 50 - 5 \\ 4 - 11 - 20 - 27 - 36 - 43 - 52 - 7 - 5 - 34 - 41 - 50 - 5 \end{bmatrix} + \cdots$$

As frequently remarked in connection with other diagrams having small numbers of conductors, the very unequal lengths of the different paths through the armature is entirely caused by this choice of a small number of conductors, and would, to a large extent, disappear with all practicable numbers of conductors.

The two independently re-entrant windings are drawn respectively with full and with dotted lines.



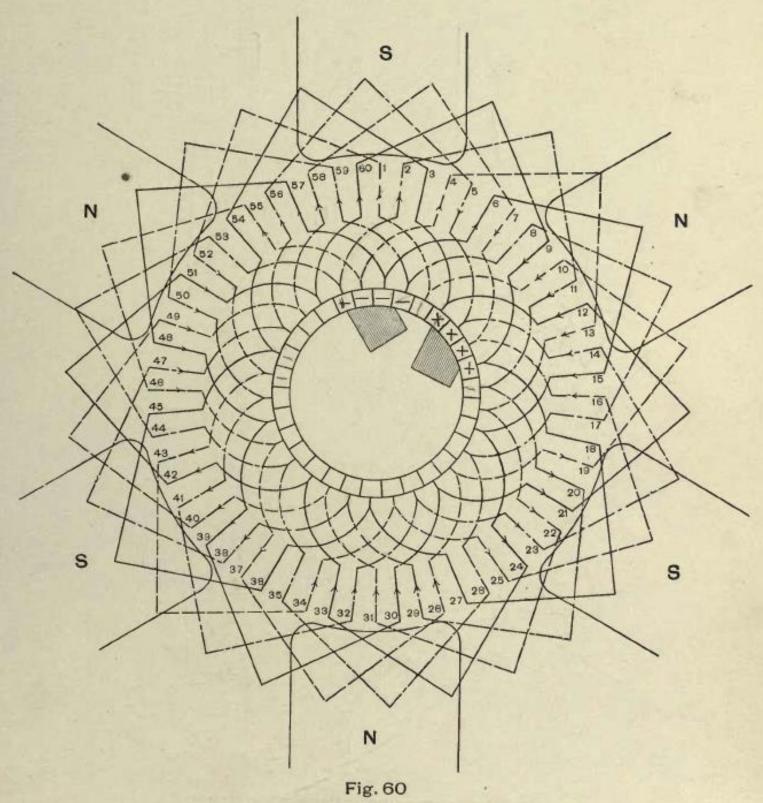
Figure 60 is a six-pole, two-circuit, triply re-entrant, triple winding. It would be represented symbolically as  $\bigcirc\bigcirc$   $\bigcirc$  n=6, and m=3. In order that it should be triply re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 3. Therefore "y" was taken equal to 9.

$$C = ny \pm 2m = 6 \times 9 \pm 2 \times 3 = 48$$
 or 60.

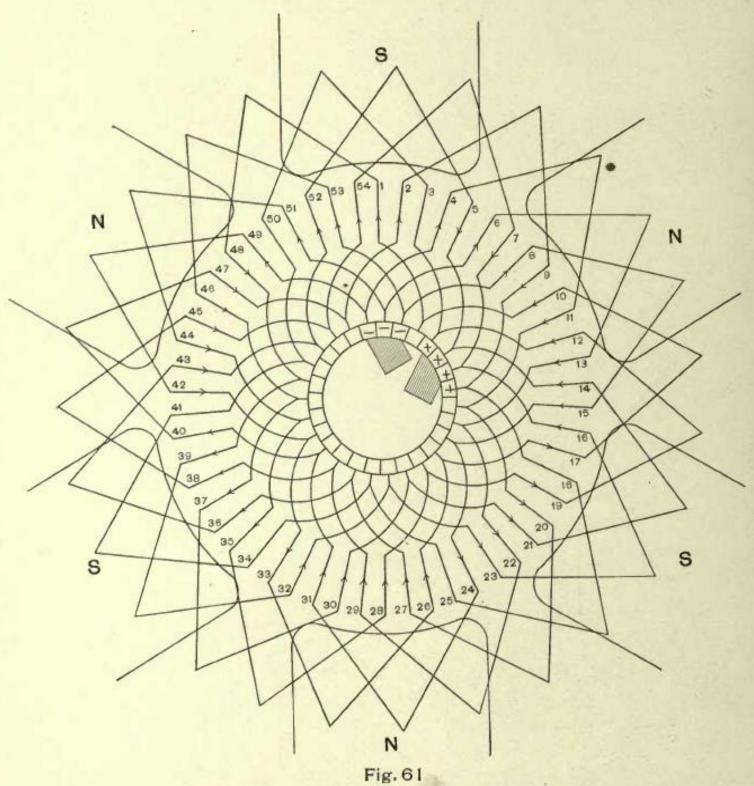
Sixty conductors have been taken.

The three independently re-entrant windings have been represented by three different styles of lines.

$$\rightarrow \begin{array}{c} \begin{bmatrix} 59-50-41-32-23-14 \\ 57-48-39-30-21-12 \\ 55-46-37-28-19-10-1-52-43-34-25-16 \\ 6-15-24-33-42-51-60-9 \\ 4-13-22-31-40-49-58-7 \\ 2-11-20-29-38-47-56-5 \end{bmatrix} + \rightarrow \\ \\ \end{bmatrix}$$



TWO CIRCUIT, TRIPLE WINDING.



TWO CIRCUIT, TRIPLE WINDING.

Figure 61 is a six-pole, two-circuit, singly re-entrant, triple winding. It may be symbolically expressed as  $\bigcirc$  n=6, and m=3. In order that it should be singly re-entrant, it was necessary for the greatest common factor of "m" and "y" to be 1. Therefore "y" was taken equal to 8.

$$C = ny \pm 2m = 6 \times 8 \pm 2 \times 3 = 42$$
 or 54.

Fifty-four conductors have been taken, "y" is alternately 7 and 9, as it would, of course, be impossible to let y=8.

$$\rightarrow \begin{bmatrix} 53-46-37-30-21-14 \\ 51-44-35-28-19-12 \\ 49-42-33-26-17-10-1-48-39-32-23-16 \\ 8-15-24-31-40-47-2-9 \\ 6-13-22-29-38-45-54-7 \\ 4-11-20-27-36-43-52-5 \end{bmatrix} + \rightarrow$$

Figure 62 is a six-pole, two circuit, triply re-entrant, triple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc$  n=6, m=3. In order that it should be triply re-entrant, it was necessary for the greatest common factor of "y" and "m" to be 3. Therefore "y" was taken equal to 12.

$$C = ny \pm 2$$
  $m = 6 \times 12 \pm 2 \times 3 = 66$  or 78.

Seventy-eight conductors have been taken, and "y" is alternately 11 and 13, as it would not be possible to let "y"=12.

The three independently re entrant windings have been represented by three different styles of lines.

In the position shown, the short-circuited conductors are those without arrow-heads. The circuits through the armature are:—

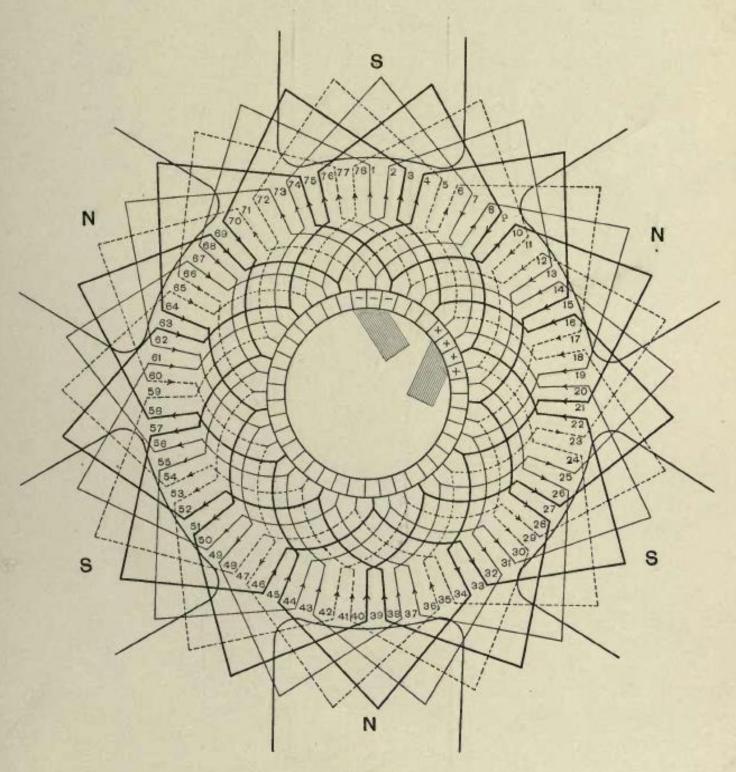


Fig. 62 TWO CIRCUIT, TRIPLE WINDING.

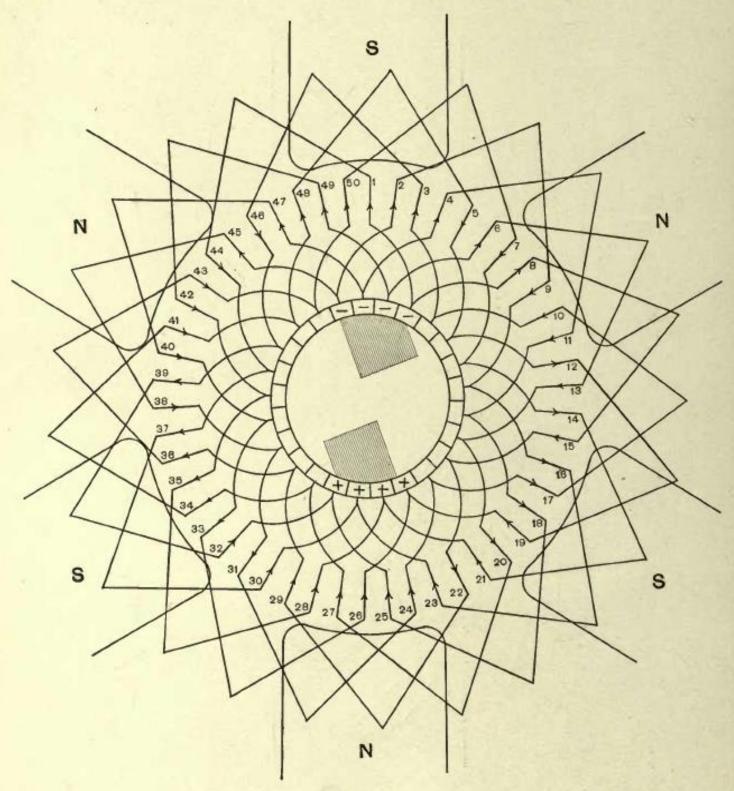


Fig. 63
TWO CIRCUIT, QUADRUPLE WINDING

$$C = ny \pm 2$$
  $m = 6 \times 7 \pm 2 \times 4 = 34$  or 50.

Fifty conductors have been taken.

$$\rightarrow \begin{array}{c} 1 - 44 - 37 - 30 - \\ 49 - 42 - 35 - 28 - \\ 47 - 40 - 33 - 26 - \\ 45 - 38 - 31 - 24 - 17 - 10 - 3 - 46 - 39 - 32 \\ 8 - 15 - 22 - 29 - 36 - 43 - 50 - 7 - 14 - 21 \\ 6 - 13 - 20 - 27 - 34 - 41 - 48 - 5 - 12 - 19 \\ 4 - 11 - 18 - 25 - \\ 2 - 9 - 16 - 23 - \\ \end{array} \right\} + \rightarrow$$

Figure 64 is a six-pole, two-circuit, quadruply re-entrant, quadruple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc\bigcirc$  n=6, and m=4. In order that it should be quadruply re-entrant, it was necessary for the greatest common factor of "y" and "m" to be 4. Therefore "y" was taken equal to 8.

$$C = ny \pm 2 m = 6 \times 8 \pm 2 \times 4 = 40 \text{ or } 56.$$

Fifty-six conductors have been taken. "y" is alternately 7 and 9, as it is obviously impossible to let y=8.

$$\rightarrow \begin{array}{c} 55 - 48 - 39 - 32 \\ 53 - 46 - 37 - 30 - \\ 51 - 44 - 35 - 28 - 19 - 12 - 3 - 52 - 43 - 36 \\ 49 - 42 - 33 - 26 - 17 - 10 - 1 - 50 - 41 - 34 \\ 8 - 15 - 24 - 31 - 40 - 47 - 56 - 7 - 16 - 23 \\ 6 - 13 - 22 - 29 - 38 - 45 - 54 - 5 - 14 - 21 \\ 4 - 11 - 20 - 27 - \\ 2 - 9 - 18 - 25 \end{array} \right\} + \rightarrow$$

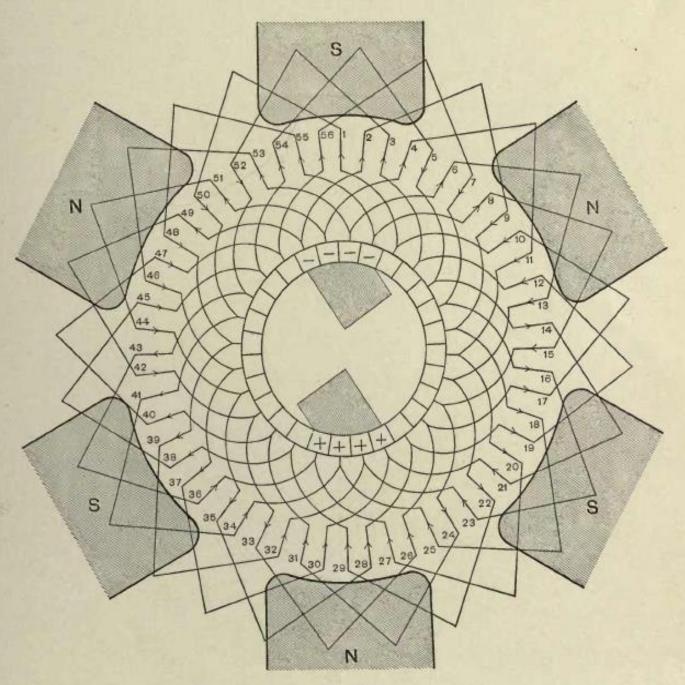


Fig. 64.
TWO CIRCUIT QUADRUPLE WINDING.



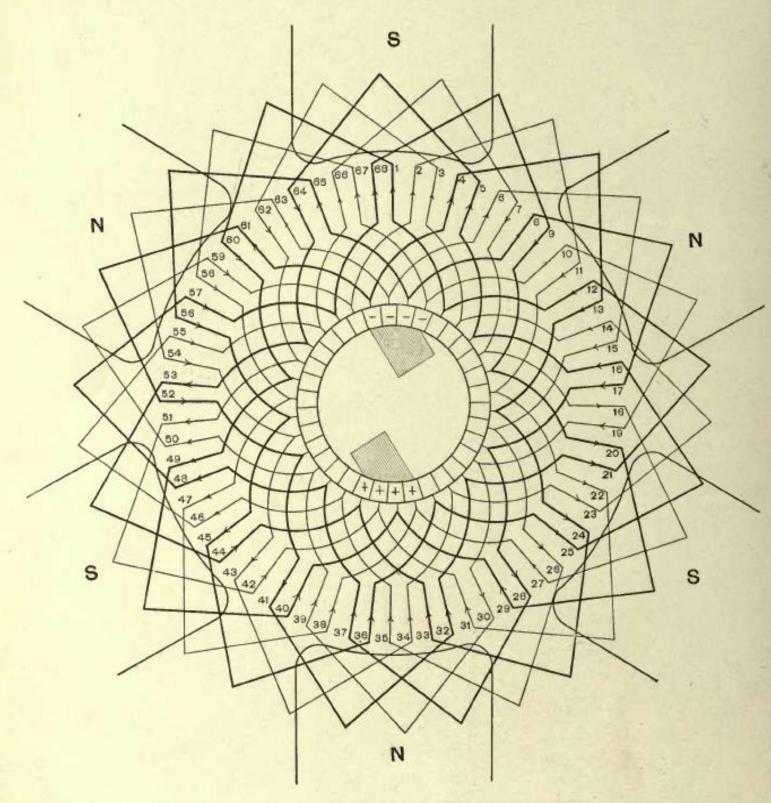


Fig. 65 TWO CIRCUIT, QUADRUPLE WINDING.

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Figure 65 is a six-pole, two-circuit, doubly re-entrant, quadruple winding. It would be represented symbolically as  $\bigcirc \bigcirc n=6$ , and m=4. In order that it should be doubly re-entrant, it was necessary for the greatest common factor of "y" and "m" to be 2. Therefore "y" was taken equal to 10.

$$C = ny \pm 2 m = 6 \times 10 \pm 2 \times 4 = 52 \text{ or } 68.$$

Sixty-eight conductors have been chosen. "y" is alternately 9 and 11, because its average value, being even, could not be used.

The two independently re-entrant windings have been represented respectively by light and by heavy lines.

$$\rightarrow \begin{bmatrix} 67-58-47-38 & & & \\ 63-54-43-34-23-14-& 3-62-51-42 \\ 65-56-45-36-25-16-& 5-64-53-44 \\ 61-52-41-32-21-12-& 1-60-49-40 \\ 10-19-30-39-50-59-& 2-11-22-31 \\ 6-15-26-35-46-55-66-& 7-18-27 \\ 8-17-28-37-48-57-68-& 9-20-29 \\ 4-13-24-33- & & \end{bmatrix} + \rightarrow$$

Figure 66 is a six-pole, two-circuit, quadruply re-entrant, quadruple winding  $[\bigcirc\bigcirc\bigcirc\bigcirc]$ . n=6, and m=4. In order that it should be quadruply re-entrant, it was necessary that the greatest common factor of "y" and "m" should be 4. Therefore "y" was taken equal to 12.

$$C = ny \pm 2 m = 6 \times 12 \pm 2 \times 4 = 64 \text{ or } 80.$$

Eighty conductors have been taken. "y" is alternately 11 and 13, its average value being even.

The four independently re-entrant windings have been represented by four varieties of lines.

$$\begin{array}{c} 77-66-53-42-29-18-5-74-61-50 \\ 75-64-51-40-27-16-3-72-59-48 \\ 73-62-49-38-25-14-1-70-57-46 \\ 71-60-47-36-23-12-79-68-55-44 \\ 10-21-34-45-58-69-2-13-26-37 \\ 8-19-32-43-56-67-80-11-24-35 \\ 6-17-30-41-54-65-78-9-22-33 \\ 4-15-28-39-52-63-76-7-20-31 \end{array} \right) + \longrightarrow \\ \end{array}$$

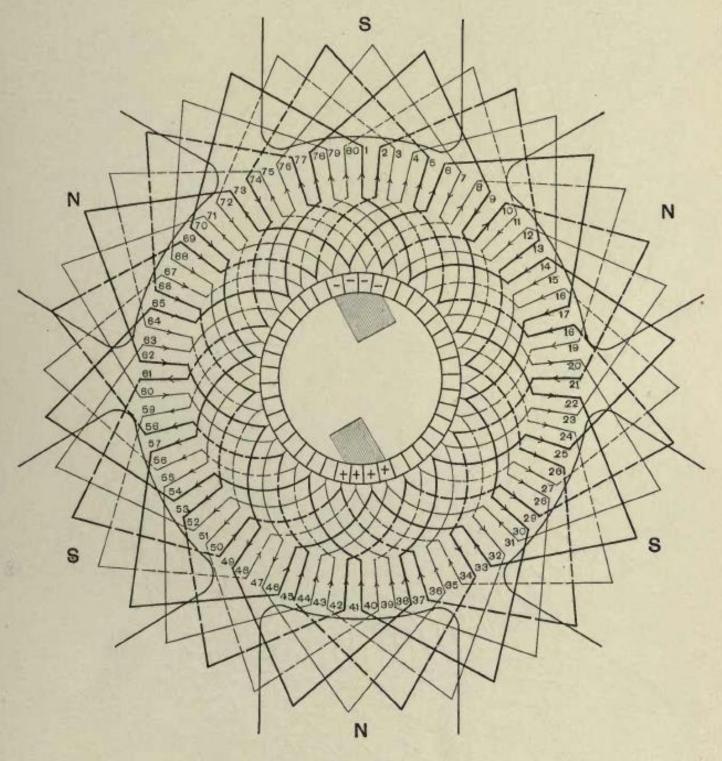
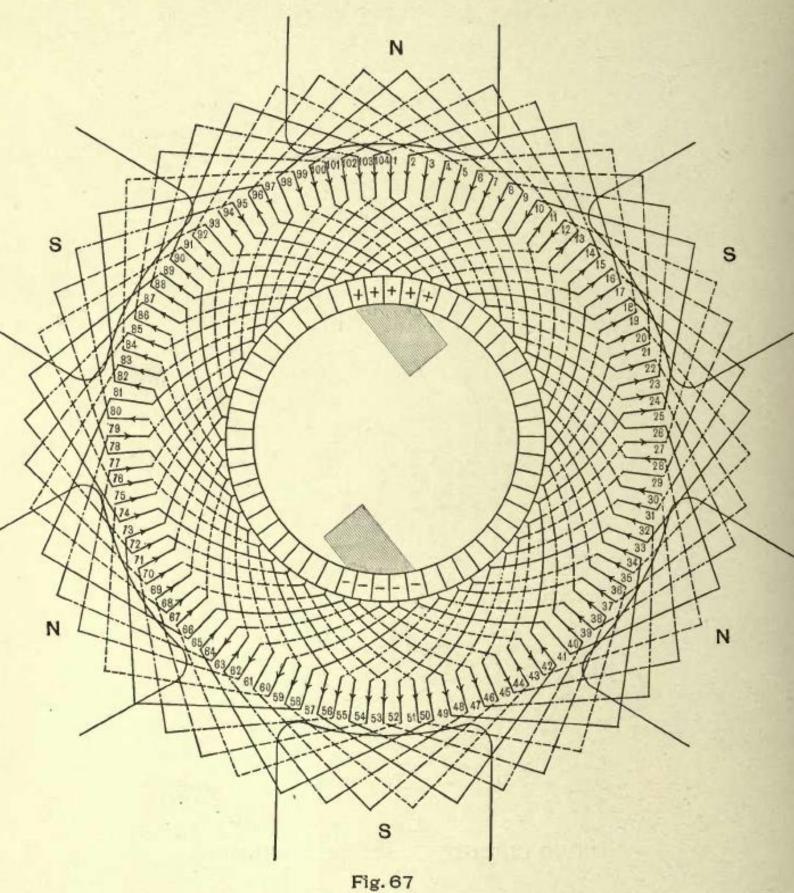


Fig. 66
TWO CIRCUIT, QUADRUPLE WINDING.





TWO CIRCUIT, QUADRUPLE WINDING.

Figure 67 is a six-pole, two-circuit, quadruply re-entrant, quadruple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc\bigcirc$  0. n=6, and m=4. In order that it should be quadruply re-entrant, it was necessary that the greatest common factor of "y" and "m" should be 4. Therefore "y" was taken equal to 16.

$$C = ny \pm 2 m = 6 \times 16 \pm 2 \times 4 = 88 \text{ or } 104.$$

One hundred and four conductors have been taken. "y" is 17 at the front end, and 15 at the back end, thus averaging 16.

The four independently re-entrant windings have been represented by four different styles of lines.

```
 \begin{array}{c} 49-34-17- \ 2-89-74-57-42-25-10 \\ 47-32-15-104-87-72-55-40-23-8 \\ 45-30-13-102-85-70-53-38-21-6 \\ 43-28-11-100-83-68-51-36-19-4-91-76-59-44-27-12 \\ 64-79-96-7-24-39-56-71-88-103-16-31-48-63-80-95 \\ 62-77-94-5-22-37-54-69-86-101 \\ 60-75-92-3-20-35-52-67-84-99 \\ 58-73-90-1-18-33-50-65-82-97 \\ \end{array} \right\} + \rightarrow
```



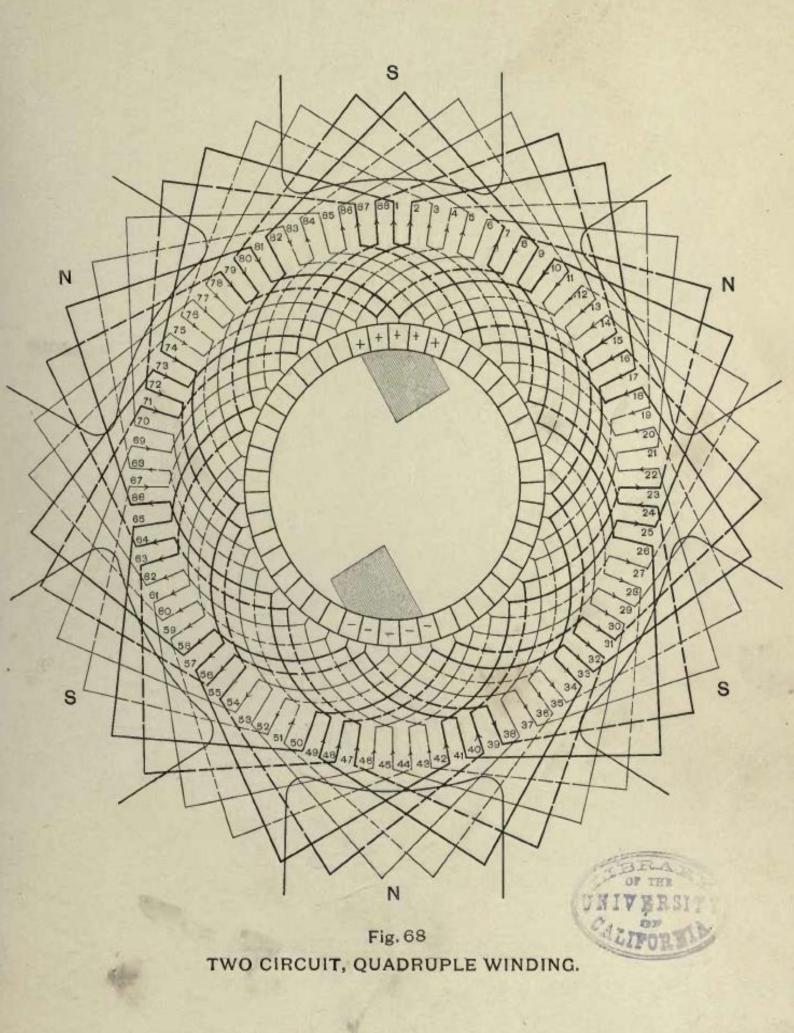
Figure 68 differs from Fig. 67 in the use of the negative instead of the positive sign in the formula. It is given to emphasize the fact that this has no influence on the type of winding. It requires, however, a greater length of copper for a given number of conductors. Like Fig. 67, it is a six-pole, two-circuit, quadruply re-entrant, quadruple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc$   $\bigcirc$  n=6, and m=4. In order that it should be quadruply re-entrant, it was necessary for the greatest common factor of "y" and "m" to be 4. Therefore "y" was taken equal to 16.

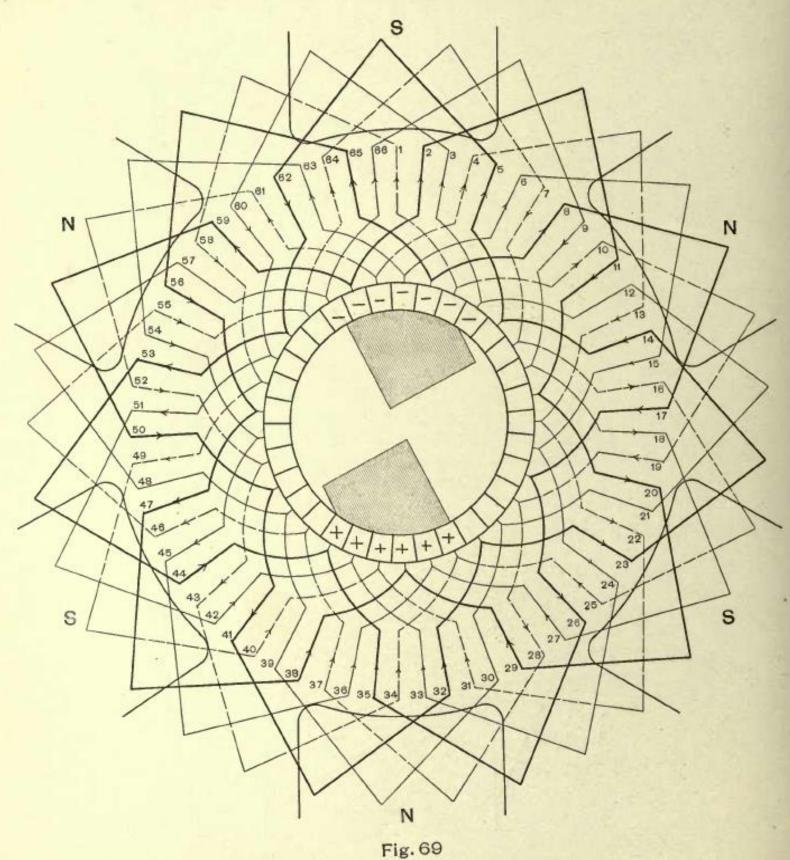
$$C=ny \pm 2 m = 6 \times 16 \pm 2 \times 4 = 88 \text{ or } 104.$$

Eighty-eight conductors have been taken. "y" is 17 at the front, and 15 at the back end.

The four independently re-entrant windings have been represented by different kinds of lines.

$$\rightarrow \begin{array}{c} 58-73-2-17-34-49-66-81- \\ 56-71-88-15-32-47-64-79- \\ 54-69-86-13-30-45-62-77- \\ 52-67-84-11-28-43-60-75-4-19-36-51-68-83 \\ 33-18-1-74-57-42-25-10- \\ 35-20-3-76-59-44-27-12- \\ 37-22-5-78-61-46-29-14- \\ 39-24-7-80-63-48-31-16-87-72-55-40-23-8 \end{array} \right\} + \begin{array}{c} \\ \\ \\ \end{array}$$





TWO CIRCUIT, SEXTUPLE WINDING.

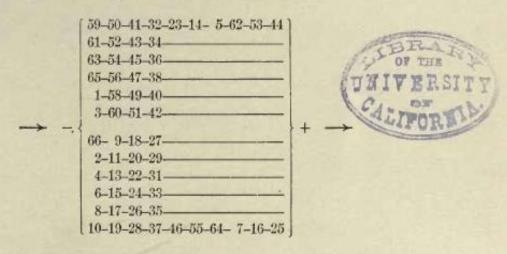
The next four diagrams (Figs. 69, 70, 71, 72) form a group of sextuple windings. It is thought that an examination of this group will bring out very clearly the method of applying and the interpretation of the rules concerning two-circuit, multiple windings. The following table will be of assistance in studying them:—

Figure,	*	у	m	C	G.C.F. of m and y.	Name of Winding.	Symbol,
69	6	9	6	66	3	Two-circuit, triply re-entrant, sextuple winding.	000
70	6	10	6	72	2	Two-circuit, doubly re-entrant, sextuple winding.	@@
71	6	11	6	78	1	Two-circuit, singly re-entrant, sextuple winding.	(00000)
72	6	12	6	84	6	Two-circuit, sextuply re-entrant, sextuple winding.	000000

Figure 69 is a six-pole, two-circuit, triply re-entrant, sextuple winding. It would be symbolically represented as 000. n=6, and m=6. In order that it should be triply re-entrant, it was necessary that the greatest common factor of "m" and "y" should be 3. Therefore "y" was taken equal to 9.

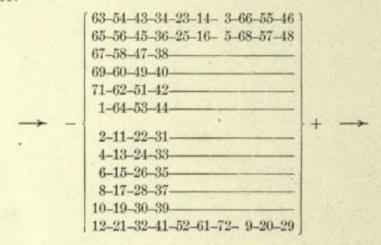
$$C = ny \pm 2 m = 6 \times 9 \pm 2 \times 6 = 42 \text{ or } 66.$$

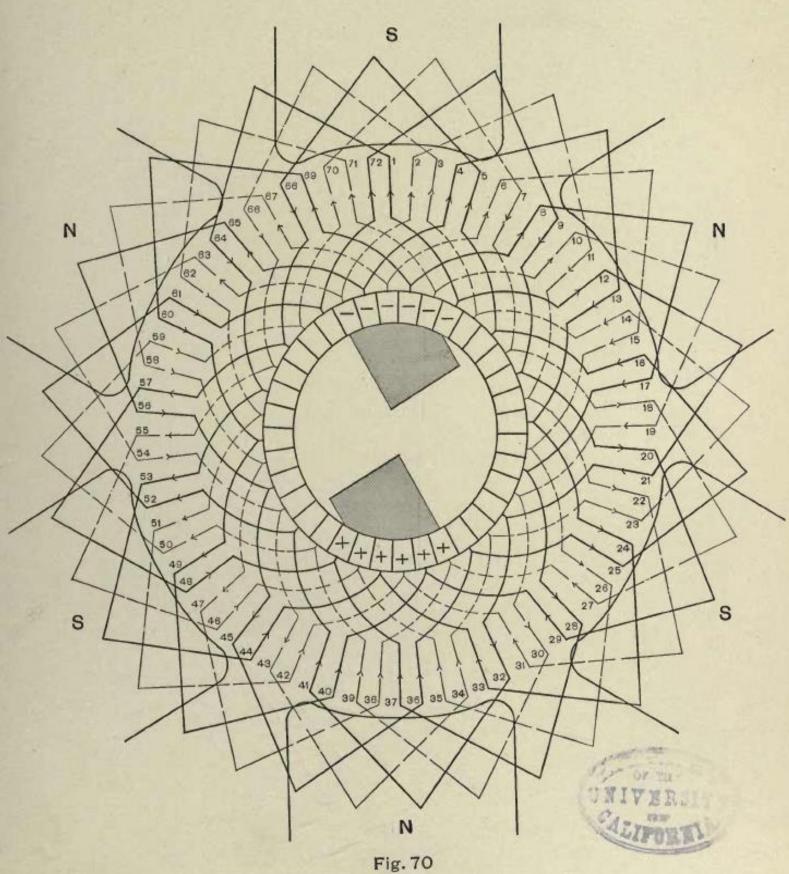
Sixty-six conductors were taken. The three independently re-entrant windings have been represented respectively by light, heavy, and broken lines.



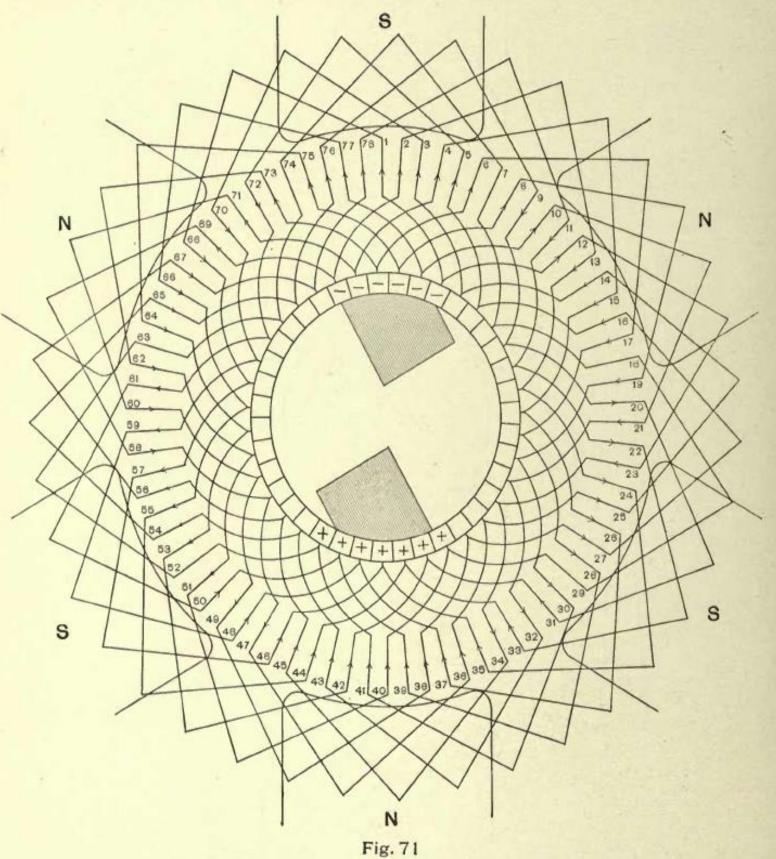
$$C = ny \pm 2 m = 6 \times 10 \pm 2 \times 6 = 48 \text{ or } 72.$$

Seventy-two conductors have been taken. The two independently re-entrant windings have been represented respectively by full and dotted lines.





TWO CIRCUIT, SEXTUPLE WINDING.



TWO CIRCUIT, SEXTUPLE WINDING.

Figure 71 is a six-pole, two-circuit, singly re-entrant, sextuple winding. It would be represented symbolically as (00000). n=6, and m=6. In order that it should be singly re-entrant, it was necessary that the greatest common factor of "m" and "y" should be 1. Therefore "y" was taken equal to 11.

$$C = ny \pm 2 m = 6 \times 11 \pm 2 \times 6 = 54 \text{ or } 78.$$

Seventy-eight conductors have been chosen.

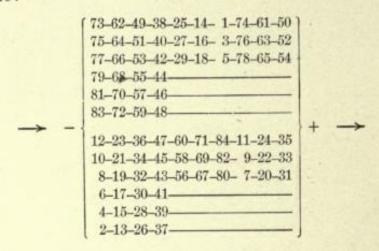
In the given position, the circuits through the armature are: —

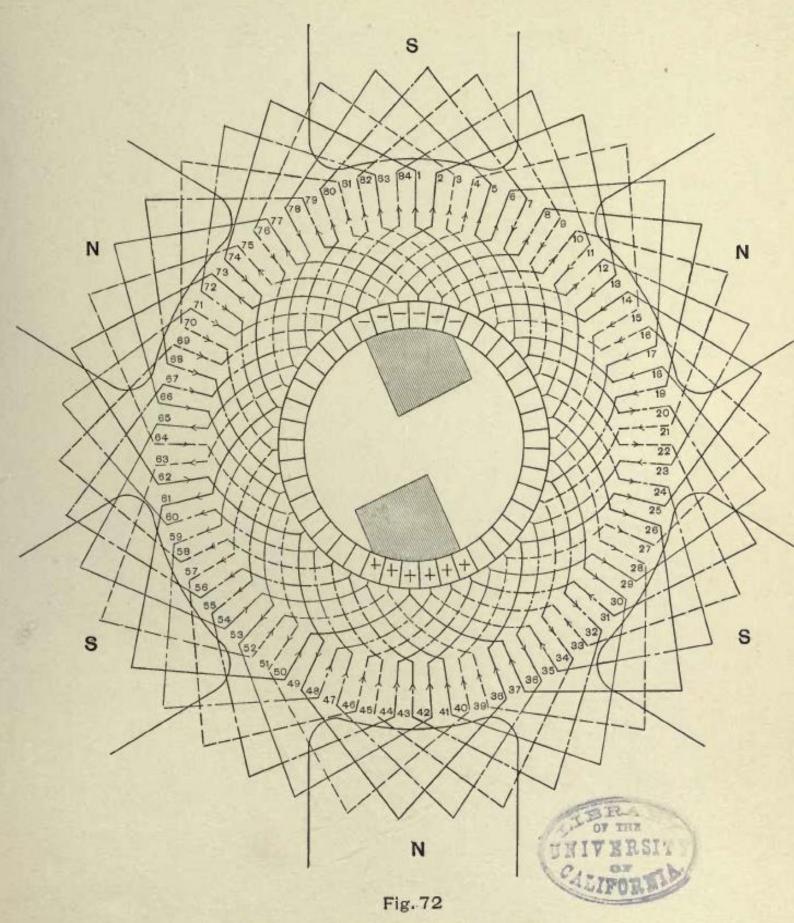
Figure 72 is a six-pole, two-circuit, sextuply re-entrant, sextuple winding. It would be represented symbolically as  $\bigcirc\bigcirc\bigcirc\bigcirc\bigcirc\bigcirc\bigcirc$ . n=6, and m=6. In order that it should be sextuply re-entrant, it was necessary that the greatest common factor of "m" and "y" should be 6. Therefore "y" was taken equal to 12.

$$C = ny \pm 2 m = 6 \times 12 \pm 2 \times 6 = 60 \text{ or } 84.$$

Eighty-four conductors have been taken.

The six independently re-entrant windings are represented respectively by different styles of lines. "y," of course, is taken alternately 11 and 13.





TWO CIRCUIT, SEXTUPLE WINDING.

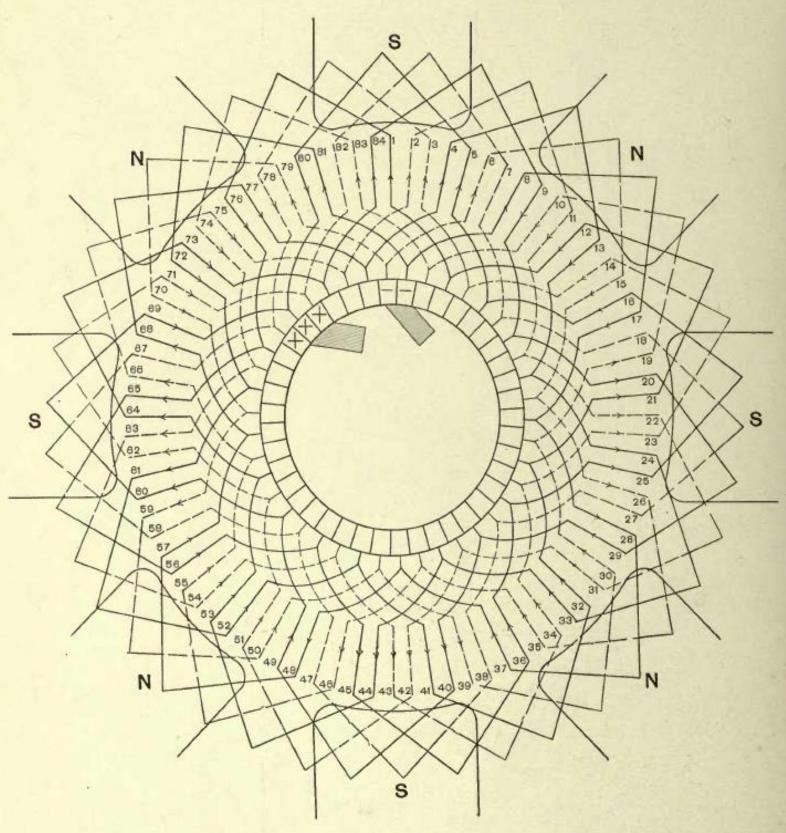


Fig. 73
TWO CIRCUIT, DOUBLE WINDING.

Figure 73 is an eight-pole, two-circuit, doubly re-entrant, double winding. It would be represented symbolically as  $\bigcirc \bigcirc$ . n=8, and m=2. In order that it should be doubly re-entrant, it was necessary that the greatest common factor of "m" and "y" should be 2. Therefore "y" was taken equal to 10.

$$C = ny \pm 2 m = 8 \times 10 \pm 2 \times 2 = 76 \text{ or } 84.$$

Eighty-four conductors have been taken.

The two independently re-entrant windings are represented respectively by full and dotted lines. "y" is taken alternately 11 and 9, the average pitch being 10.

In the given position, the circuits through the armature are: -

Figure 74 is an eight-pole, two-circuit, singly re-entrant, double winding. It would be represented symbolically as n. n=3, and m=2. In order that it should be singly re-entrant, it was necessary that the greatest common factor of "y" and "m" should be 1. Therefore "y" was taken equal to 11.

$$C = ny \pm 2 m = 8 \times 11 \pm 2 \times 2 = 84 \text{ or } 92.$$

Eighty-four conductors have been taken just as in the preceding figure. In the given position, the circuits through the armature are:—

$$\rightarrow \left. - \left\{ \begin{array}{c} 8-19-30-41-52-63-74-1-12-23-34-45-56-67\\ 6-17-28-39-50-61-72-83-10-21-32-43-54-65-76-3-14-25-36-47-58-69\\ 81-70-59-48-37-26-15-4-77-66-55-44-33-22-11-84-73-62-51-40-29-18-7-80\\ 79-68-57-46-35-24-13-2-75-64-53-42-31-20-9-82 \end{array} \right\} + \rightarrow$$

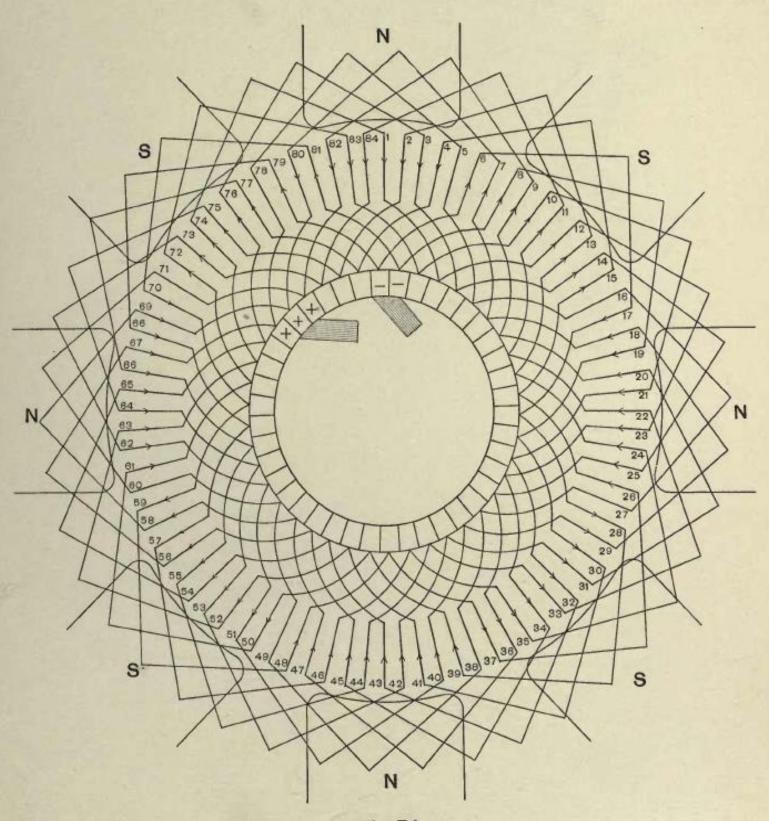


Fig. 74
TWO CIRCUIT, DOUBLE WINDING.



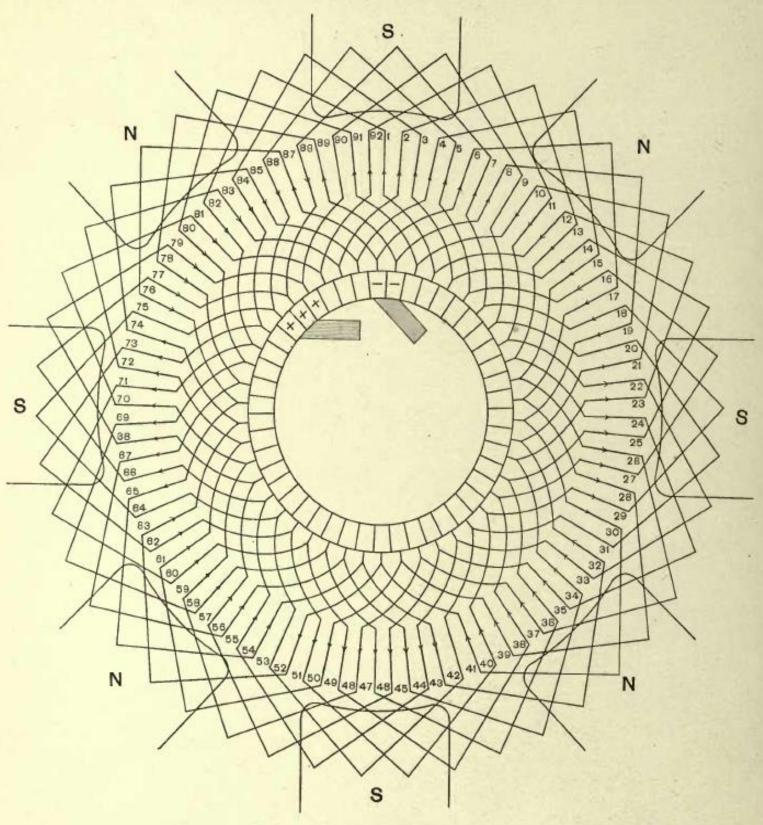


Fig. 75
TWO CIRCUIT, DOUBLE WINDING.

Figure 74 was obtained by using the negative sign in the formula—

 $C = ny \pm 2 m$ .

This is, as has been pointed out, rather wasteful of copper, and was only done to demonstrate the fact that in certain cases with a given number of conductors, either a singly or a doubly re-entrant, double winding may be used.

In Fig. 75, the positive sign was used. It will, however, not be necessary to analyze it, it not being materially different from Fig. 74.

Numerous interesting deductions concerning two-circuit, multiple-wound, drum armatures may be made from the data contained in the tables in Chapter XVIII.

## CHAPTER XI.

## THE SAYERS WINDING.

The armature coils of dynamos have, in addition to their function of establishing the electromotive force required external to the armature, the function of setting up in the arc of commutation an electromotive force to reverse the current in them as they successively pass the collecting brushes (by arc of commutation is meant the arc in which the current in the armature coils is reversed, the extent of this arc being determined by the length of the arc of contact of the collecting brushes). In the ordinary methods of armature winding the electromotive force for reversing the current in the coils is obtained by giving the collecting brushes an angular lead, the amount of which depends upon the distribution of the magnetic flux in the air gap, the coefficient of self-induction of the armature coils when in the arc of commutation, and the rate of change of the current in the coils, while the current is being reversed. In generators this angular lead is in such direction that the magnetomotive force of the armature is opposed to the magnetomotive force of the field magnets to an extent proportional to the angle of lead, in consequence of which the reversing field becomes of diminished intensity for an increase of current in the armature, when it needs to be increased.

Mr. Sayers, of Glasgow, has patented a winding in which the commutation of the current in the main armature coils is effected by an additional set of coils which may be termed commutating coils. These coils are applicable to any form of armature winding suitable for commutating machines. One of these coils is connected between each commutator bar and the connections joining the main armature coils in series with each other. These commutating coils are located on the periphery of the armature in such a position with respect to the main coils that the magnetomotive force of the main coils tends with increasing current to increase the flux through them, and further so that the magnetomotive force of the armature acts with the magnetomotive force of the field magnets instead of against it as in ordinary dynamos. It is possible, therefore, through a certain range of output to sparklessly operate a generator at constant voltage without changing the lead of the brushes or the excitation of the field magnets. It may be noted that when one of the main coils is short-circuited by the collecting brushes it is through two of these commutating coils, and the electromotive force from these coils effective for reversing the current in the main coil is the excess of the electromotive force generated in the leading coil over that in the following coil. The position, then, of the reversing field, if effective, is fixed as to angular extent between very narrow limits. It does not appear to the writers that the reversing field can be so localized for great changes of current in the armature as one might infer from reading the discussion of Mr. Sayers' paper at the Institution of Electrical Engineers. (See Vol. XXII., pages 377-441, Journal Ins. Elect. Engrs., London). Within certain limits, however, it appears that the magnetomotive force of the armature may be utilized in creating proper strength of reversing field.

This method, as applied to a bi-polar drum winding, is illustrated in Fig. 76. It will be seen to consist of a regular drum winding, with the difference that the connections from the winding to the commutator segments,

instead of consisting of short leads, consist of auxiliary force conductors which pass from the winding, backward, a short distance against the direction of rotation, and then parallel to the regular face conductors to the back of the armature. The conductor then passes forward in the direction of rotation, and again crossing the armature, is carried to the commutator segment.

In the diagram, the current in the coil  $A^2$  has just been reversed. The coil  $A^1$  is, by the two adjacent commutator segments under the brush, short-circuited while its main conductors are still moving through intense fields, tending to maintain the current in its original direction. But this short circuit contains, in series with the main coil, the two connections to the commutator segments, both of which are so linked with the magnetic flux from the pole piece, that electromotive forces are induced. Of the electromotive forces induced in the two commutator loops, that in the loop drawn in the figure is added to that of the short-circuited main coil, but this loop is farther out of the magnetic field than the remaining loop (not drawn) of the short-circuited section. This latter loop, leading from the segment next adjacent on the left of that shown at

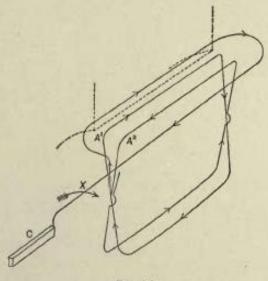


Fig. 76.

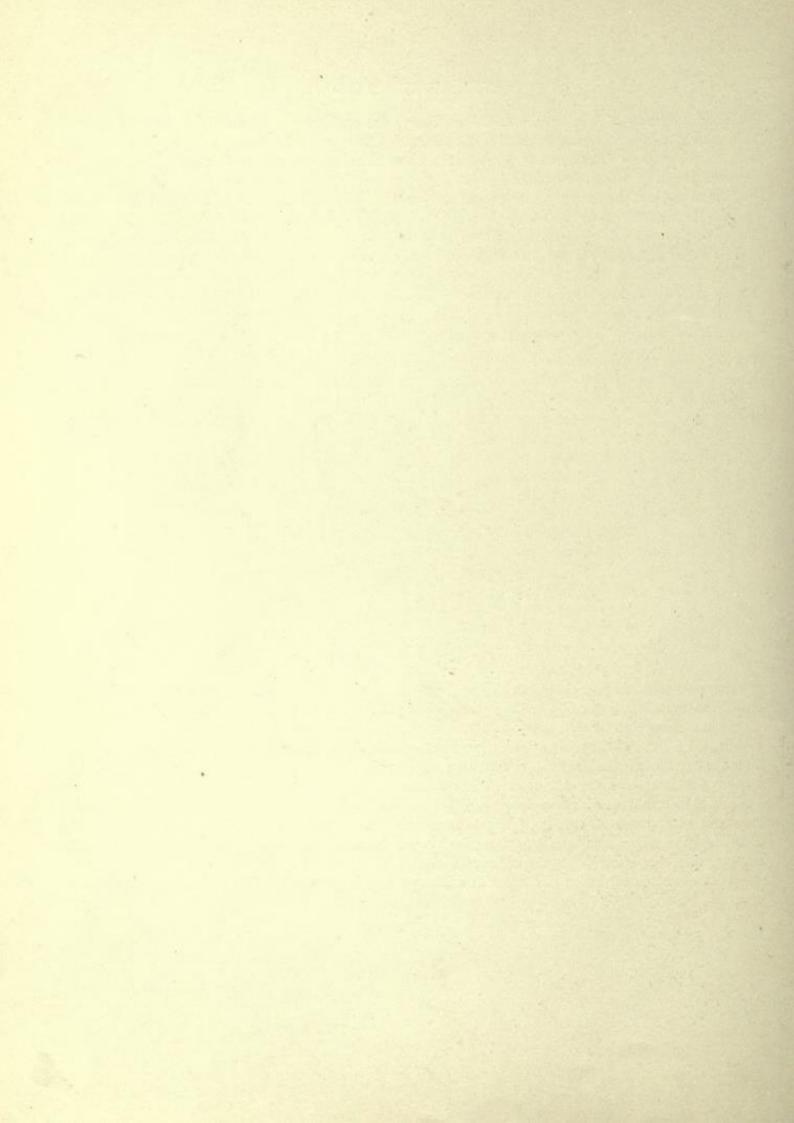
C, being well under the pole pieces, has induced in it a strong electromotive force, which opposes that in the rest of the short-circuited section, and enables a current to be generated in the direction of that in the half of the armature circuit of which it is soon to become a part.

In such a drum winding, Mr. Sayers refers to these commutator connections as "reverser bars." As they carry the current only during the short time that their corresponding sections are passing under the brushes, they may be of much smaller cross-section than the main conductors.

It will be seen from the above description that the winding is particularly adapted for use with ironelad armatures with very small air gaps, for the effectiveness of the arrangement is largely dependent upon the differential inductive action upon two successive reverser bars, and the more abrupt the demarcation of the magnetic flux, the greater will be this differential effect.

It should be clearly understood that this winding is equally applicable to rings, discs, and other types of armature.

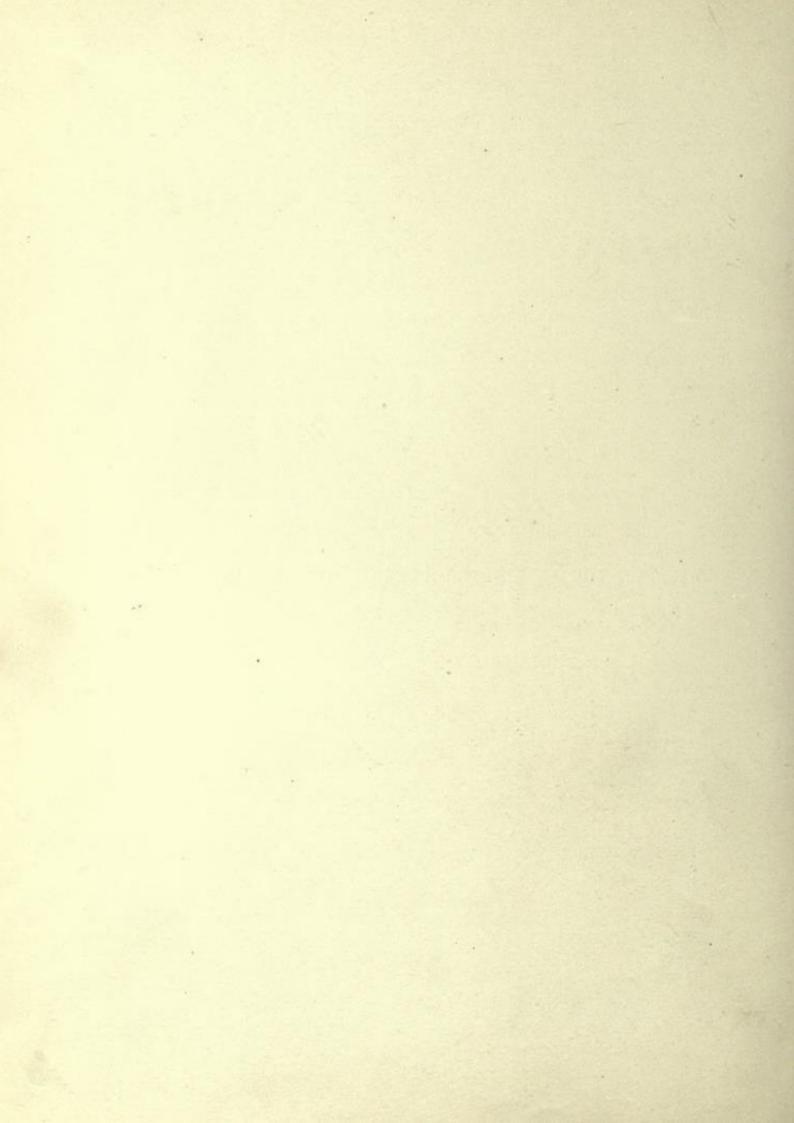




# PART II.

WINDINGS FOR ALTERNATING-CURRENT DYNAMOS AND MOTORS.





## CHAPTER XII.

#### ALTERNATING-CURRENT WINDINGS.

In general, any of the continuous-current armature windings may be employed for alternating-current work, but the special considerations leading to the use of alternating currents generally make it necessary to abandon the styles of winding best suited to continuous-current work, and to use windings specially adapted to the conditions of alternating-current practice.

Attention should be called to the fact that all the re-entrant (or closed circuit) continuous-current windings must necessarily be two-circuit or multiple-circuit windings, while alternating-current armatures may, and almost always do from practical considerations, have one-circuit windings, i.e. one circuit per phase. From this it follows that any continuous-current winding may be used for alternating-current work, but an alternating-current winding cannot generally be used for continuous-current work. In other words, the windings of alternating-current armatures are essentially non-re-entrant (or open circuit) windings, with the exception of the ring-connected polyphase windings, which are re-entrant (or closed circuit) windings. These latter are, therefore, the only windings which are applicable to alternating-continuous current, commutating machines.

Usually, high voltages are desired, and in such cases windings are generally adopted in which heavily insulated coils are imbedded in slots in the armature surface. Often, for single-phase alternators, one slot or coil per pole piece is used, as this permits of the most effective disposition of the armature conductors as regards generation of electromotive force. If more slots or coils are used, or, in the case of face windings, if the conductors are more evenly distributed over the face of the armature, the electromotive forces generated in the various conductors are in different phases, and the total electromotive force is less than the algebraic sum of the effective electromotive forces induced in each conductor. But, on the other hand, the subdivision of the conductors in several slots or angular positions per pole, or, in the case of face windings, their more uniform distribution over the peripheral surface, decreases the self-induction of the windings with its attendant disadvantages. It also utilizes more completely the available space and tends to bring about a better distribution of the necessary heating of core and conductors. Therefore, in cases where the voltage and the corresponding necessary insulation permit, the conductors are sometimes spread out to a greater or less extent from the elementary groups necessary in cases where very high potentials are used.

Windings in which such a subdivision is adopted, will be referred to as having a multi-coil construction, as distinguished from the form in which the conductors are assembled in one group per pole piece, which latter will be called uni-coil windings.

The terms uni- and multi-slot have been applied to alternating-current ironclad armatures, but the modified nomenclature described in the preceding paragraph will be preferable, in that it does not distinguish between armatures where the groups are arranged on the periphery, and those in which the groups are imbedded in slots. A little consideration will show the advisability of this nomenclature, as it will often permit one description to suffice for a winding which may be used either for ironclad or smooth-core construction.

It will be seen later, that in most multiphase windings, multi-coil construction involves only very little sacrifice of electrometive force for a given total length of armature conductor, and in good designs is generally adopted to as great an extent as proper space allowance for the insulation will permit.

Often in alternating current installations, step-up or step-down transformers, or both, are used, and in such cases the other extreme is approached, and the apparatus is built for very low voltages. This permits the use of very small space for insulation; and conductors of large cross-section, often arranged with only one conductor per group, are used. Here the multi-coil construction is less difficult, although still attended to some extent with the disadvantage of obtaining less than the maximum possible voltage per unit length of armature conductor.

Examples of windings adapted respectively to both of the above extremes will be given in the following chapters.

It will now be readily understood that the ordinary continuous-current windings are not, in the great majority of cases, adaptable to the work to be done. They should, however, always be kept in mind, and will often be found to work in nicely in special cases.

A class of apparatus, best termed alternating continuous-current, commutating machines, is now being found of much value in various ways. They are in a general way used for feeding continuous-current circuits, from single-phase or multiphase circuits (or vice versa), and also sometimes for feeding alternating circuits of one class (for example, single- or quarter-phase) from those of another (say three-phase). This type of armature may usually be best laid out by employing regular continuous-current windings and tapping them off in the proper manner. Examples will be given.

A wide variety of styles of armature construction have been employed in alternating-current machinery. Rings, drums (both ironclad and smooth-core), discs, and very many other types have been successfully built. Iron cores are used by some makers, and carefully avoided by others. Internal and external rotating parts have each found advocates. This great variety renders detailed treatment difficult, and in the following discussion it has been generally assumed that the windings are laid on the periphery of a drum, either on the surface, or imbedded in slots, and that the necessary connections are made at the ends of the armature. These peripheral conductors are represented diagrammatically by radial lines, and the end connections by crooked lines. Thus, re-entrant polygons drawn with heavy lines may be taken to represent coils of the desired number of turns, the lighter lines representing the connections of these coils to each other.

In the case of bar windings, no difficulty will be found in understanding the diagrams, as they correspond quite nearly to the continuous-current windings. Small, heavy circles in the middle of the diagram represent collector rings. If a winding is desired, for a disc or some other type, the diagrams will generally be found amply suggestive. Pancake coils and other types of windings, not specifically described, may be readily planned by slight modifications of the diagrams.

No examples have been given of gramme-ring alternating-current windings, as these may be found in text books, and are so easily understood as to require no discussion.

Before concluding these general considerations, it is desirable to emphasize the following points regarding the relative merits of uni- and multi-coil construction:—

With a given number of conductors arranged in a multi-coil winding, less terminal voltage will be obtained at no load than would be the case if they had been arranged in a uni-coil winding, and the discrepancy will be greater in proportion to the number of coils into which the conductors per pole piece are subdivided, assuming that the spacing of the groups of conductors is uniform over the entire periphery.

Thus, if the terminal voltage at no load be taken as 1 for a uni-coil construction, it will, for the same total number of conductors, be .707 for a two-coil, .667 for a three-coil, .654 for a four-coil, etc.

But when the machine is loaded, the current in the armature causes reactions which play an important part

in determining the voltage at the generator terminals, and this may only be maintained constant as the load comes on, by increasing the field excitation, often by a very considerable amount. Now, with a given number of armature conductors, earrying a given current, these reactions are greatest when the armature conductors are concentrated in one group per pole piece, that is, when the uni-coil construction is adopted, and they decrease to a considerable degree as the conductors are subdivided into small groups distributed over the entire armature surface, that is, they decrease when the multi-coil construction is used. The ratios given above for the relative voltages at no load, for uni- and multi-coil construction, do not, therefore, represent the relative values of the windings under working conditions, and it is believed that careful consideration should in many cases be given to both styles of winding, before deciding upon the one best suited for the purpose.

Multi-coil design also results in a much more equitable distribution of the conductors, and, in the case of ironelad construction, permits of coils of small depth and width which cannot fail to be much more readily maintained at a low temperature for a given cross-section of conductor, or, if desirable to take advantage of this point in another way, it should be practicable to use a somewhat smaller cross-section of conductor for a given temperature limit. And similarly, when we consider smooth-core construction, we find that the distribution of conductors over the entire surface carries with it great advantages from a mechanical standpoint.

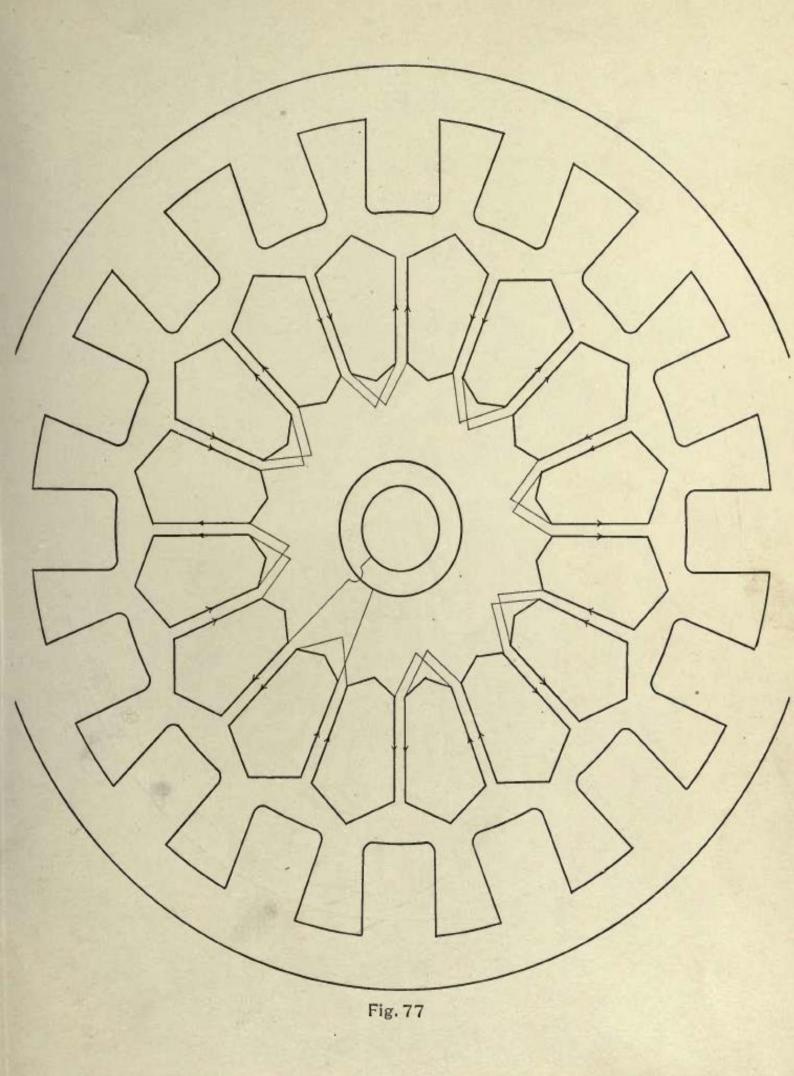


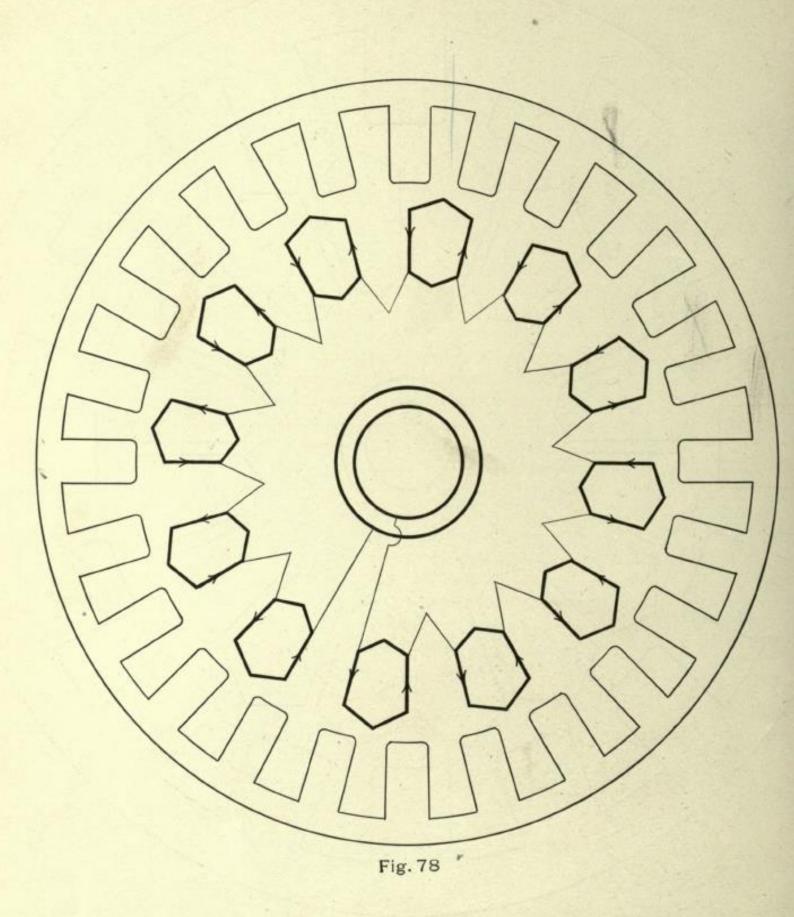
## CHAPTER XIII.

### SINGLE-PHASE WINDINGS.

FIGURE 77 is a diagram of a winding for single-phase alternating-current generators and synchronous motors, which has been very extensively used. It has one group per pole piece, consisting of adjacent halves of two coils of the proper number of turns. These are interconnected as shown by the light lines. The adjacent halves of the two coils are usually arranged side by side, but it might sometimes be of advantage to place them one over the other. The arrangement of two coils side by side has been satisfactorily applied in various types of ironclad armatures. In Figs. 102 and 119 are given examples of this style of winding connected respectively for quarter-phase and for three-phase work. It should be noted, however, that the same armature can be used for three-phase purposes only by having fields with different numbers of pole pieces.

The avoidance of crossings at the ends, and the extreme simplicity of this style of winding, are its chief advantages.





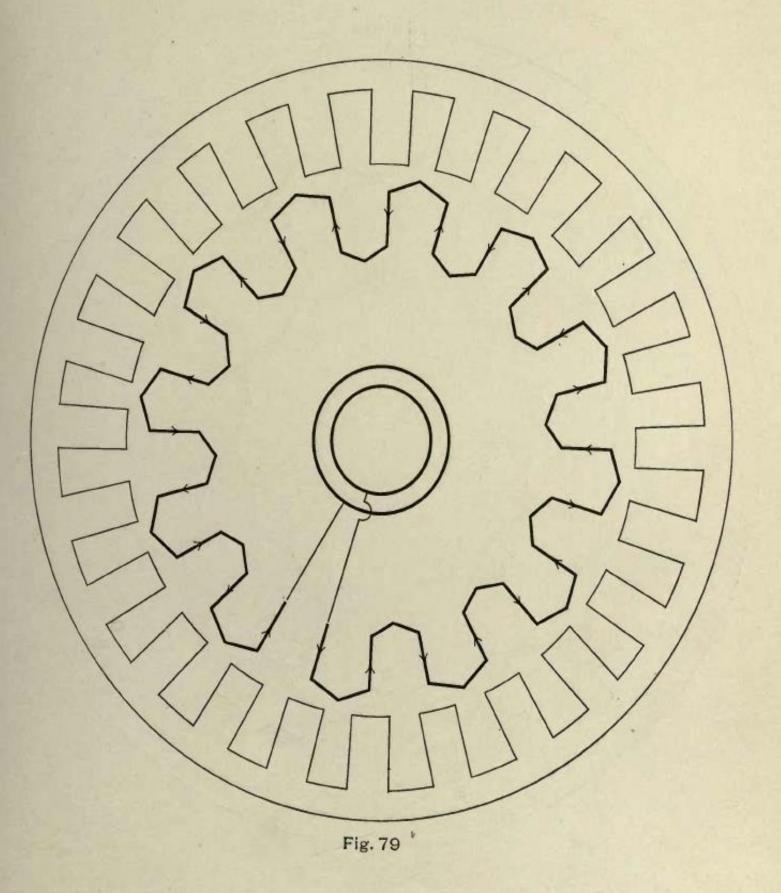
In Fig. 78 is given another uni-coil winding, but here only one coil is placed in each slot. In many cases this might be preferable to the arrangement shown in Fig. 77, but the ends of the armatures are not so completely occupied by the ends of the coils, which wastes room and tends to bring about a less even distribution of the loss by heating. The use of only half as many coils is, of course, generally an advantage, on account of simplicity, but it is usually necessary for each coil to be wound deeper, which is objectionable from a thermal standpoint, as well as from the fact that a greater depth of space has to be allowed for the winding at the ends of the armature.

It should not be overlooked that if half the number of pole pieces is odd, the armature coils could not be connected up in two parallels, which would in practice be a very considerable objection, as it would limit the use of the armature for other purposes than that contemplated in laying out the original design.

One feature of this winding worthy of consideration is the great ease of insulation, it being, in this respect, superior to Fig. 77, one of the groups of which consists of adjacent halves of two coils, having between them the entire voltage of the armature.



Figure 79 is a bar winding, with one bar per pole piece, corresponding to the coil winding of Fig. 78. This would be used for low voltages, and in the case of generators of large capacity, such windings are practicable for high voltages. It is typical of the simplest form of a multipolar, single-phase alternator, and has been used in some very large machines.



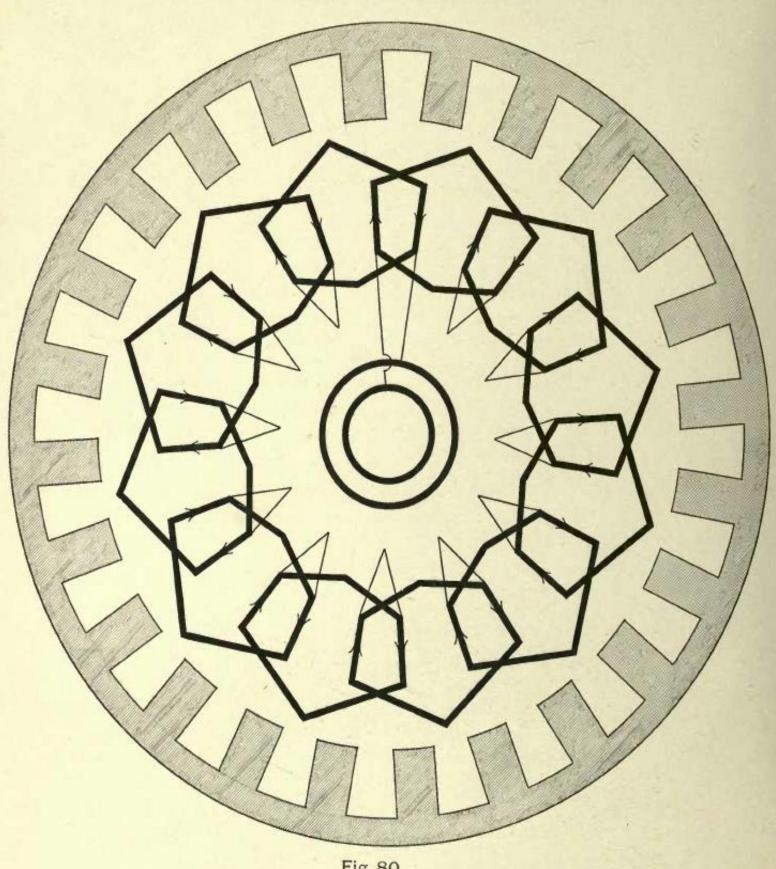


Fig. 80

Figure 80 is another uni-coil winding. It is given largely as a matter of interest; for, as will be seen, it has undesirable crossings and very long end connections, which would be very wasteful of copper unless the length of the magnet cores parallel to the shaft is great compared with the length of the pole arc. Even in such a case there would be no advantage over Fig. 78, unless for the fact that Fig. 80 is a very good winding for a three-phase alternator of one-third the number of poles, and the case might occur where it would be of advantage to use the same armature and winding for both cases. This would make an excellent three-phase winding for one-third as many poles, and would then be similar to the three-phase winding given in Fig. 116.

The corresponding diagram for a bar winding, with one bar per pole piece, is sufficiently evident from Fig. 80, and, in view of its unimportance, will not be given.



The following diagrams are multi-coil, single-phase alternators. As a class they have been very thoroughly discussed in the general remarks of the preceding chapter.

Figure 81 represents a very simple two-coil winding. It is to be noted that this winding is mechanically identical, with the exception of the interconnection of the coils, with the winding of Fig. 78, but it is put in a frame with half as many poles as there are groups of conductors, instead of, as was the case in Fig. 78, being laid out for a frame with a number of poles equal to the number of groups of conductors.

As already pointed out, such multi-coil windings do not at no load generate so great an electromotive force per unit of length of face conductor, as uni-coil windings. It has, however, been also shown on page 164 that this objection does not have such great weight as would at first sight appear to be the case.

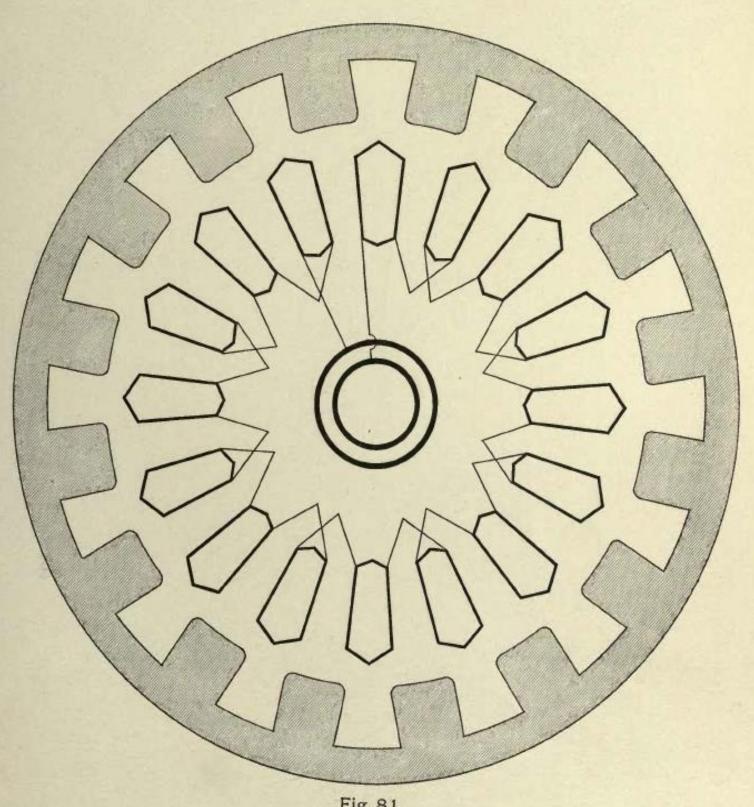


Fig. 81

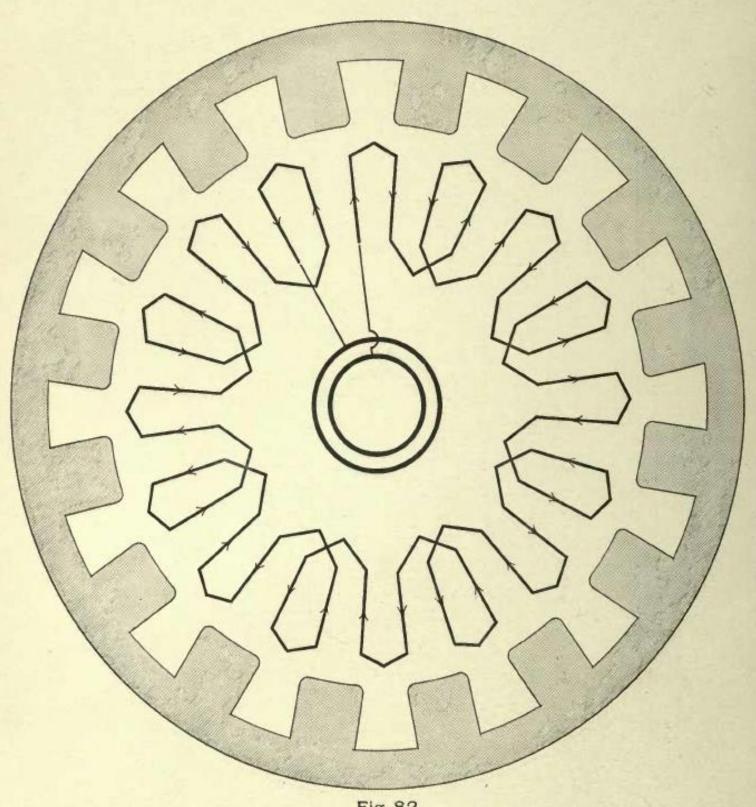


Fig. 82

Figure 82 gives a bar winding with two bars per pole piece. It corresponds to the coil winding of Fig. 81. These two windings (Figs. 81 and 82) could probably be used to advantage in many cases, but, of course, their disadvantages should be carefully considered.



Figure 83 represents another two-coil winding. It would seldom be used, as it has the faults and lacks the merit of the winding given in Fig. 81.

If, however, the coils, instead of being evenly spaced, were brought into groups of two, not very far apart, it would, to some extent, have part of the advantages of the uni-coil construction, and would partly overcome some of the faults of the latter. If modified in this way, it would partake of the nature of the windings given in Figs. 97, 98, and 99, and the remarks made in connection with these figures should be referred to.

If Figs. 81 and 82 should be similarly treated (that is, if the coils should be brought into groups of two coils each, not very far apart), the result would be a winding comparable to those given in Figs. 97 and 99.

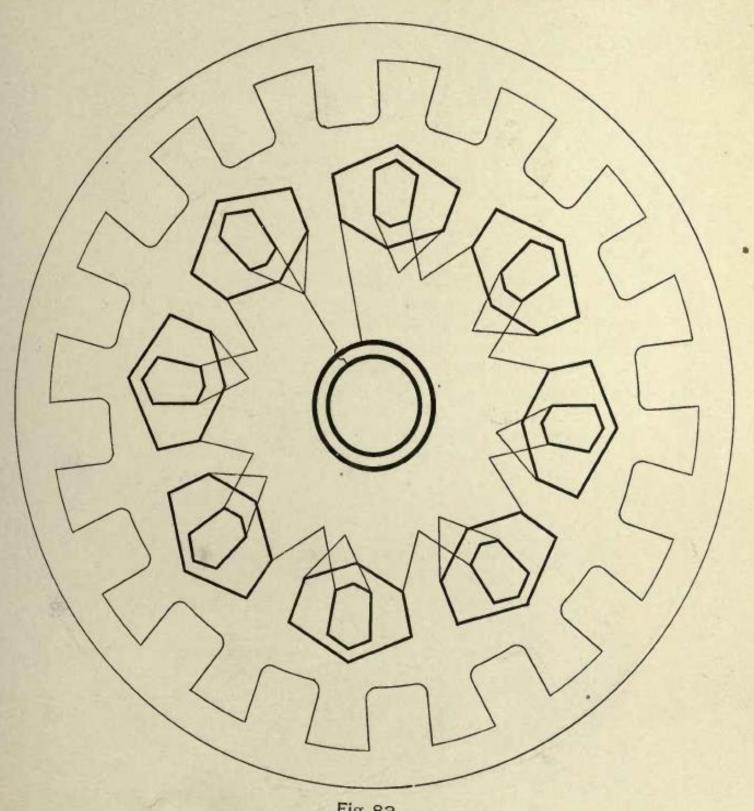


Fig. 83

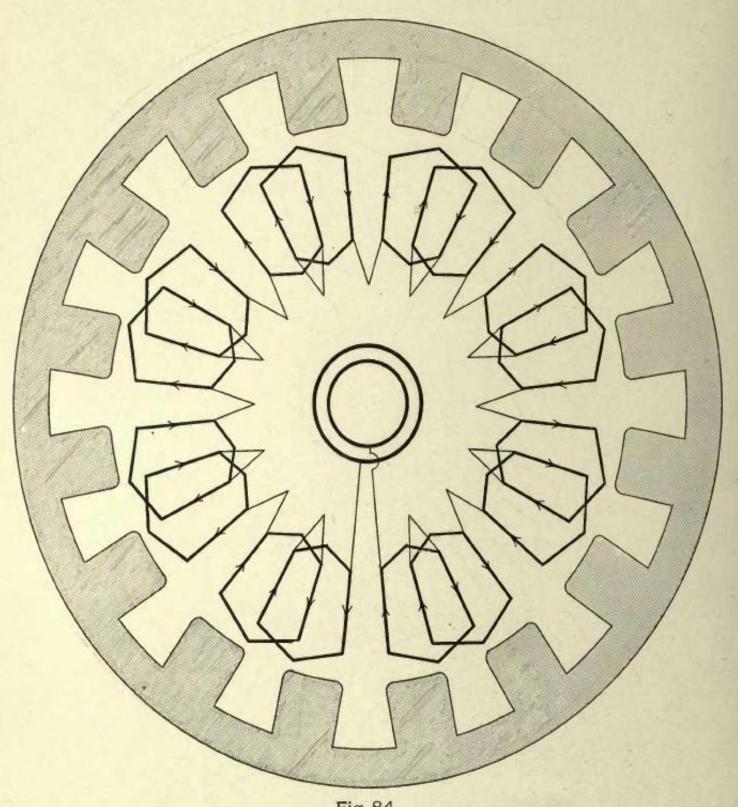


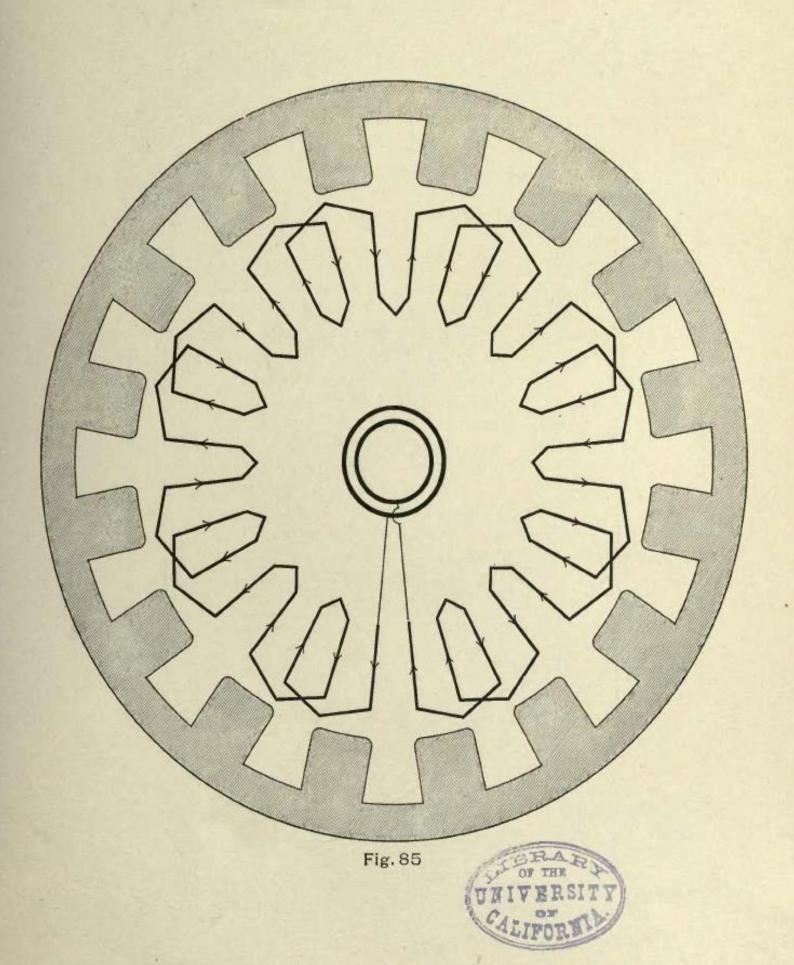
Fig. 84

Figure 84 is a diagram of another two-coil winding. It is connected as a single-phase alternator, but except for the manner of interconnection of the coils it is identical with the quarter-phase winding given in Fig. 100. It will give the same voltage as would Fig. 100, if the two components of the quarter-phase winding should be connected in series.

For this reason (that is, because when reconnected, it makes a good quarter-phase winding), it might sometimes be used, but of course, would, as stated in connection with previous windings, require a greater length of wire to generate the same voltage than a uni-coil winding, and would naturally have a greater armature self-induction. But, of course, the decrease in self-induction due to the multi-coil construction would somewhat compensate for this increase.

Figure 85 gives a diagram for a single-phase bar winding, corresponding to Fig. 84. It is only of interest as showing that it is identical with Fig. 82, except that the long-end connections which were at the collector ring end in Fig. 82 are now at the other end.

It should be noted that all these multi-coil windings now under consideration would, for a given terminal voltage, require much more field excitation at no load than corresponding uni-coil windings. But at full load they would, in some cases, require little if any more field excitation than would be the case with uni-coil windings. As a result of these considerations it will be seen to be necessary in any particular case to observe the requirements for the field excitation as regards permissible regulation, heating, etc., when deciding upon the type of armature winding to adopt.



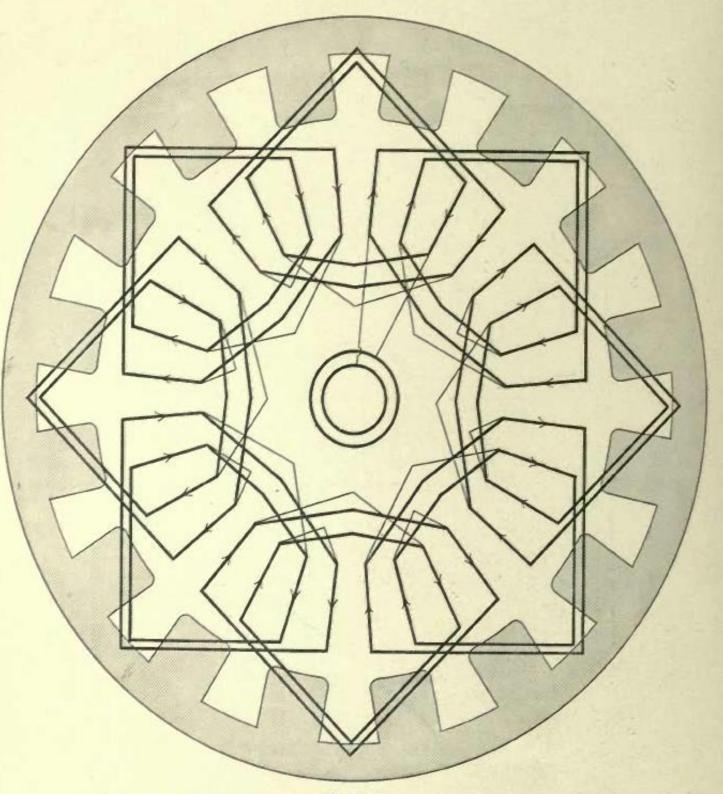


Fig. 86

Figure 86 should be compared with Fig. 80. It is quite like the latter, except that it has two coils per pole piece instead of one. It would, of course, not be used, as it has such long end connections.

The number of poles is sixteen. Such a winding with twelve, eighteen, or twenty-four poles could be used in a three-phase armature of one-third the number of poles by merely changing the interconnections of the coils. Figure 123 gives such a diagram for a three-phase alternator in an eight-pole frame.

The mechanical arrangement of such windings as those given in Figs. 80, 86, and 123 is exceptionally good, although in the case of Figs. 80 and 86, they are much less simple, as single-phase windings, than those that do not cross.

Figure 87 represents a winding with two groups of coils per pole, and two coils per group. It will be seen to be identical with the two-phase winding of Fig. 103, except that it is connected up as a single-phase winding. With the exception of the sequence of interconnection of the coils, it may be considered to be two windings like Fig. 77, one of which is displaced 90°, so that its conductors lie half way between those of the other.

Its end connections permit of good mechanical arrangement; very much, in fact, like that of Figs. 80, 86, and 123.

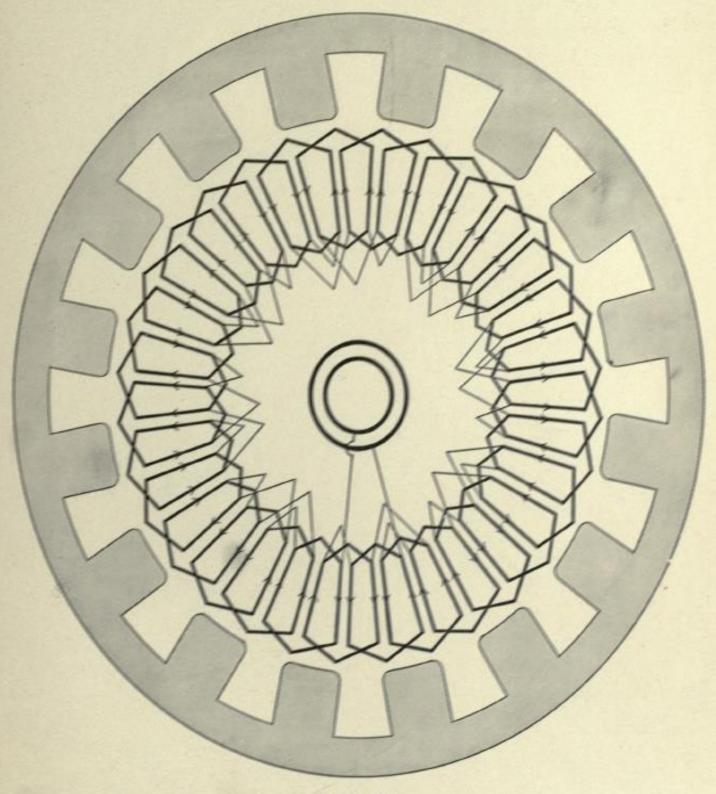


Fig. 87

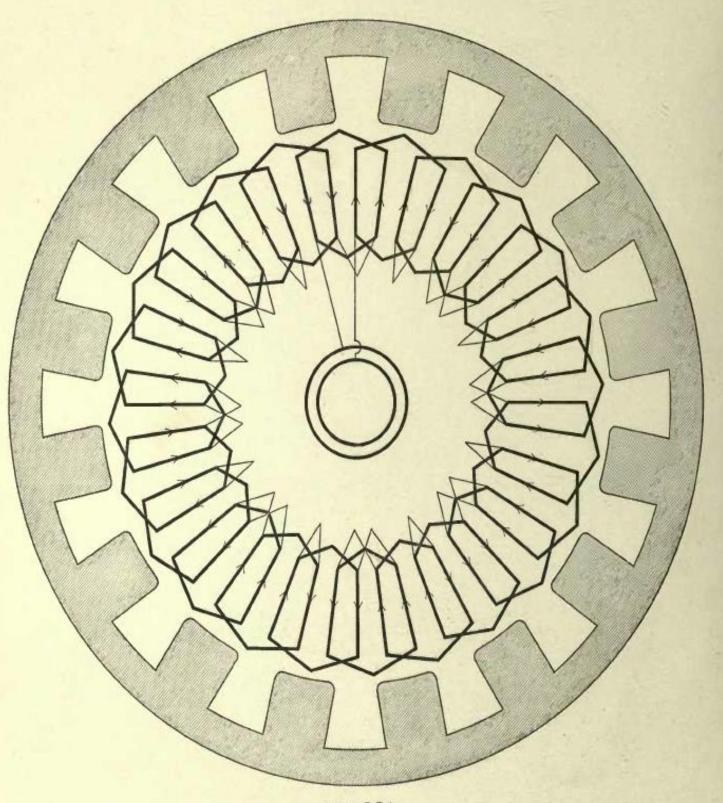


Fig. 88'

Figure 88 shows a useful three-coil winding. It has all the advantages and disadvantages already noted of multi-coil armatures.

The end connections can be very nicely arranged, so as to permit of winding on forms and slipping them into slots. Only two different shapes of forms are necessary; one-half of the coils would be wound in one of them, and the rest in the other.

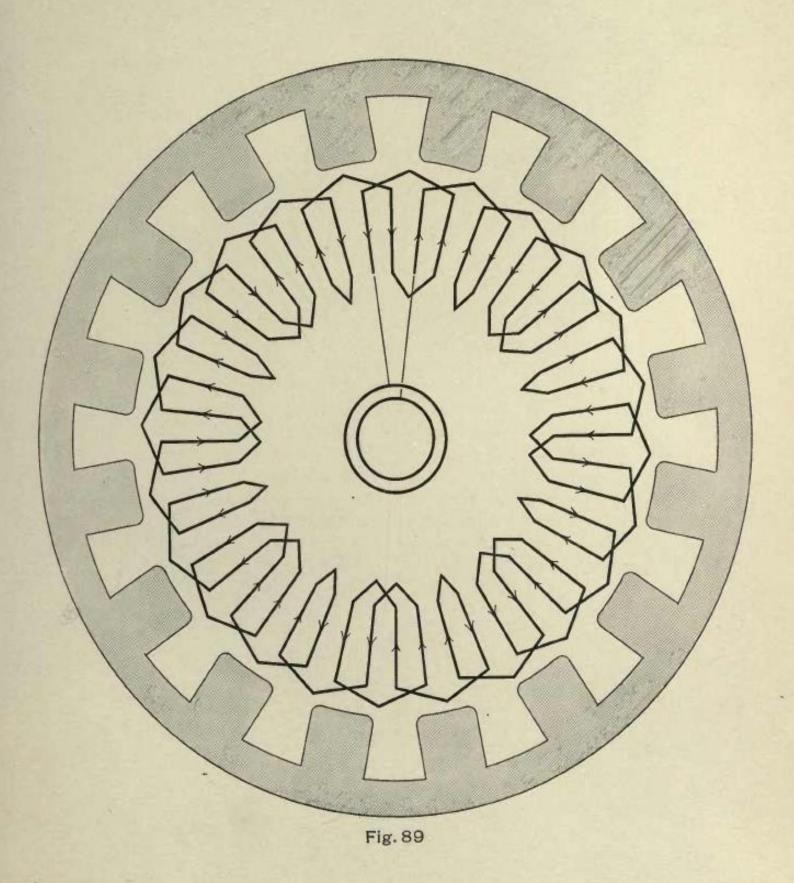
It will be seen that it is really the three-phase winding of Fig. 116 connected up as a single-phase winding. For this reason, among others, it might be expected to be of service where it would be of advantage to have armatures which could be used interchangeably for single- or three-phase work. Most three-phase windings could, of course, be similarly used.

As a single-phase winding per se, Fig. 88 is excelled by the windings of Figs. 92 and 94, which require a smaller length of end conductors.



Figure 89 is the bar winding corresponding to the coil winding of Fig. 88. It is not a generally useful winding. Among other faults it has three different lengths of end connections, half of them being very long. In this respect it is excelled by the winding given in Fig. 93. The end connections at one end are perfectly regular, but this would seldom be considered to compensate for the needlessly great length of copper employed.

This winding is an example of the importance of thoroughly examining many diagrams before adopting a winding for a certain case; for it is not at once apparent that this winding could be improved upon, and if thought of first, might be chosen without further investigation.



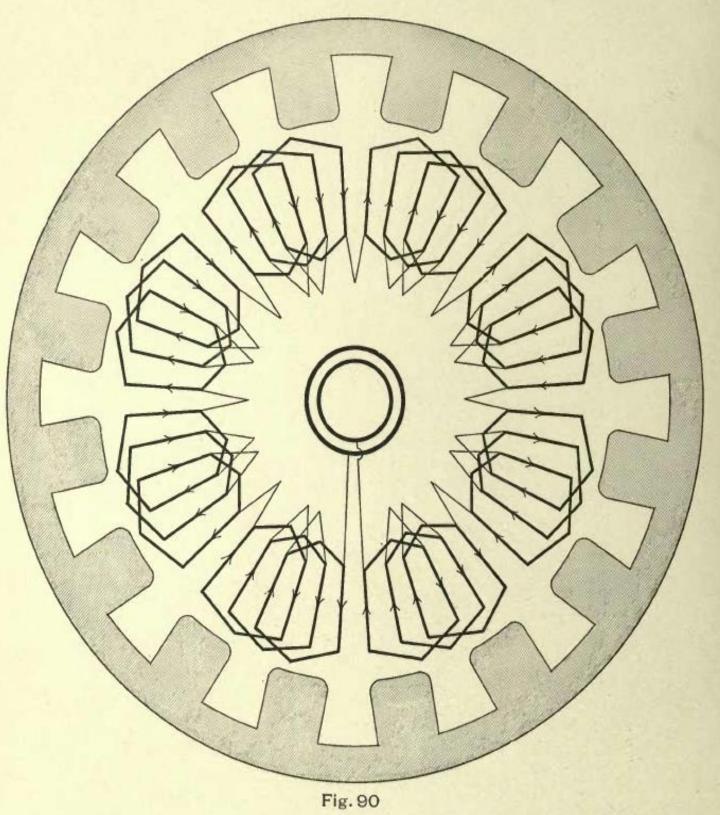


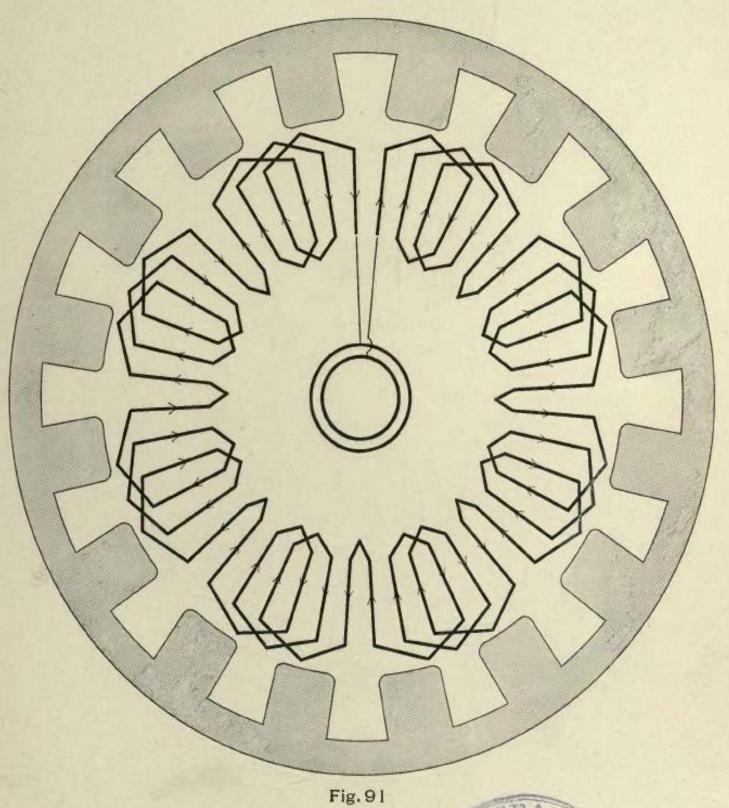
Figure 90 gives a coil winding very similar to that of Fig. 88. But the end crossings would render it very inconvenient, and the space at the ends of the armature is not so well utilized as it was in Fig. 88. This would tend to an undesirable concentration of the heating.

Unlike Fig. 88, the winding would not interfere with the armature, being made in segments for convenience of shipment. But Figs. 92 and 94, which require less copper in the end connections, also possess this advantage, Fig. 94 to the greatest extent of all.



Figure 91 has all the faults of Figs. 89 and 90. It is the bar winding corresponding to Fig. 90. It is inferior to the winding shown in Fig. 93.

It has the advantage that the winding is more symmetrical as a whole than many better windings, and it is for this reason readily constructed and connected up, with little liability of error. It is a great help for the winder to be able to intelligently perform his work, and windings that are, electrically and mechanically, to a small extent inferior, might in some cases consistently be adopted because of the simplicity of winding. They also permit of the more ready locating and correcting of faults that are liable to develop during the practical operation of the machinery.





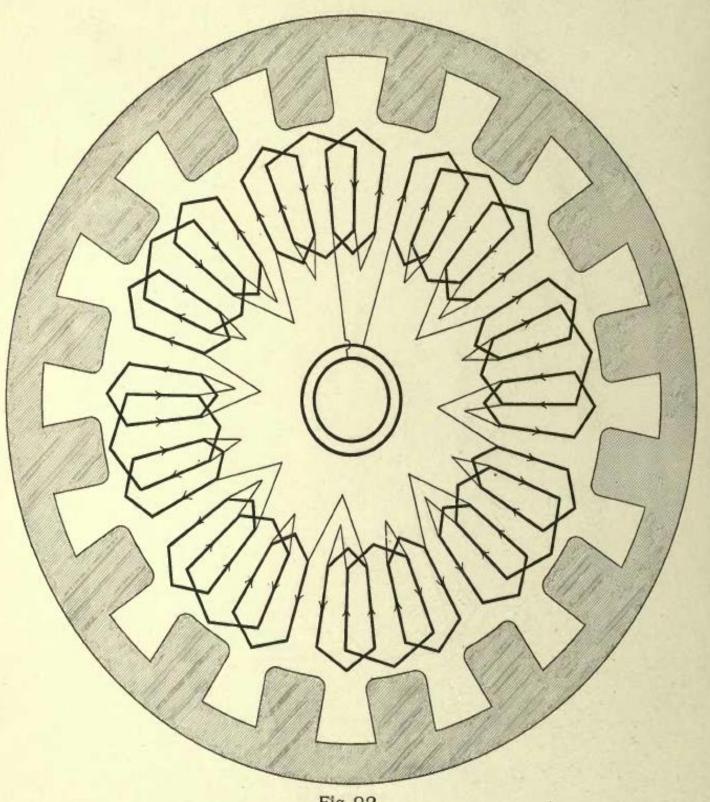


Fig. 92

Figure 92 is another three-coil winding. It gives the same results as Figs. 88 and 90, but with less copper, as it has shorter end connections. It is also simpler, as there is much less overlapping at the ends. Only two sizes of coils are necessary.

The chief point of inferiority to Figs. 88 and 90 is that it cannot be connected up as a three-phase armature.

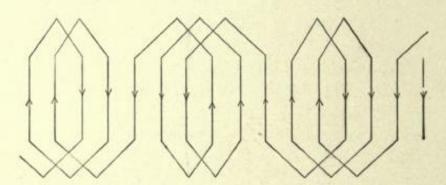
Even Fig. 92 is not so good as Fig. 94 (to be described later), which latter has still shorter end connections and less crossings.

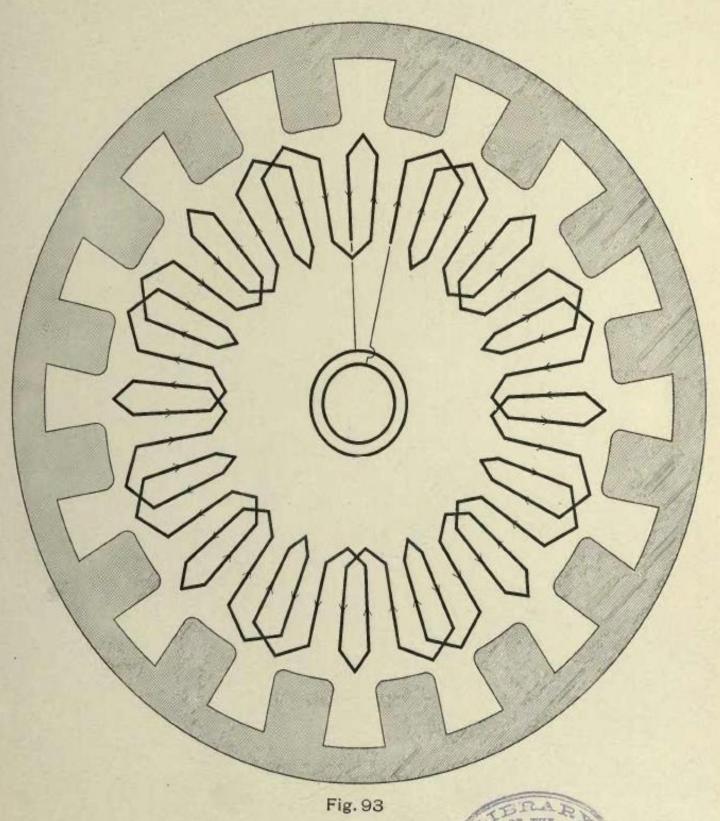
There is no good bar winding corresponding to Fig. 92.

Figure 92 possesses the advantage noted in the discussion of Fig. 90, that the armature may be built and shipped in sections without interfering with the winding.

Figure 93 is the best bar winding for three bars per pole piece. It is distinctly superior to Figs. 89 and 91, as it has much shorter end connections. It requires, moreover, only two different lengths of end connections, whereas Figs. 89 and 91 each require three.

The following diagram is a section of a bar winding with five bars per pole piece: —







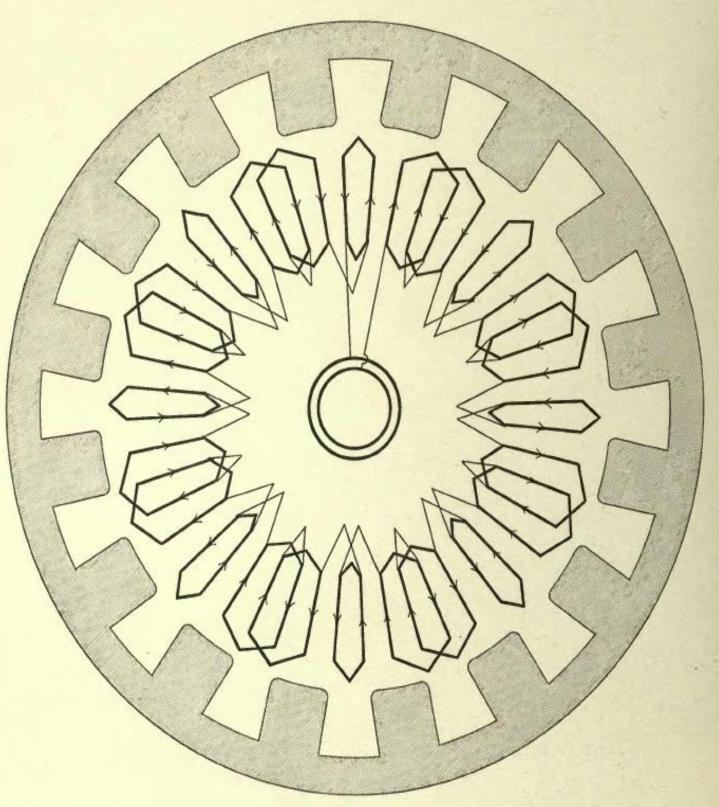


Fig. 94

Figure 94 is the coil winding corresponding to the bar winding of Fig. 93.

This coil winding is superior to that of Figs. 88, 90, and 92, in that it gives the same result with much shorter end connections and with fewer crossings of the end connections. Like Fig. 92, it cannot be connected up as a three-phase alternator, it being in this respect inferior to Figs. 88 and 90.

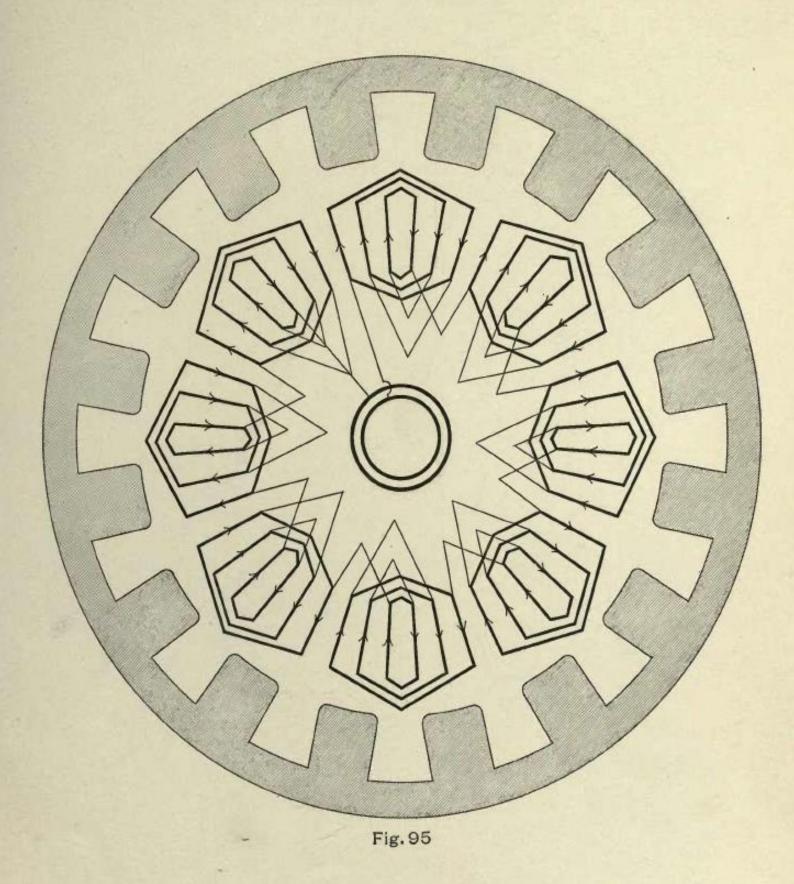
The winding of Fig. 94 could readily be built in sections in cases where it would be necessary to ship the armature in segments.



Figure 95 is a coil winding electrically equivalent to Figs. 88, 90, 92, and 94.

Windings of this class may readily be derived from the example given in Fig. 95, for any desired number of coils per pole piece. It often works out well from a mechanical standpoint, and although the end connections are necessarily longer than in the preceding windings, it will frequently be found useful.

The various coils might with advantage be grouped to a greater or less extent, in accordance with the principles exemplified in Figs. 97, 98, and 99, which, together with the accompanying text, should be consulted in this connection.



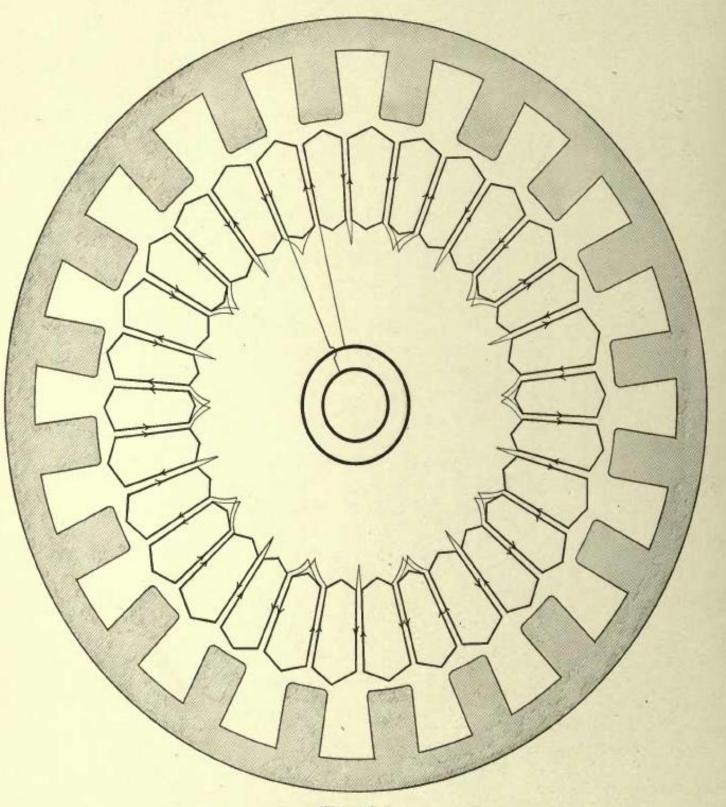


Fig. 96

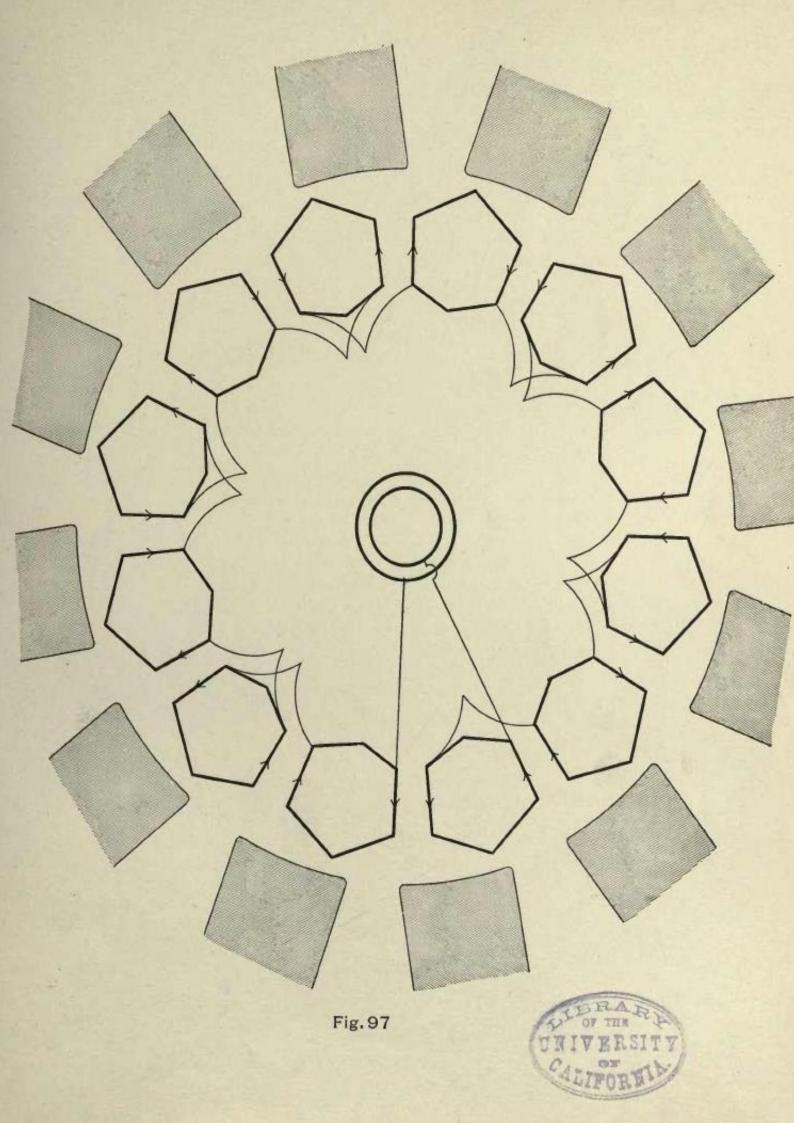
Figure 96 gives a coil winding with one and one-half coils per pole piece. It has two coils per group. It is really a winding such as Fig. 77, put in a field with two-thirds as many poles as the armature has coils. Thus in Fig. 96 there are thirty armature coils and twenty field poles. There is disadvantageous counter-induction which makes the use of more armature copper necessary than would be used in a uni-coil winding. The armature could, however, be used interchangeably in fields with n and with  $\frac{2}{3}n$  poles, which property permits of the use of the armature in cases where different speeds or periodicities may be called for.

Also by changing the interconnections of the coils, an excellent three-phase armature is obtained. The three-phase connections of such a winding are given in Fig. 119.

Moreover, owing to the fact that when one side of a coil is under a field pole, the other is between two poles, the selfinduction of such a winding is low, and is fairly uniform for all positions of the armature. Many of the multi-coil windings given heretofore have been somewhat undesirable by reason of the counter-induction, which made it necessary to have a greater length of conductor for a given voltage than would have been necessary if the conductors had been concentrated in one coil per pole piece.

Figure 97 is a winding which, while retaining to a great extent many of the advantages of multi-coil windings, is usually as good with regard to its freedom from counterinduction as a uni-coil winding with evenly spread coils.

It is in fact one of the two windings of the quarter-phase diagram of Fig. 104.



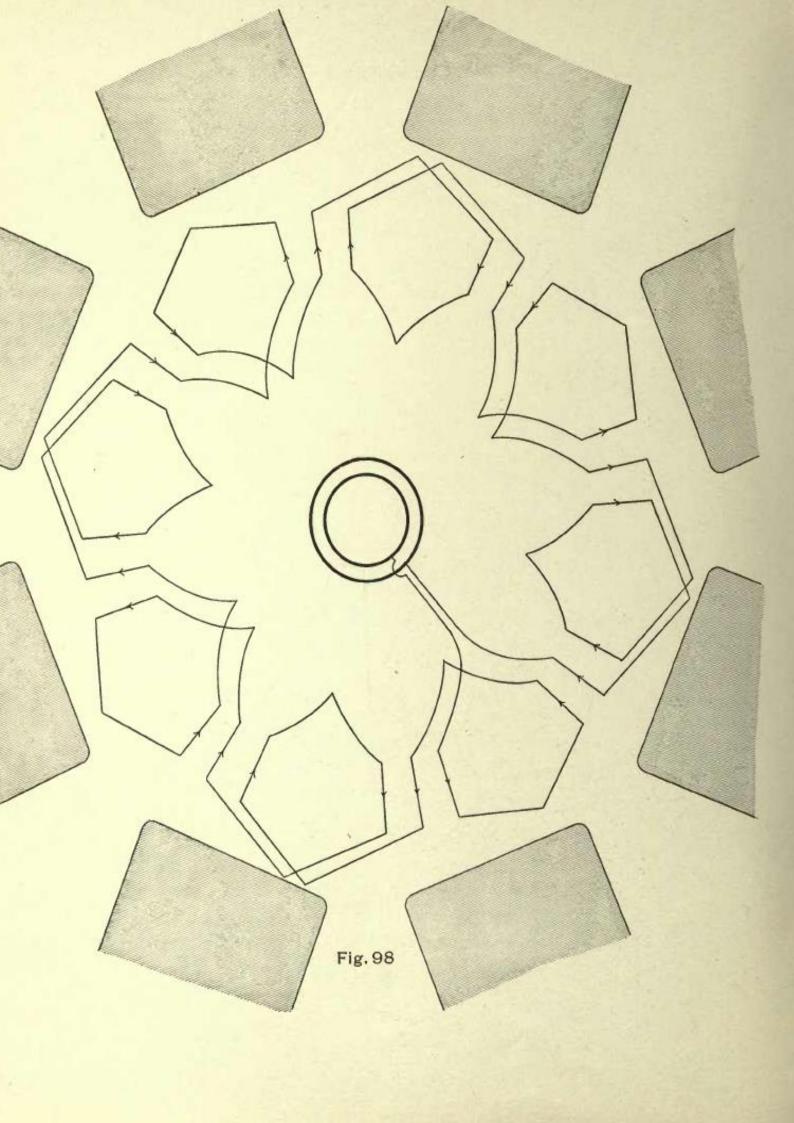


Figure 98 does not differ essentially from Fig. 97 as far as regards the point that it is intended to illustrate. It, also, is one of the two windings of a quarter-phase armature, being in fact derived from the quarter-phase diagram of Fig. 112.

Other excellent diagrams of this type may be derived by considering one of the two windings of the quarter-phase armatures shown in Figs. 105, 106, 107, and 111.



Figure 99, like its predecessors, Figs. 97 and 98, has its coils arranged in groups in the periphery of the armature. It has to some extent their advantages and disadvantages. It differs from them in utilizing two-thirds of the available space, instead of one-half, and is more of a compromise with the uniformly distributed windings.

It is obvious that windings such as the three just given may readily be derived from any of the evenly distributed multiphase windings by simply discarding one or more of the windings belonging to the respective phases of such diagrams. They may also be derived from many of the single-phase windings by shifting the coils laterally from the normal position into the desired groups.

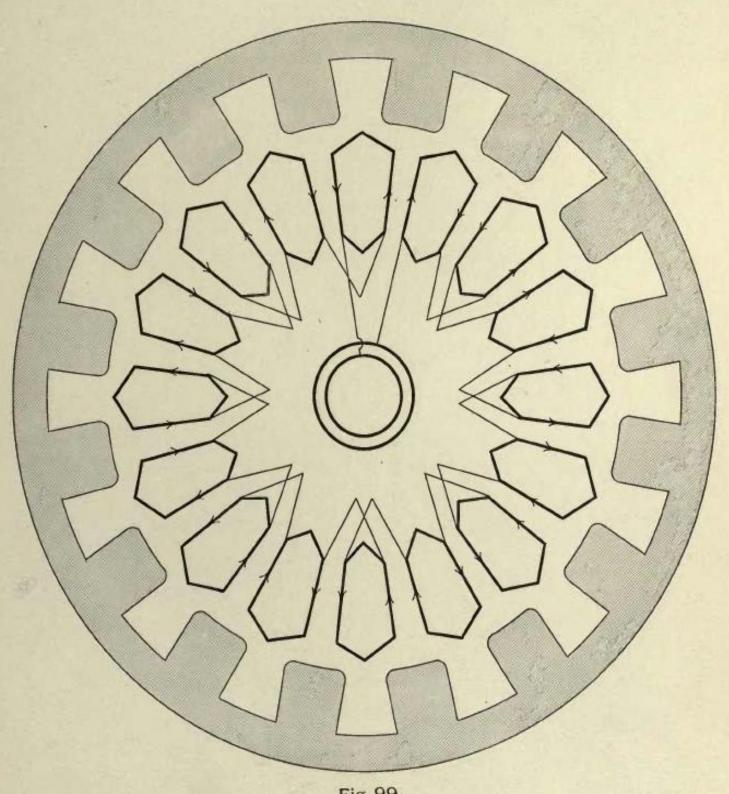
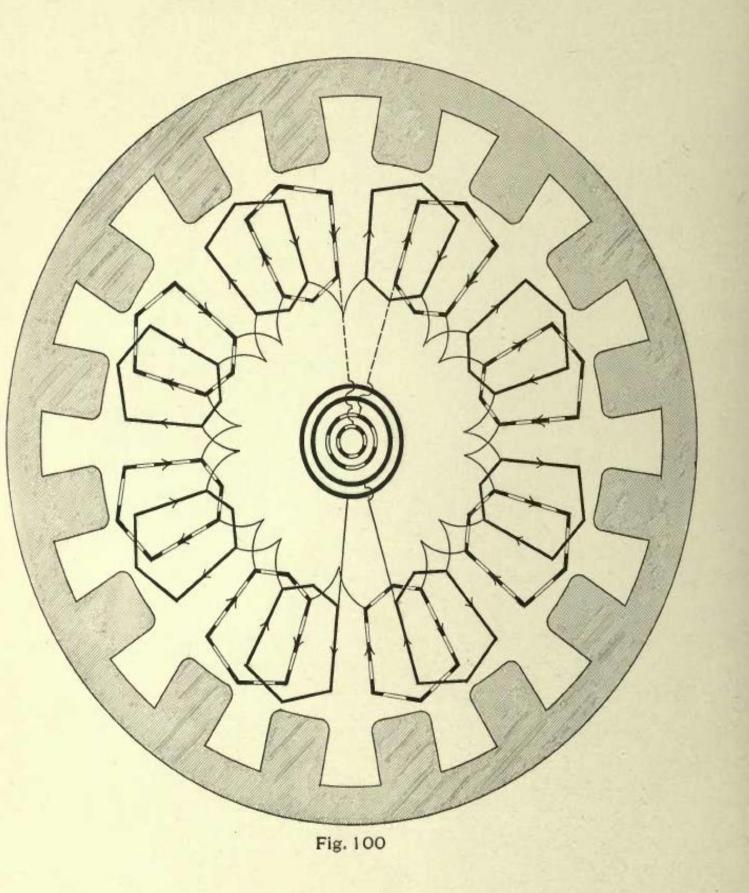


Fig. 99



## CHAPTER XIV.

## QUARTER-PHASE WINDINGS.

FIGURE 100 represents a quarter-phase coil winding with one group of conductors per pole piece per phase. In accordance with the nomenclature already adopted, this would be known as a uni-coil winding; although it has but one coil per pole piece per phase, it has two coils per pole piece.

The two windings are represented, respectively, by full and broken lines. The winding is quite simple, but has the objection of crossings at the ends. In this respect it is inferior to the style of winding represented by the diagram of Fig. 102.

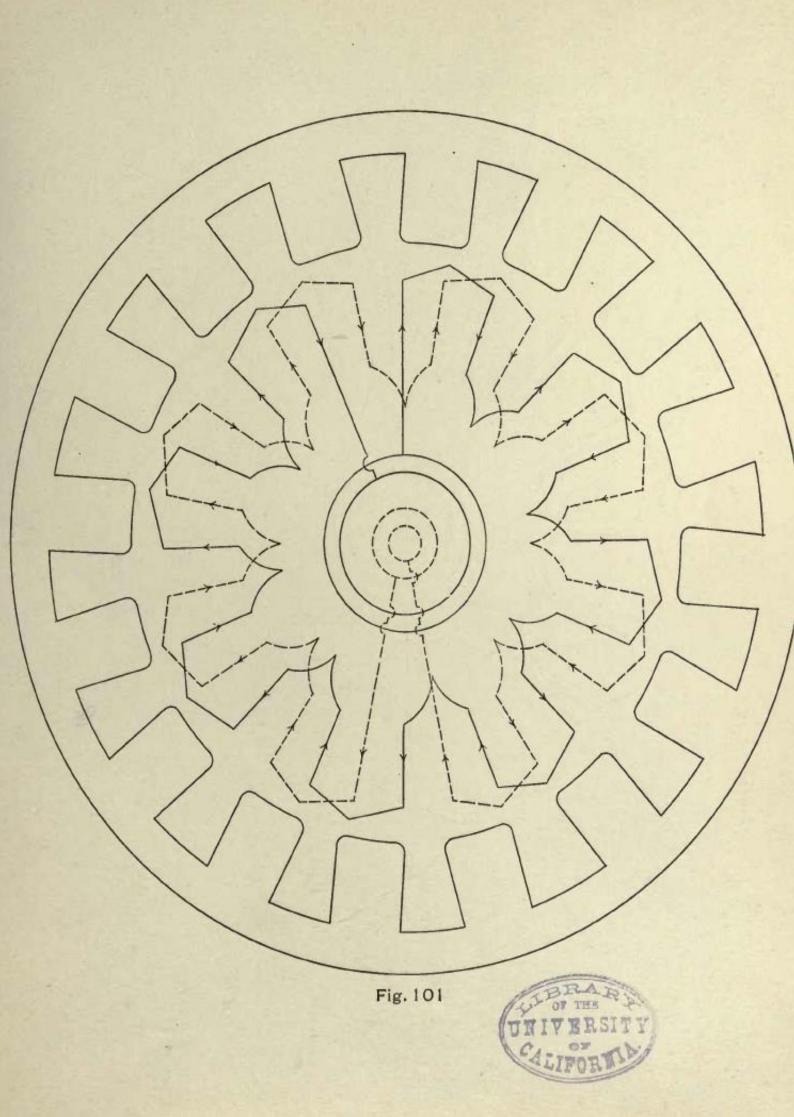
Three collector rings could be used, one of them being common to each winding. In the diagrams, however, four collector rings will be shown, this being the method now generally used. In connection with a system employing three collector rings, the standard quarter-phase commutating machines (to be described later) could not be used.

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Figure 101 is the bar winding corresponding to Fig. 100. It does not well utilize all of the available space on the armature ends. This is generally not a great objection in the case of uni-coil windings, as there is in such cases plenty of room on the ends, but, other things being equal, it is of course preferable to have windings uniformly distributed at the ends as well as on the surface. In this connection Fig. 109 should be studied, and it will be seen that by placing two conductors in a group a perfectly symmetrical design is obtained with one group per pole piece.

A decided objection to this arrangement would be that adjacent conductors would have between them large differences of potential, whereas in Fig. 101 there are but few points in which neighboring conductors have between them any considerable percentage of the total terminal voltage.



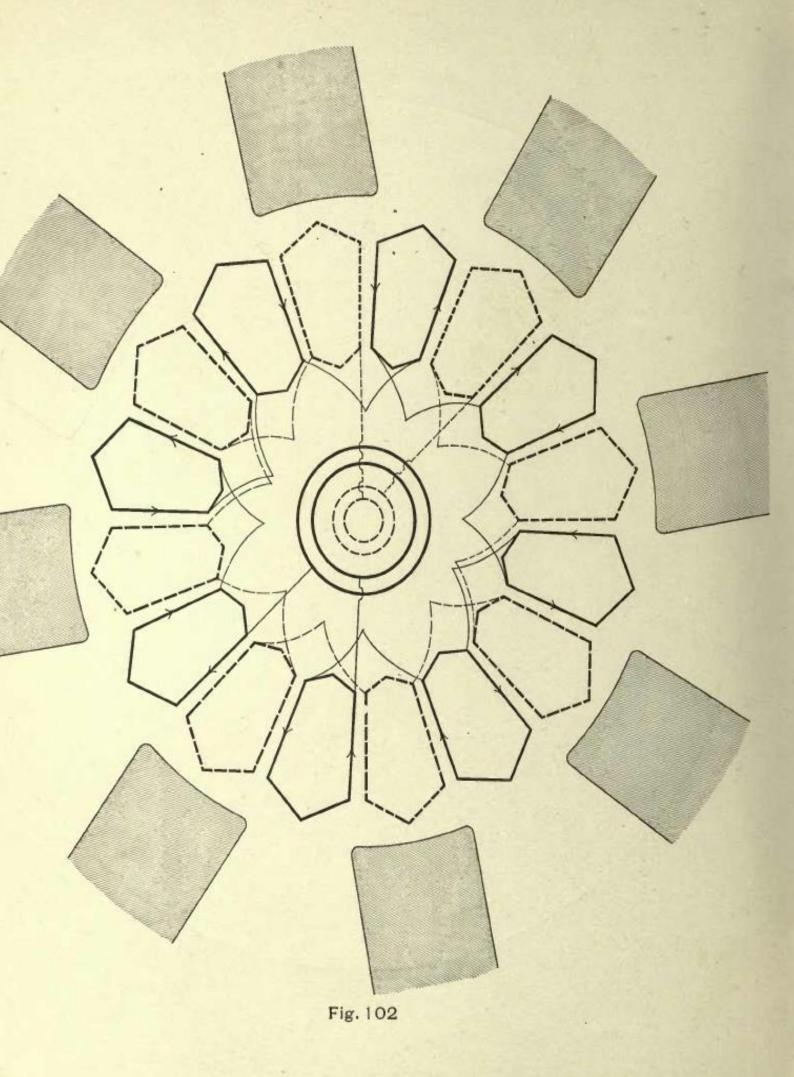


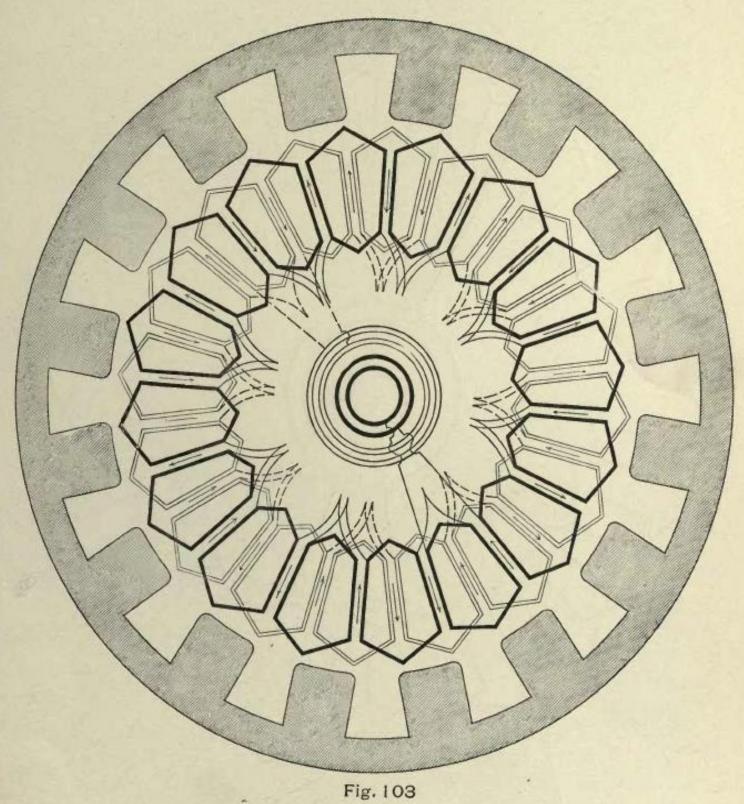
Figure 102 is a non-overlapping quarter-phase winding with one group of conductors per pole piece per phase. It has the advantage over Fig. 100 that there are no crossings at the ends of the armature, and that it utilizes the end space more completely, thus bringing about a better distribution of the necessary heating losses in the copper. Its chief fault is that if the width of the pole face is over one-half of the distance between pole centers, the coils never embrace the total flux from one pole piece. However, at full load, the area occupied by the flux is narrower, and a greater portion would be included than at no load, so that this objection would not be so serious as would appear at first sight. Moreover, the necessary space allowance for the field winding will in many cases not permit the width of the pole piece to be sufficiently great to cause any trouble in this respect. Mechanically, this is an excellent winding, being, in fact, the single-phase winding given in Fig. 77, for double the number of poles.

The remarks made in connection with Fig. 96 (singlephase alternating winding with one and one-half slots per pole piece) should also be considered in studying this winding. Consult also Fig. 119 and corresponding text.



Figure 103, which like Fig. 102 has two coils per group, is not open to the objection discussed on the preceding page. It has, however, crossings at the ends. It is to be preferred to Fig. 100 for the reason that the end space is more effectively utilized, but the additional crossings would require a somewhat greater length of wire than would be necessary in Fig. 100.

Bar windings could be built corresponding to the coil windings of Figs. 102 and 103. They would not be symmetrical at both ends, but might advantageously prove applicable for certain cases. The two bars of a group could be placed either over each other, or side by side. With smooth-core construction the latter arrangement would be adopted, and often also in ironclad armatures with bar windings.



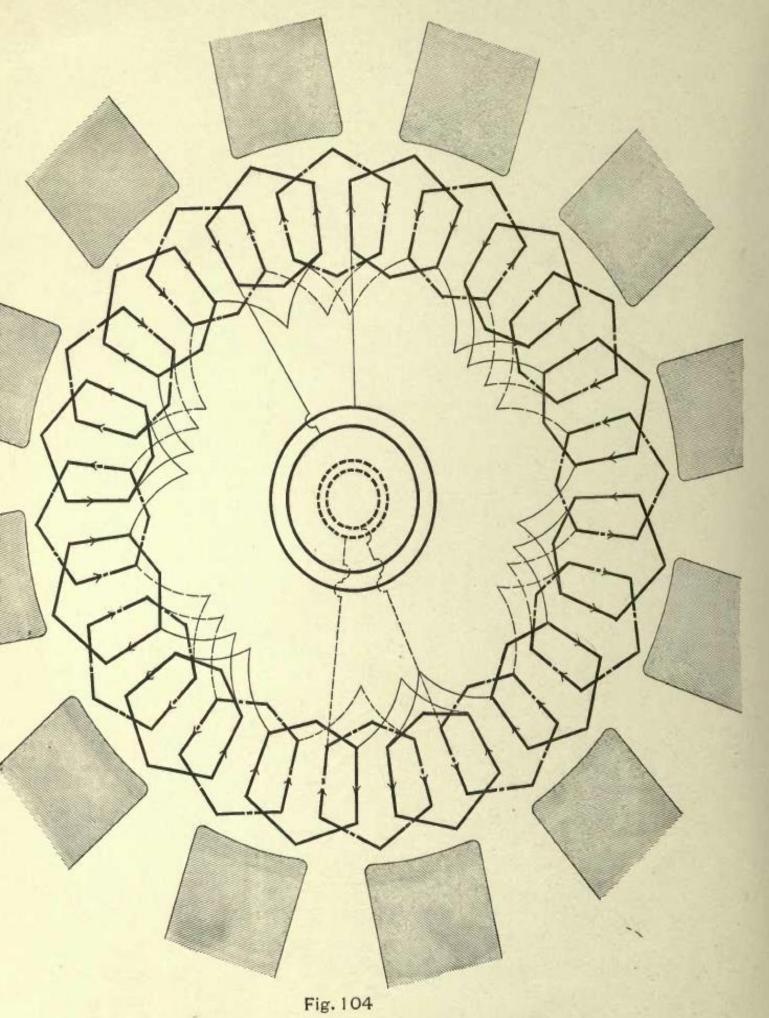


Figure 104 is a quarter-phase coil winding with two conductors per pole piece per phase. It is entirely symmetrical, and utilizes all the winding space to the best advantage. The crossings at the ends are unavoidable, but may be made thoroughly satisfactory from a mechanical standpoint by proceeding in the manner shown most clearly in the diagram of Fig. 123.

Such windings are applicable to quarter-phase armatures with any even number of coils per pole piece per phase.

In studying Fig. 104 it will be instructive to examine Fig. 97, which is one of the two windings of Fig. 104.



Figure 105 is electrically equivalent to Fig. 104. The winding might sometimes be used, although it would for most purposes be excelled by Fig. 104.

It will be noted that the end connections are longer, and that they occupy a greater depth. Much of the end space is wasted. This winding is superior to that of Fig. 104, in that the coils are so located as to make it very plain how the connections should run. This would be of great assistance to the winder, and would, moreover, facilitate the detection and correction of faults that might develop in practical working.

An armature with such a winding could be built and shipped in segments.

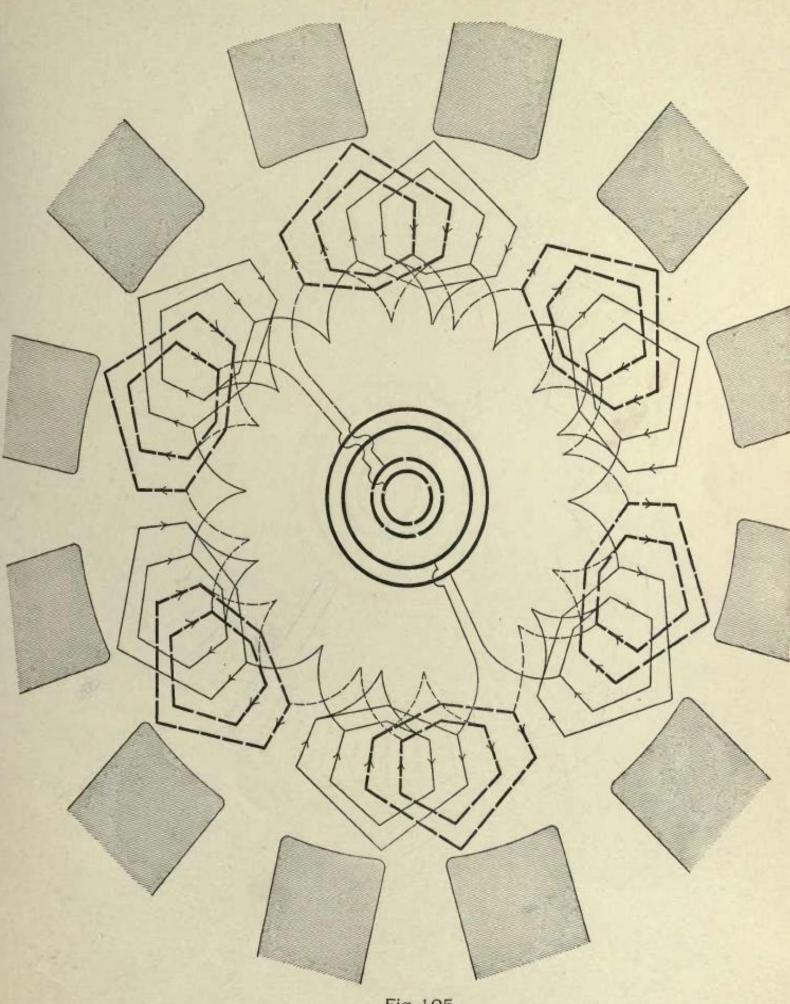


Fig. 105

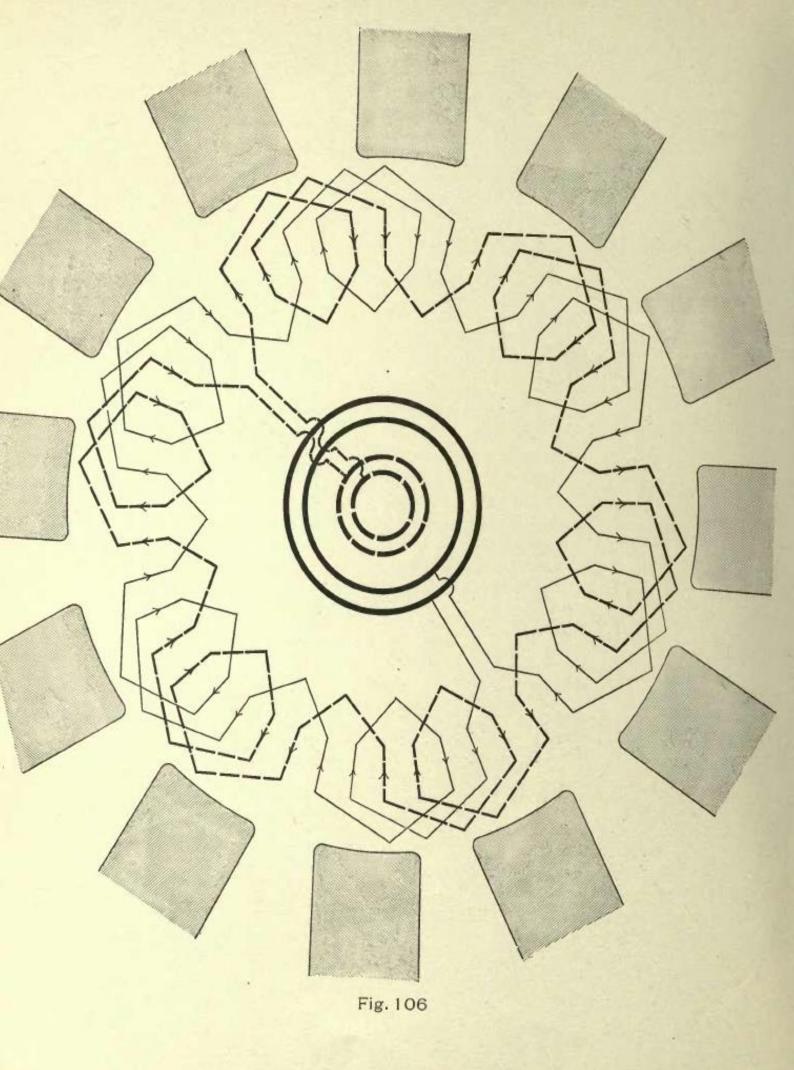


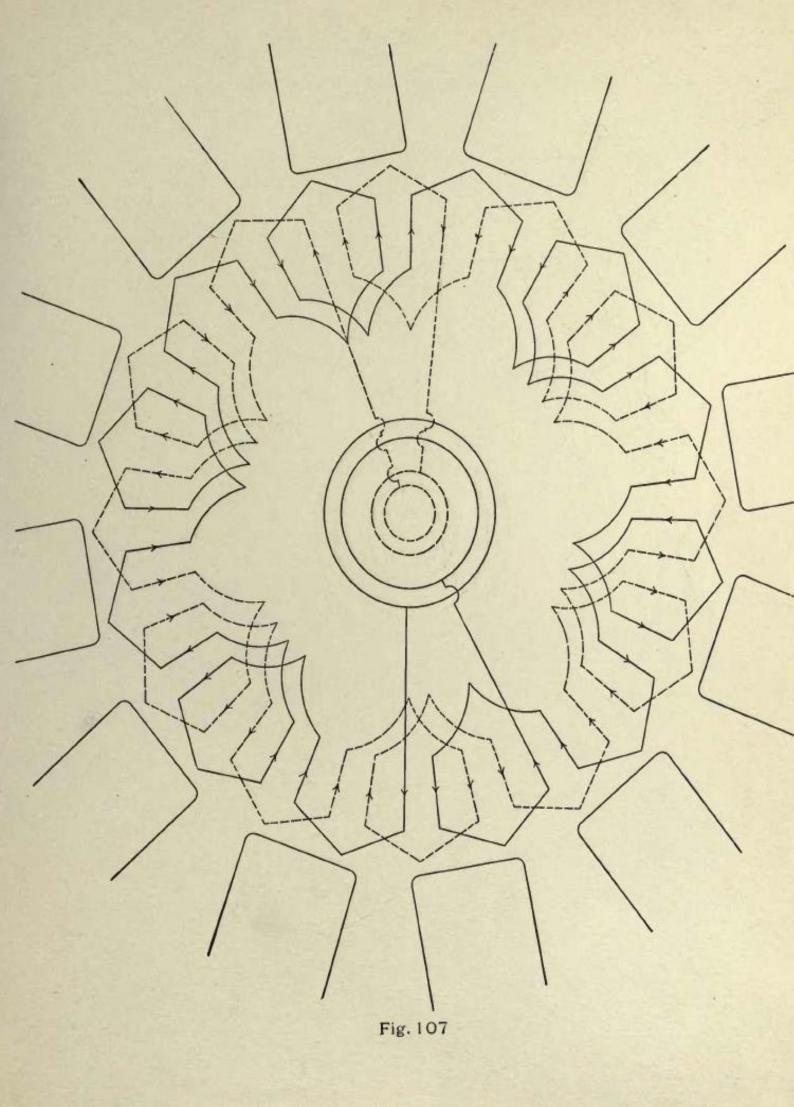
Figure 106 is a bar winding differing but little in principle from the coil winding of Fig. 105. The space is uniformly occupied at the collector ring end, but is not at the other end.

This lack of uniformity in end connections is not of very great moment in bar windings with few bars per pole piece. Other things being equal, however, it would on the whole seem best to avoid it, although in special cases such disposition of the end-connections allows room much needed for mechanical arrangements.



Figure 107 is a bar winding corresponding to Fig. 104. It is a good example of the fact that very symmetrical coil windings often correspond to very unsymmetrical bar windings, and vice versa. But, as noted on the preceding page, this lack of symmetry is in such cases not a great objection, and has, incidentally, some redeeming features.

One of the two windings of this diagram would, as mentioned on page 209, work out very well for a single-phase armature.



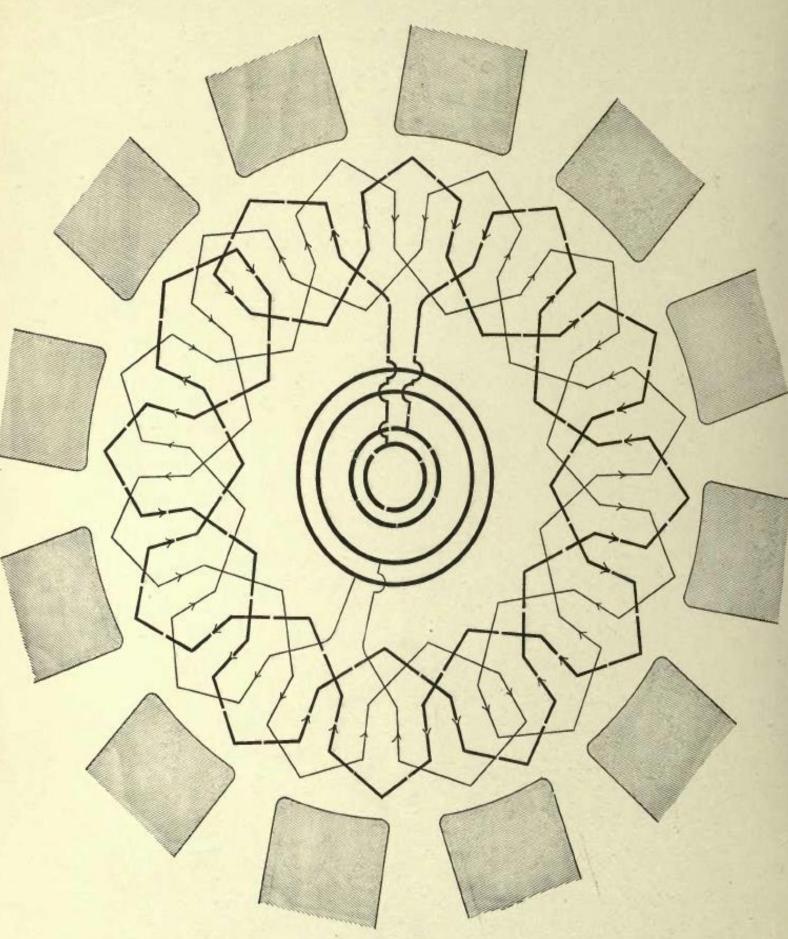


Fig. 108

Figure 108 is a much better bar winding than Fig. 107, though electrically equivalent.

It will be seen to be unsymmetrical at two points at the end distant from the collector This irregularity consists in the end connections of the two adjacent bars starting off in the same direction, instead of, as in all other parts of the winding except these two, going in opposite directions. Four of the end connections have to be longer than the rest.

This winding is practically the same as the following one, Fig. 109, except that the above described irregularity is introduced instead of making use of the cross-connections shown in Fig. 109.



Figure 109 is a symmetrical quarter-phase bar winding with two conductors per pole piece per phase. If used for an ironelad or projection armature, it may have four slots per pole piece with one conductor per slot, or two slots per pole piece with two conductors per slot.

Examination will show that it is essentially a twelve-pole armature with four separate series of windings of twelve bars each. These four windings are connected up into two windings of twenty-four conductors each.

At the front end y=5, and at the back end y=3, therefore average y=4.

As pointed out in the discussion of Fig. 101, Figs. 108 and 109 have the fault that neighboring conductors have between them large percentages of the total potential of the armature, and this would sometimes be objectionable in cases of high potential windings.

It will doubtless have been observed that in the case of quarter-phase windings, multi-coil construction does not have to so great an extent the fault pointed out in the case of corresponding single-phase windings, of useless counter-electromotive forces.

The coils of one phase usually embrace practically the entire flux, because the two groups of conductors, forming respectively the two sides of a coil, are usually separated by a group forming one side of a coil belonging to the winding of the other phase.

This advantage is possessed in a still greater degree by the three-phase windings, which will be discussed later.

Exceptions to the above statement often occur in cases where single and multi-phase alternating windings are obtained from ordinary directcurrent windings.

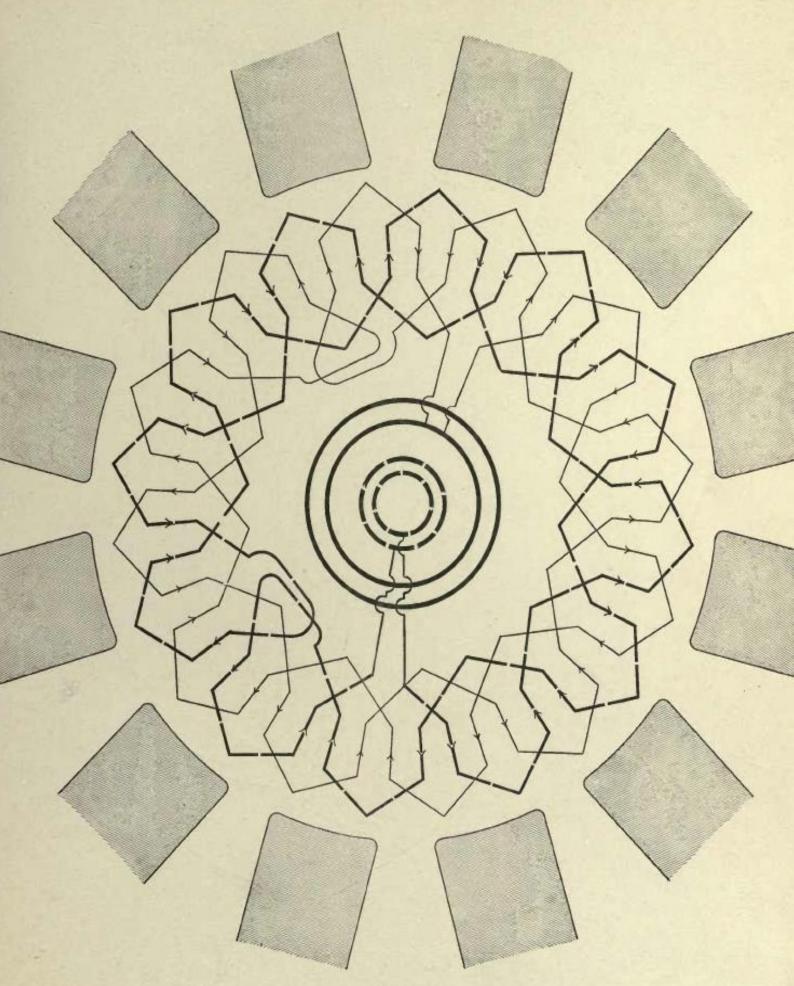


Fig. 109

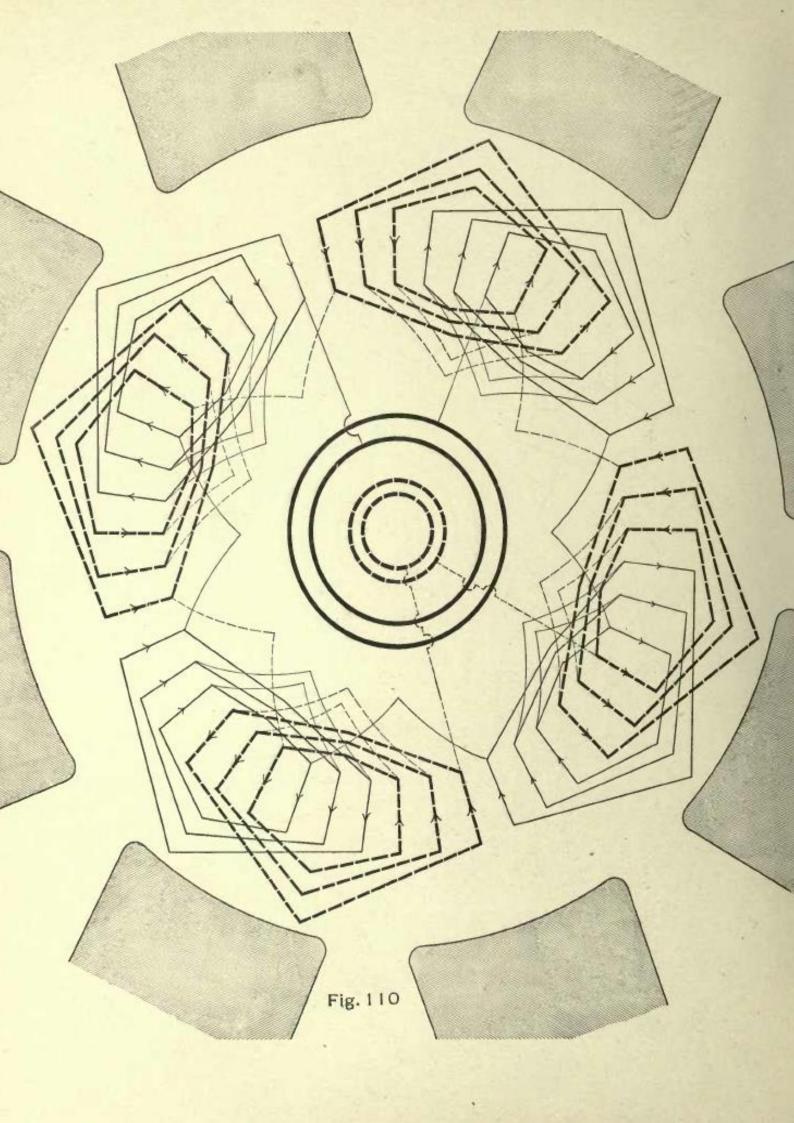
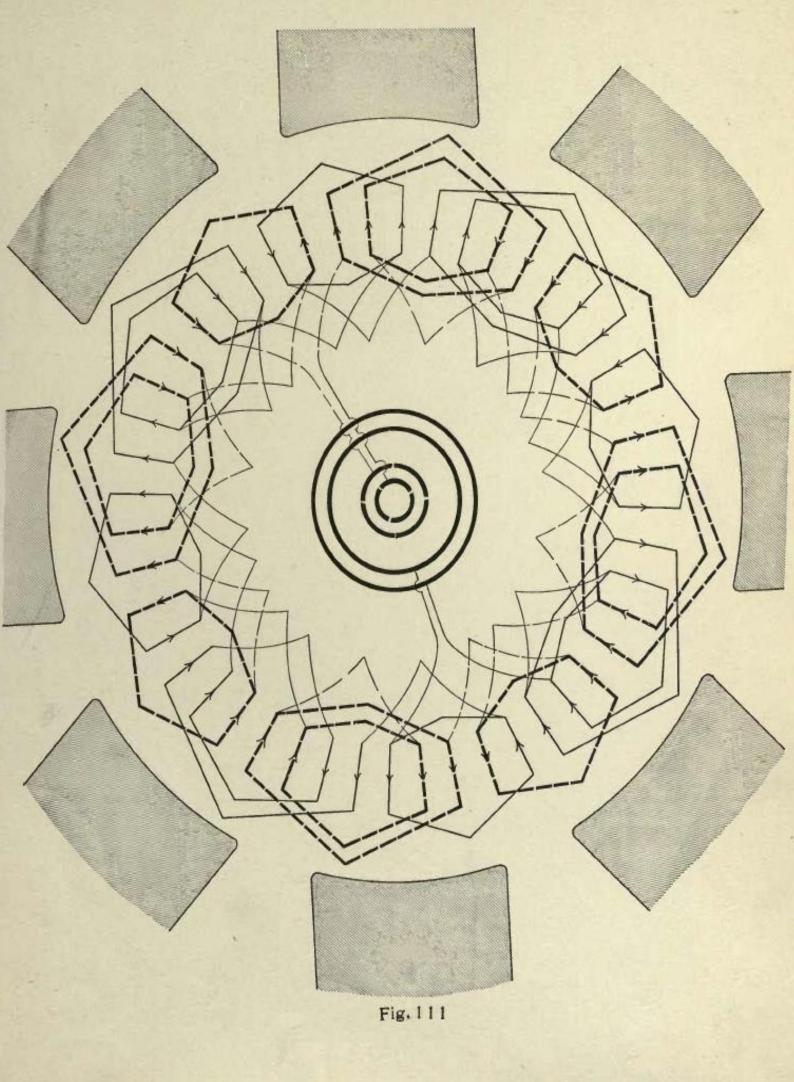


Figure 110 represents a quarter-phase coil winding with three slots per pole piece per phase. It does not utilize very uniformly the end space on the armature, the end connections being three layers deep at some points and much less at others.

An advantage of this winding is the well-defined nature of the coils, rendering it easy to see just how they should be connected. The winding might also be necessary, if it should be required that the armature should be built so that it could be shipped in segments.



Figure 111 is electrically equivalent to Fig. 110, but the end connections are only two layers deep, are shorter, and are better distributed over the ends of the armature. Where the number of coils per pole piece per phase must be odd, windings such as those given in Figs. 110 and 111 must for quarter-phase armatures often be chosen. It is quite apparent that, except in special cases, the style of diagram shown in Fig. 111 will give the best result.



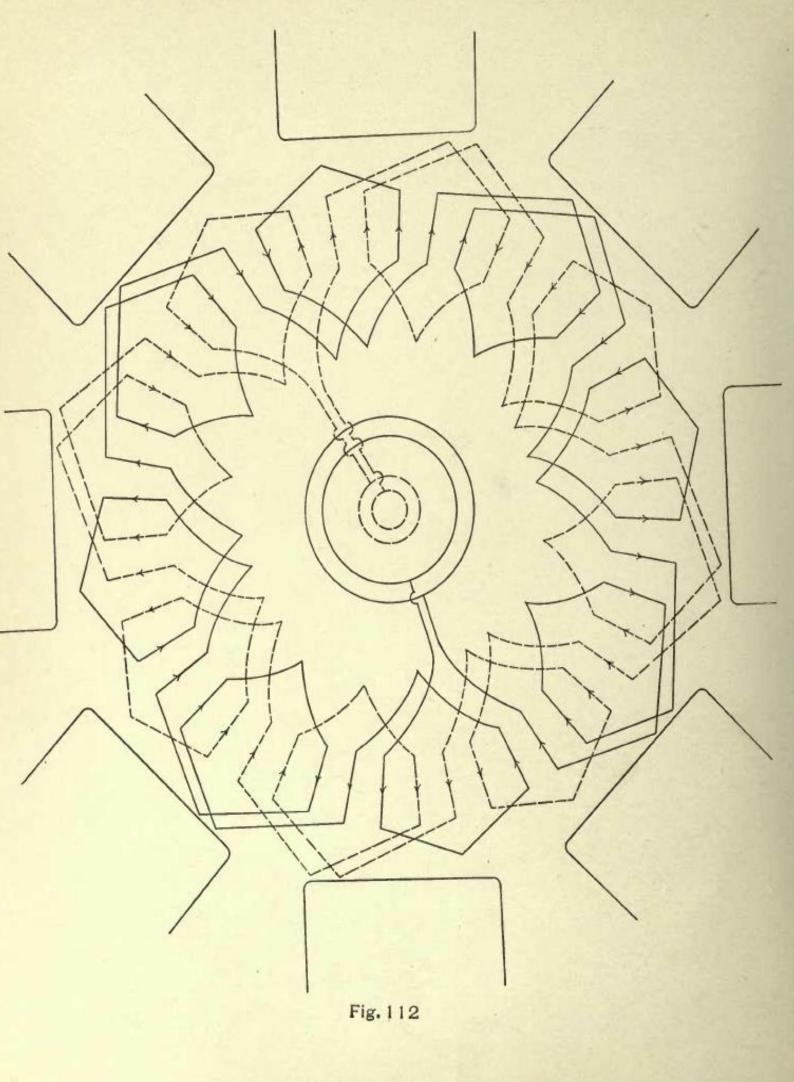


Figure 112 is a bar winding corresponding to the coil winding of Fig. 111. Although not symmetrical, the end connections are fairly well distributed, and there would be in but very few places any great percentage of the total difference of potential between adjacent conductors. Several different lengths of end connections would necessarily have to be employed.

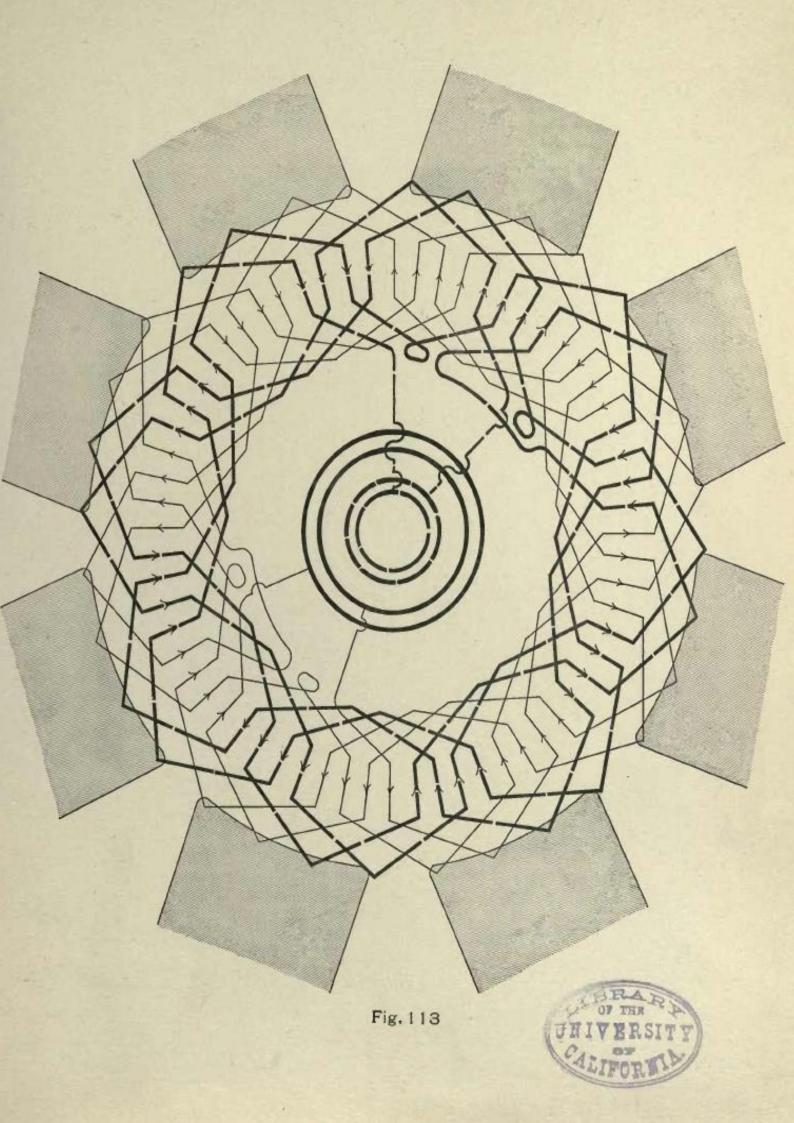
One of the two windings of this diagram has already been given in Fig. 98 in Chapter XIII. on Single-Phase Windings.



Figure 113 represents a quarter-phase bar winding with four conductors per pole piece per phase. It is perfectly symmetrical, and may have one, two, or four conductors per slot, as desired.

This winding is like that of Fig. 109, except that four sets of elementary windings are connected in series to form one of the two phases, instead of two sets, as was the case in Fig. 109.

If one-half or one-quarter as great a terminal electromotive force should be desired, two, or all four, of these elementary windings could be connected in parallel between the collector rings, instead of joining them in series as shown.



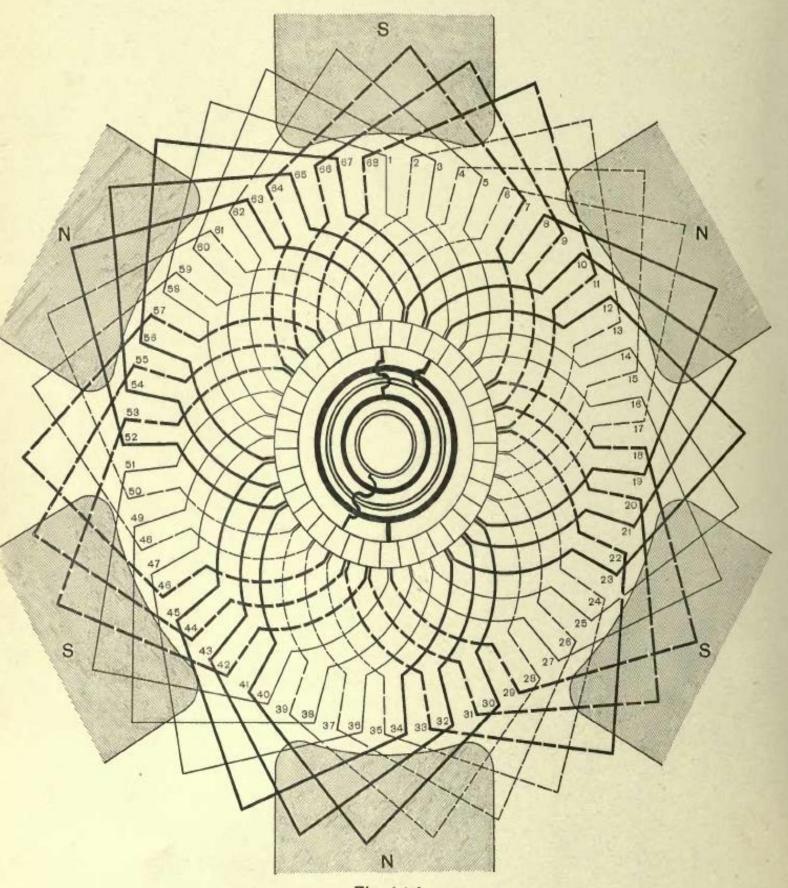


Fig. 114

## TWO-CIRCUIT WINDING FOR QUARTER-PHASE CONTINU-OUS CURRENT COMMUTATING MACHINE.

Figure 114 is the diagram for the winding for a commutating machine for deriving a continuous current from a quarter-phase alternating supply, or vice versa, or for a generator for supplying both continuous and quarter-phase systems.

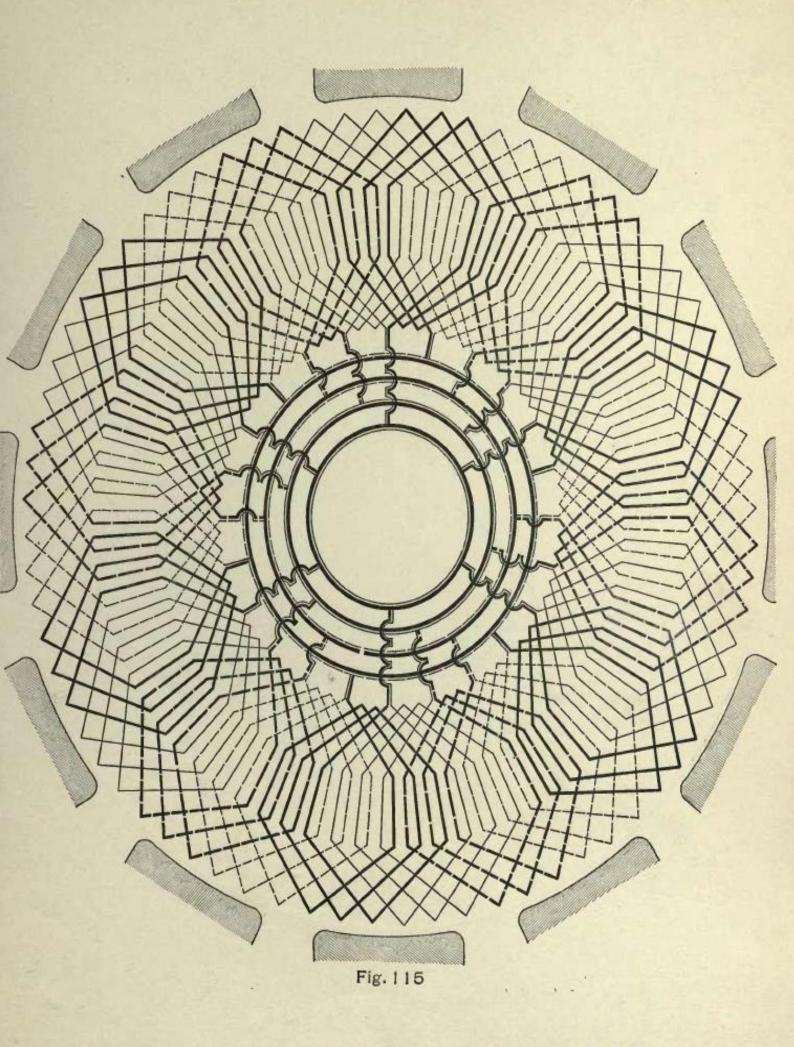
Examination will show that it is the two-circuit single winding of Fig. 43 (Chap. VIII.), tapped off from four approximately equidistant points to four collector rings. As the winding consists of sixty-eight conductors, there should be seventeen conductors in each section, but for the convenience of having all the connections to the collector rings made at one end, the divisions are 16, 16, 18, and 18. With the large numbers of conductors used in practice, the irregularity produced by one conductor more or less would be of less importance, though always undesirable. In such a winding four points only of the armature are tapped independently of the number of poles.



TWELVE-CIRCUIT WINDING FOR QUARTER-PHASE CON-TINUOUS-CURRENT COMMUTATING MACHINE.

Figure 115 is another winding for a quarter-phase continuous-current commutating machine. It is fundamentally a multiple-circuit, continuous-current winding, and requires four leads (one to each collector ring) for each pair of poles.

It is to be remembered that in quarter-phase continuouscurrent commutating machines, the effective voltage between collector rings 180° apart equals the continuous-current voltage multiplied by .707 (or divided by 1.414).



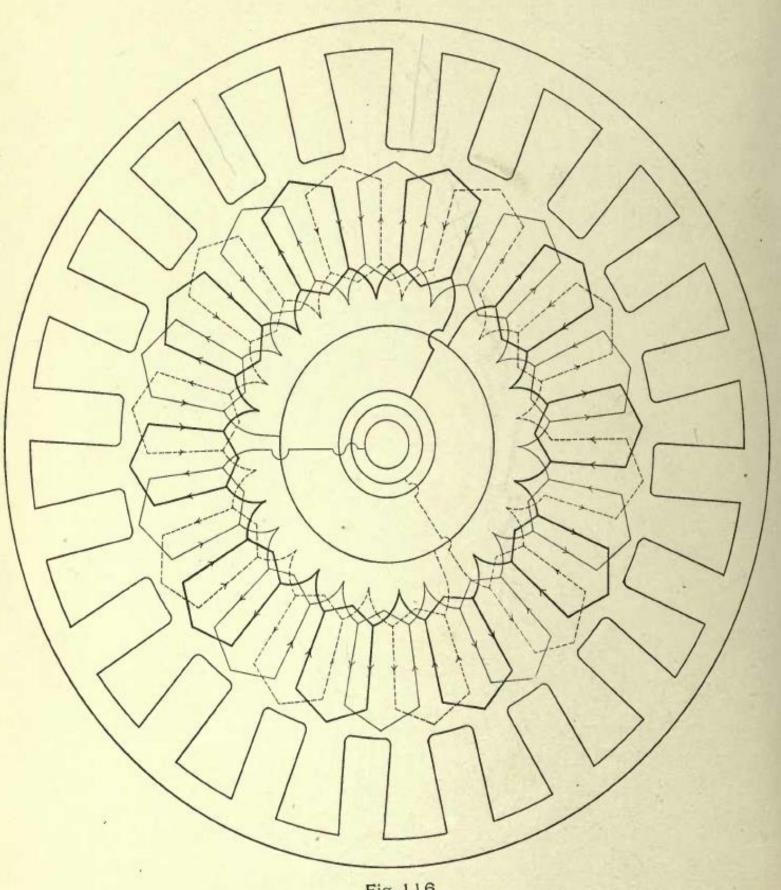


Fig. 116

## CHAPTER XV.

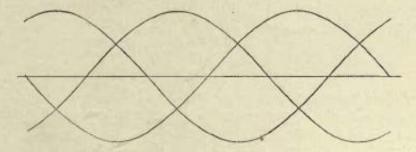
## THREE-PHASE WINDINGS.

FIGURE 116 is a three-phase coil winding with one set of conductors per pole piece per phase. The coils belonging to the three windings may be distinguished from each other by the three different styles of lines. The armature is connected in a manner technically known as the "Y" connection. The characteristic of this style of connecting three-phase windings is that one end of each of the three windings is brought to a common connection, the other three ends being carried to three collector rings.

Inasmuch as three-phase alternators have but recently been used to any considerable extent in practice, it may not be out of place to give as concisely as possible a few of the leading considerations involved in their practical construction and operation, as far as relates to the armature windings.

One complete cycle is passed through by any armature conductor while passing from a certain point opposite one pole piece, say the middle of the north pole, to the corresponding point opposite the next pole piece of the same polarity. This angular distance is usually spoken of as 360°, independently of the number of poles of the machine. Now, a three-phase armature winding is merely three single-phase windings, laid on the same armature, the conductors of the three windings, however, being located 120° (one-third of a cycle) behind each other. Any conductor of one winding is, therefore, at any instant, in a different phase from that of the conductors of the other windings. Thus, in the position represented in Fig. 116, the conductors represented by heavy lines are directly opposite the middle of the pole pieces, the light line conductors are located 120° behind them, and the dotted conductors are 120° behind the light conductors and 240° behind the heavy conductors.

Now it follows from the relative positions of the conductors of the three phases, that the electromotive forces generated in the three windings are 120° behind each other, and if they are sine waves, they may be represented, as in the following figure, by three sine curves displaced 120° behind each other.



If the three circuits are equally loaded, these curves may also be considered to represent the corresponding instantaneous values of the current.



It will be noted that at every instant, the algebraic sum of the three currents is zero. Now instead of having three pairs of lines and brushes and collector rings, one end of each of the three windings is brought to a common connection, and a conductor from this common connection could be used as a common return for each of the three circuits. But, since the resultant current at every instant is zero, this conductor becomes superfluous and is omitted.

If the voltage between any ring and the common connection, that is, the voltage per phase, is equal to v, then the volts V between any pair of collector rings will be,—

$$V = \sqrt{3} v \text{ or } 1.732 v.$$

The effective current will be equal in each of the three lines, and may be represented by C. With a non-inductive load, the watts output, W, will be,—

$$W=3 Cv = \frac{3 CV}{\sqrt{3}} = 1.732 CV.$$

If the load is inductive, the current C, for a given output W, will be greater than with a non-inductive load.

A safe and easily understood way of connecting the three windings correctly to the three collector rings and the common connection, is to consider that the winding whose conductors occupy the position in the middle of the pole piece, is carrying the maximum current, and to indicate its direction on the winding diagram by an arrow. The currents at the same instant in the conductors immediately next to it on the right and left are in the same direction, and should be so marked by arrow-heads. Now, from the sine curves given above, it will be seen that where one curve has a maximum value, the other two have a value half as great, and in the opposite direction. Therefore consider that the current in the winding occupying the position at the middle of the pole face is flowing away from the common connection. Then the currents in the other two windings, which are each of half the magnitude of the former, must both be flowing into the common connection; therefore join those ends of the three windings to the common connection, which will bring about this condition at this instant. Carry the other three ends to the three rings. This has been done in the upper diagram of Fig. 117, which represents a "Y" connected three-phase winding.

Another way of connecting up three-phase armatures is to connect the three windings in series in a closed circuit, and at every third of the total way through the circuit thus formed, to carry off a lead to one of the collector rings.

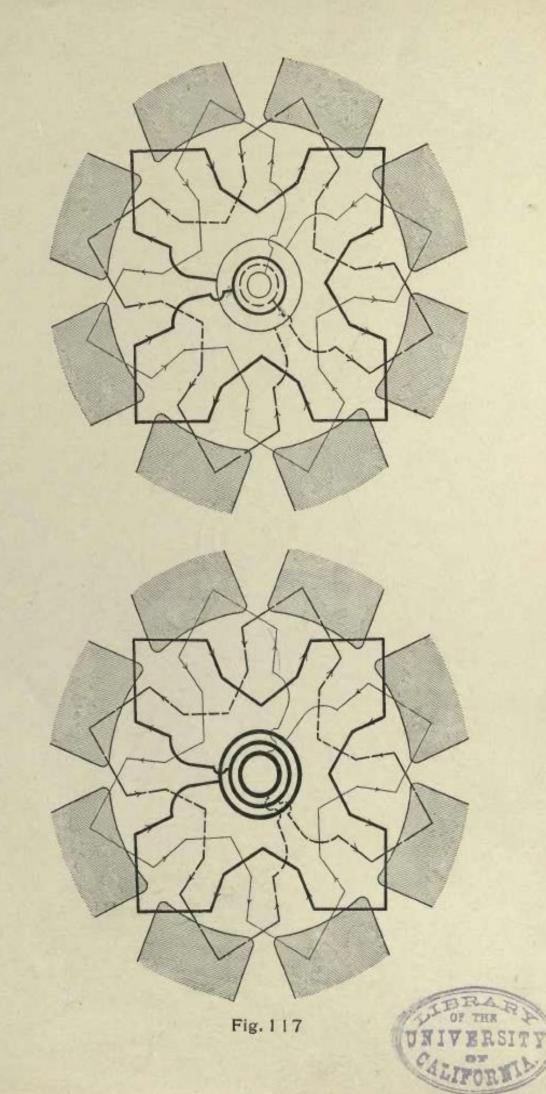
In the case of this, technically called the "delta" ( $\Delta$ ) connection, the current C in the line (i.e. beyond the collector rings) is  $C = \sqrt{3}c$ , or C = 1.732c, where c = current in the winding. The volts per winding are in this case equal to the volts between each pair of collector rings; that is, to the volts per phase. The watts output of a machine are,—

$$W=3 cV = \frac{3 CV}{\sqrt{3}} = 1.732 CV.$$

Examples of each of these two connections are given in Fig. 117.

The upper diagram represents a "Y" connected three-phase armature, and the lower diagram represents the very same armature, but with a "delta" ( $\Delta$ ) connection.

In connecting up the separate windings for a "delta" ( $\Delta$ ) connection, it is most convenient to choose the instant when the conductors of one phase are opposite the middle of a pole piece. Then assume these conductors to be carrying the maximum current, which is illustrated in the figure by the larger arrow-head.



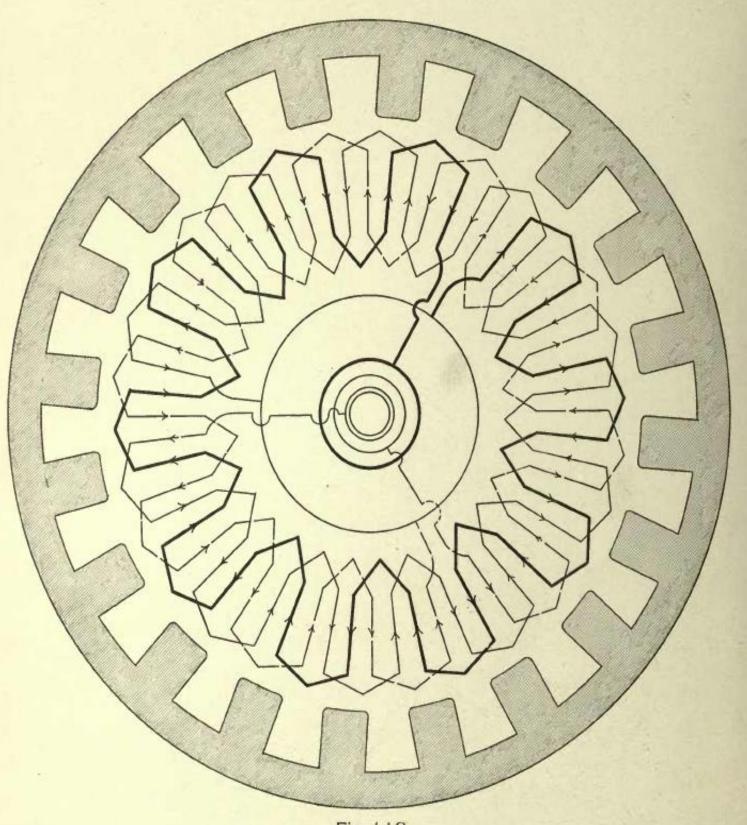
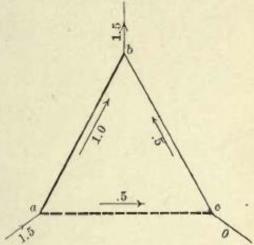


Fig. 118

The other two windings are at the same instant having induced in them currents of only one-half this magnitude. The condition of affairs in line and in winding is, for the instant, as represented in the following diagram.

From this it is seen, that, starting from the middle collector ring (corresponding to point a in the diagram), and following the direction of the current, we must pass through the heavy winding, carrying the large current to the outer ring (corresponding to point b of diagram). In the other direction, we must pass from the middle ring (i.e. point a), through the dotted winding, which carries one-half as great a current, to the inner collector ring (corresponding to point a of diagram). Then we must continue through the light winding, still in the direction of the current, until we again reach the outer collector ring, or point a of diagram.

Any of the following three-phase diagrams may be connected either "delta" or "Y," but they will usually be shown with the "Y" connection.



It is well to keep in mind that if a "Y" connected armature is changed over to the "delta" connection, it may with the same regulation and heating give 1.732 times as much current, but only  $\frac{1}{1.732}$  times the voltage. The reverse holds true in changing from "delta" ( $\Delta$ ) to "Y."

Figure 118 is the bar winding corresponding to Fig. 116. It has one bar per pole piece per phase. This winding, while partaking of all the advantages and disadvantages of multi-coil construction, would be particularly unsatisfactory for a three-phase motor on account of the dead points that it would develop at starting. These dead points are much less marked with multi-coil windings and with windings like those in Figs. 119 and 120.

In the case of induction motors, it is customary to make use of such windings as those given in Figs. 126 and 127, where smoother action is obtained partly by virtue of the choice of a number of conductors, prime, or nearly so, to the number of poles.



Figure 119 is a non-overlapping, three-phase, coil winding, with only one and one-half coils per pole piece per phase. It is the winding which was given with its single-phase connection, in Fig. 96. This should make a very excellent three-phase winding, as there is no crossing of the coils. It is a regular thirty-pole, single-phase winding, connected up as a three-phase armature for twenty poles. This diagram should be compared with Fig. 77, Fig. 96, and Fig. 102. It should be particularly suitable for use in three-phase motor work, as it should have very weakly defined dead points. In a projection armature, when a slot is opposite a certain pole piece, spaces between two slots will be opposite the adjacent pole pieces, thus giving a more equitable distribution of the magnetic flux.

The inductance of such a winding is low and fairly uniform, for the reason that when one side of a coil occupies a position under a pole piece, the other side of the coil is between two pole pieces.

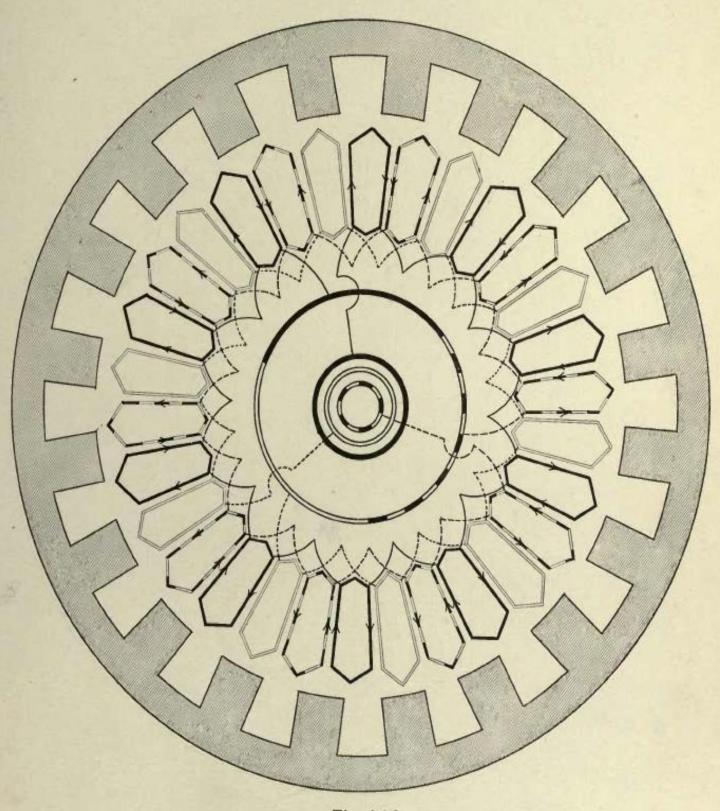


Fig. 119

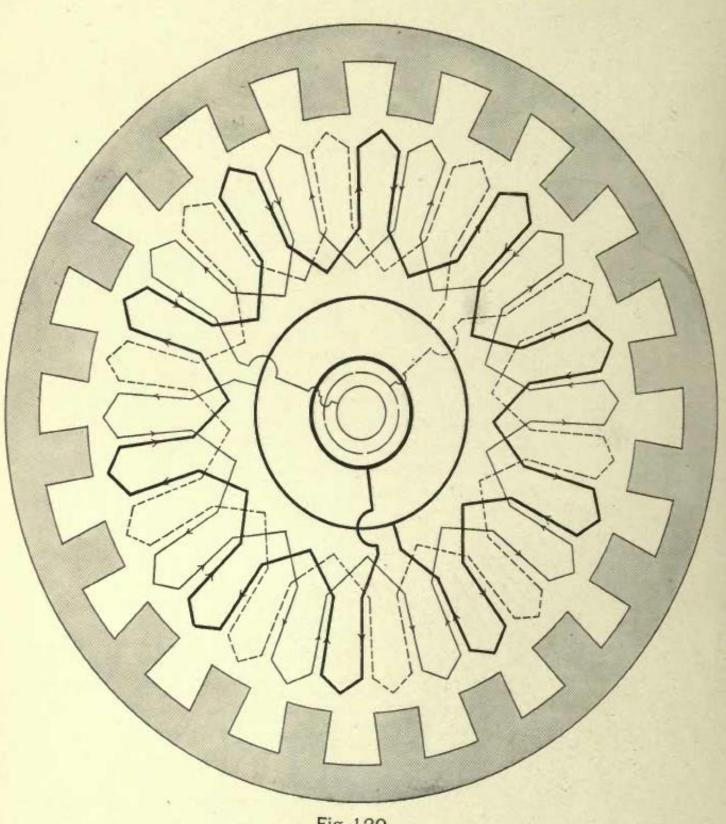
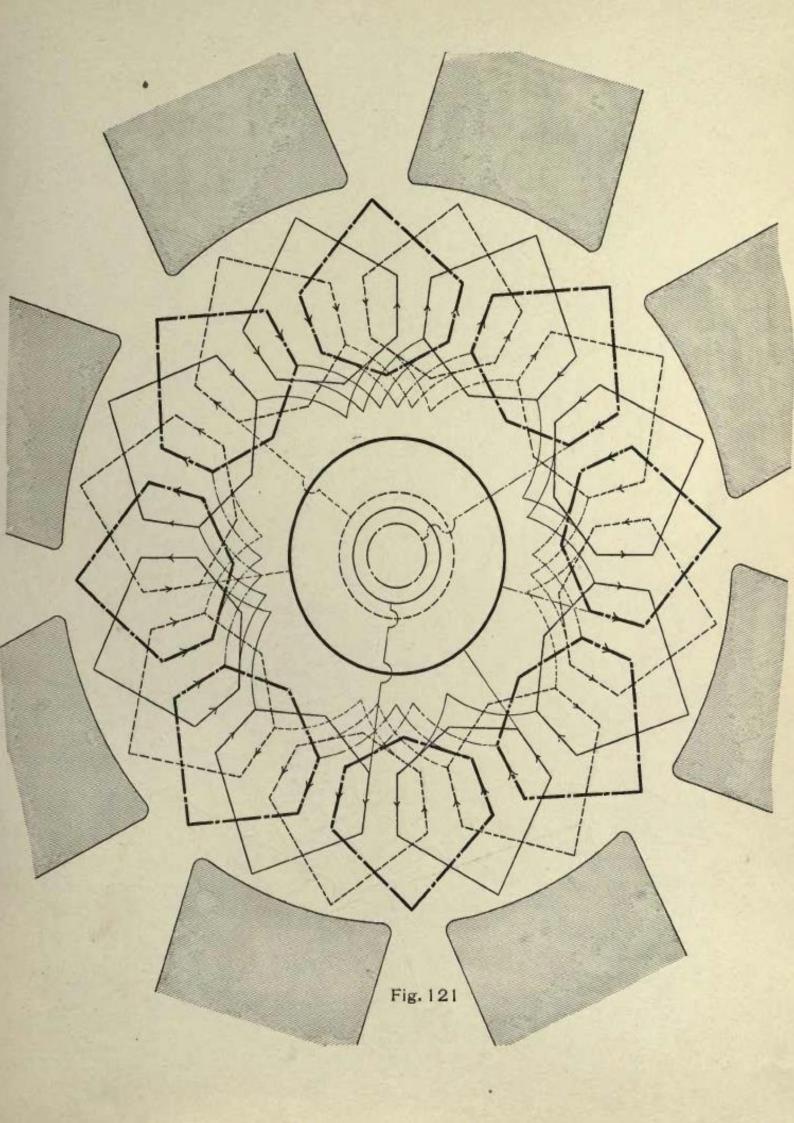


Fig. 120

Figure 120 represents the corresponding bar winding. In the case of projection or ironclad armatures, it would have two bars per slot, which might be arranged one over the other or side by side. It is interesting to note that each slot would contain one bar of each of two windings, two bars of the same winding never occupying the same slot.

All the remarks regarding the winding of Fig. 119 apply equally well to Fig. 120.





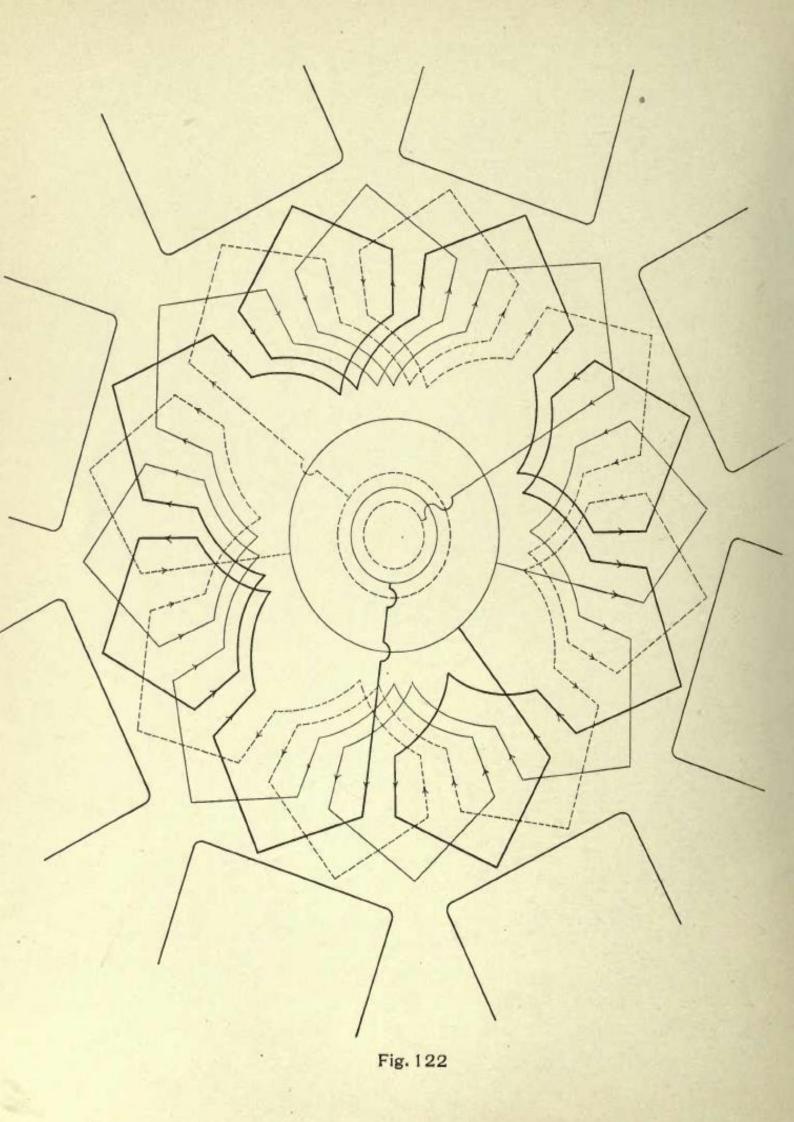


Figure 122 is the bar winding corresponding to Fig. 121. The end connections are perfectly symmetrical and well distributed at one end, but are far from it at the other. Its point of superiority over Fig. 124 is that it has, as a rule, no great differences of potential between adjacent conductors.

As already stated, the irregular distribution of the end conductors is not, at least in the case of bar windings, so great an objection in cases where there are comparatively few bars per pole piece. And in this instance there is a sort of a regularity about their grouping, that might be found of advantage on account of the large spaces that it makes available for mechanical arrangements.



Figure 123, which was devised by Mr. Thorburn Reid, who has devised a number of useful windings, is superior in the mechanical arrangement of the coils, to the winding of Fig. 121. The corresponding bar winding is not drawn, but it may be readily seen that it would have no very obvious advantages.

Coil windings of the same style as that of Fig. 123 may be constructed with any number of coils per pole piece per phase, and are frequently superior to other arrangements.

It is thought that the style of lining adopted in the diagram will indicate fairly well the arrangement of the end connections, if care is taken to note that the conductors of some groups of coils are carried directly over in the same plane as the face wires, to the conductors forming the other side of the group. The end conductors of the other coils have to be bent down out of the plane of the face conductors and then back again into their plane. The coils are usually wound in forms and then laid in place on the armature.

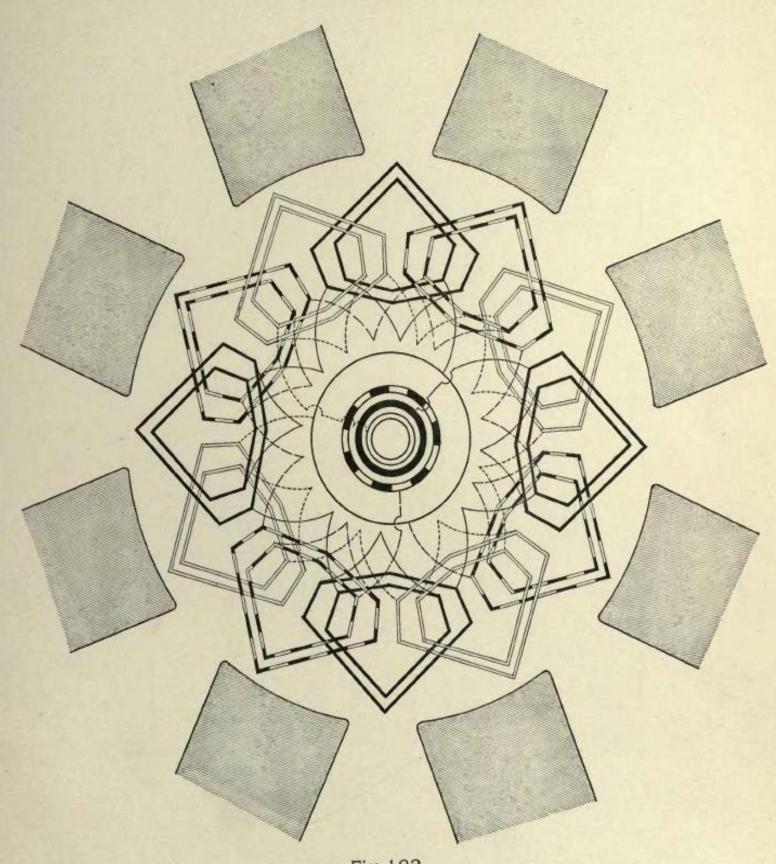


Fig. 123

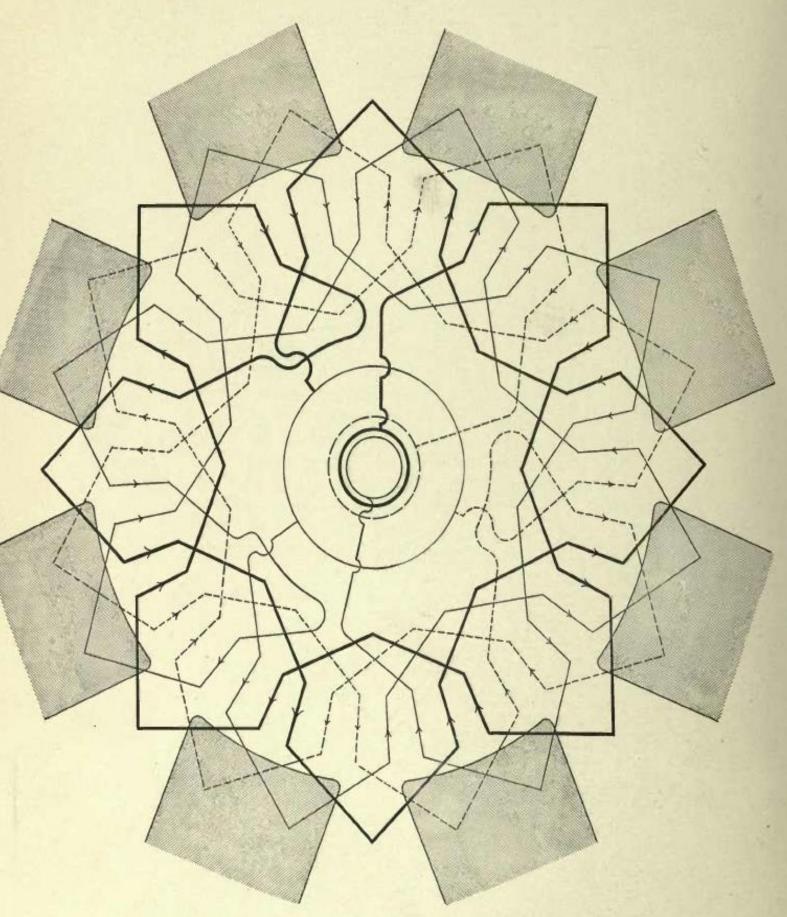


Fig. 124

Figure 124 is a three-phase bar winding, with two bars per pole piece per phase. It is perfectly symmetrical, and may have either one or two conductors per group. It is inferior to Fig. 122, in that, from the nature of the winding, there are much greater differences of potential between adjacent conductors than in Fig. 122.

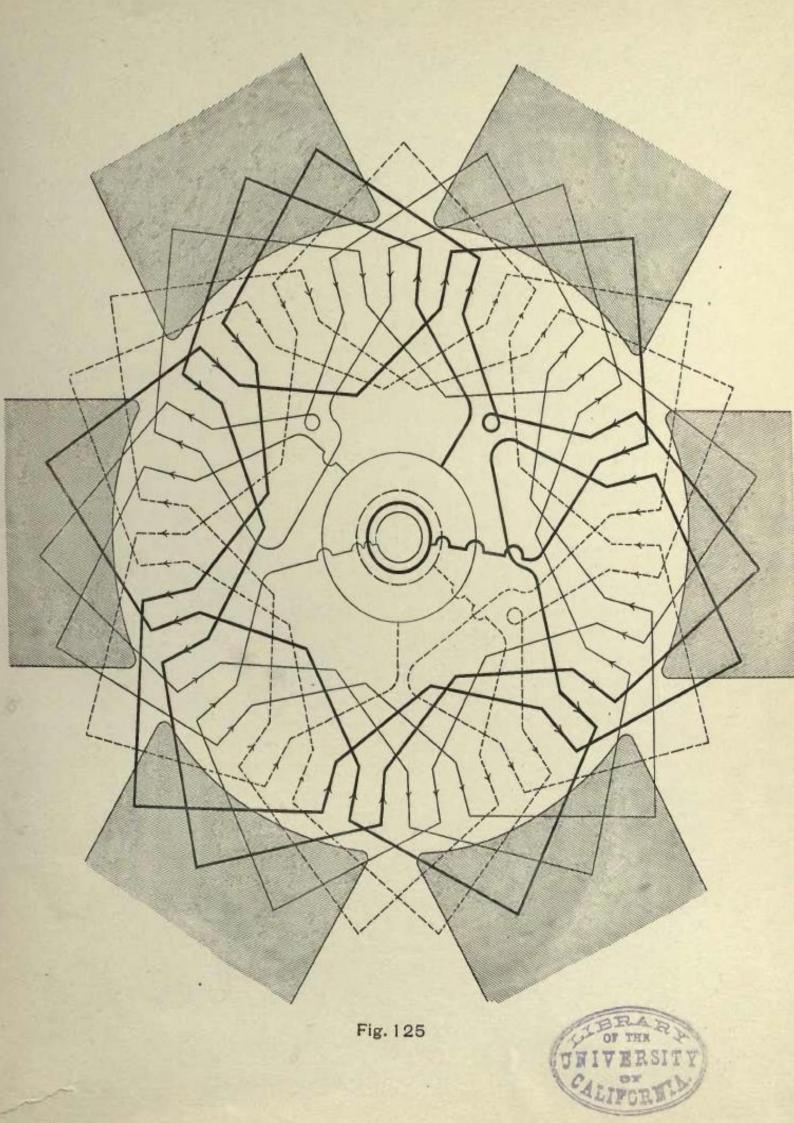
In Fig. 124, the pitch is 5 at one end and 7 at the other. Two sets of conductors, each set having as many conductors as there are pole pieces, are joined in series to form each one of the three windings. If an armature for half the voltage had been wished, the two sets of conductors forming each winding would have been connected in parallel.

This winding, as well as the next (Fig. 125), is of the same general character as those shown in Figs. 109 and 113.



Figure 125 is similar in all respects to Fig. 124, except that it has three conductors per pole piece per phase. The pitch is 9 at both ends. It could be connected so as to give one-third as great a terminal electromotive force by joining the three elementary groups of which each winding is formed, in parallel, instead of in series.

In connection with Figs. 124 and 125, emphasis should be laid on the fact that in virtue of the nature of these windings, whereby adjacent conductors have between them large differences of potential, valuable space has to be sacrificed to make room for the proper thickness of insulation, which, with types of winding not possessing this character, could be usefully employed.



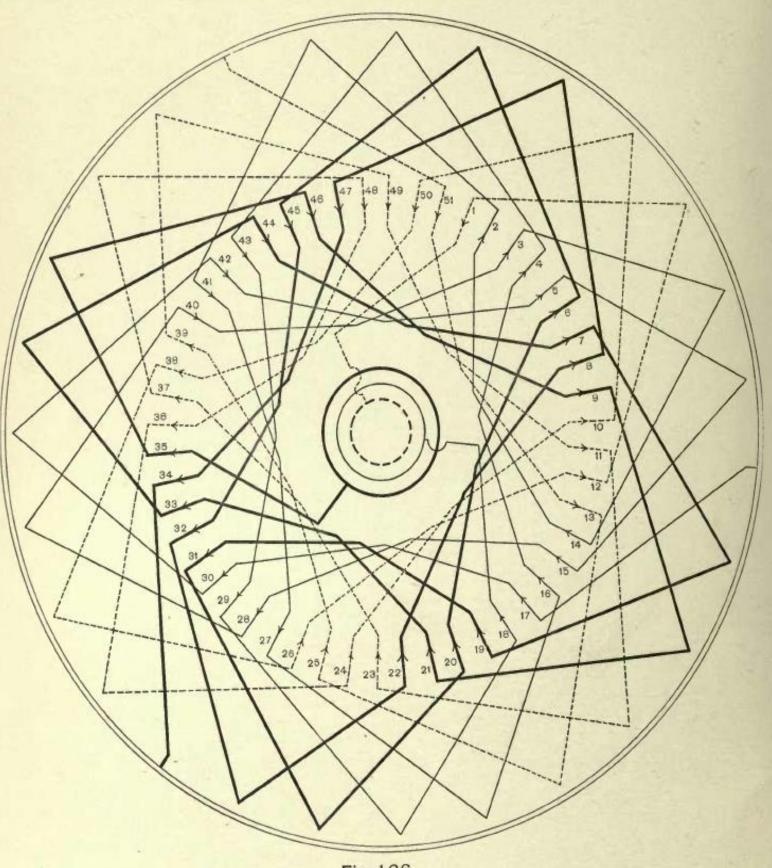


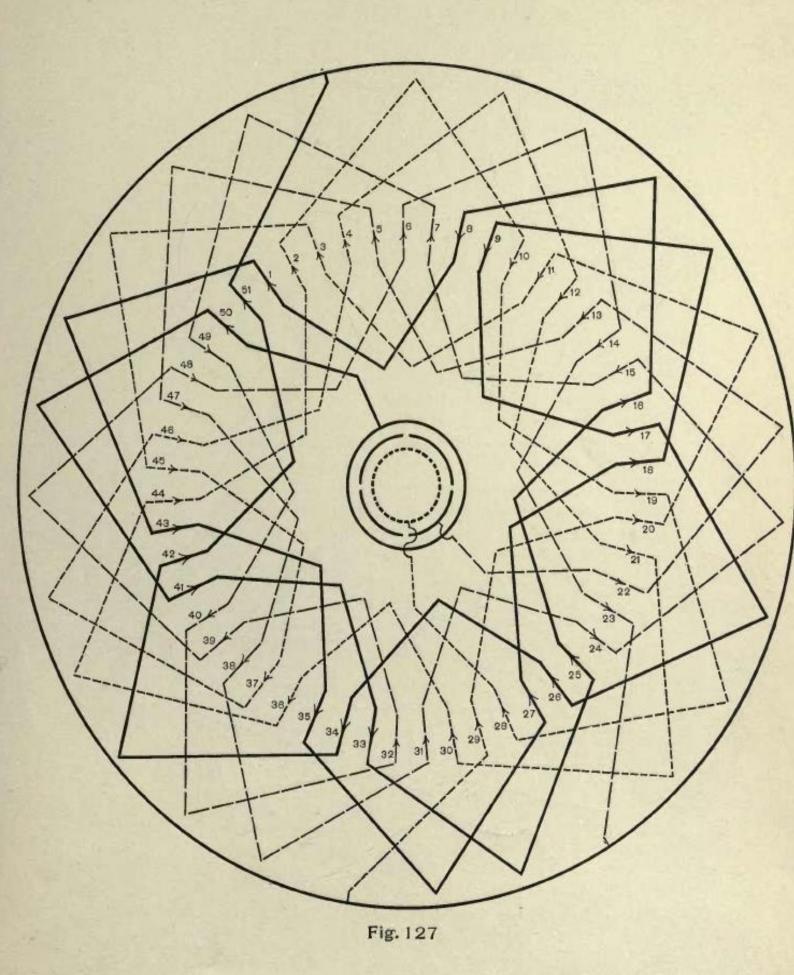
Fig. 126

Figure 126 is a four-pole, three-phase bar winding of a very irregular character. It has fifty-one conductors, seventeen per phase. There are, therefore, unequal numbers of conductors, both per phase and per pole, opposite the different pole pieces.

This style of winding has been used with success in induction motors, where it is important to choose a number of slots on the armature, which is prime, or nearly so, to the number of slots on the field. It may be well to state that, in the case of induction motors, the field, in the most successful types, consists merely of an assembly of annular punchings with radial slots within which the cylindrical drum armature revolves. It is practically a transformer, one of the elements, usually the secondary, being movable. It has become customary to call the moving element, the armature, and the stationary, the field. In the types, and for the voltages generally employed, it has been found best to use a coil winding for the field, the coils often being wound on forms and slipped into the slots. In the armature, which is practically a short-circuited secondary, the number of conductors and slots is determined by the permissible inductance, the actual voltage of the armature being to a great extent immaterial. In certain types the ratio of field to armature conductors has been something like 6:1. It is in connection with such motors as these, that the winding diagram of Fig. 126 will be found of greatest service. There cannot well be more than one bar per slot, because of the irregularity of the end connections.



Figure 127 is another three-phase bar winding with fiftyone conductors. It has six poles, and is even more irregular
than the winding of Fig. 126. It, like Fig. 126, will find
its chief use in the design of induction apparatus. Windings, almost as irregular, might be used in large polyphase
generators, where it is desired to have but one conductor
per slot.



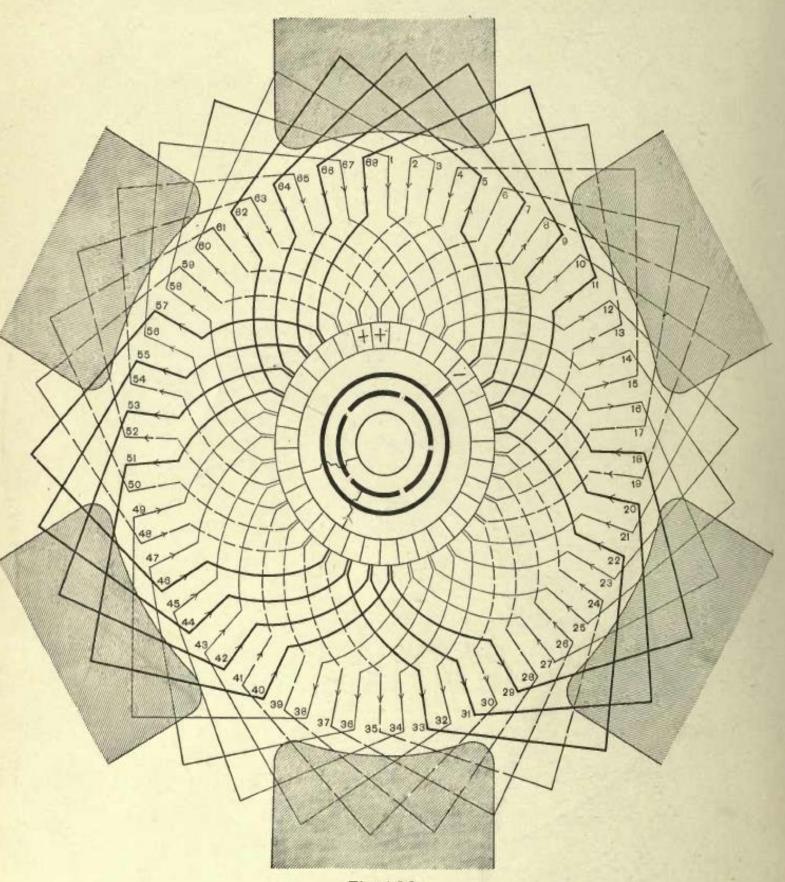


Fig. 128

## TWO-CIRCUIT WINDING FOR THREE-PHASE CONTINUOUS-CURRENT, COMMUTATING MACHINE.

Figure 128 represents the same winding as Fig. 114, except that here it is tapped off at three nearly equidistant points instead of at four, as was the case in Fig. 114.

The result is a winding for a three-phase, continuouscurrent, commutating machine.

The total sixty-eight bars are divided up into sets of twenty-two, twenty-two, and twenty-four conductors, respectively, which are represented on the diagram by heavy, light, and dotted lines.

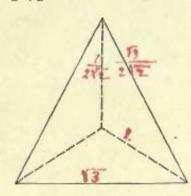
If the conductors are arranged in groups of two each, as would frequently be the case in projection armatures, where two conductors would often be placed together in each slot, it is of interest to note that these two conductors never belong to the same phase.

## SIX-CIRCUIT WINDING FOR THREE-PHASE, CONTINUOUS-CURRENT, COMMUTATING MACHINE.

Figure 129 is still another three-phase, continuous-current, commutating machine, but with a six-circuit winding. It requires three leads per pair of poles; therefore, in this case, nine leads. It is quite analogous to the quarter-phase, continuous-current, commutating machine of Fig. 115.

It is of interest to notice the relation of the voltage between collector rings to the continuous-current voltage at the commutator, in the case of three-phase, continuous-current, commutating machines. It will have been observed that they have "delta" connected windings.

Let V= continuous-current voltage at the commutator; then, taking the point of zero potential to be at the middle of the winding, the electromotive force of each half of the winding is  $\frac{V}{2}$ . But the corresponding effective alternating electromotive force will be  $\frac{V}{2\sqrt{2}}$ . This, therefore, will correspond to the voltage between common connection



(point of zero potential), and collector ring, for an equivalent "Y" connected three-phase armature winding. Now the voltage between the collector rings of the "delta" connected armature winding will be  $\sqrt{3}$  times as great as the voltage to the common connection of this equivalent "Y" winding, therefore the voltage between the collector rings will be,—

$$\frac{\sqrt{3}V}{2\sqrt{2}} = .612V,$$

where V=continuous-current voltage at commutator.

Inasmuch as a "delta" connected winding cannot be readily conceived to have a point of zero potential, the above subterfuge of substituting for it, the equivalent "Y" connected winding, will often be found to facilitate the handling of three-phase winding problems. When doing so, the equivalent "Y" potential and the equivalent "Y" current may be spoken of as attributes of a "delta" connected armature. In the accompanying figure, an equivalent "Y" connected winding is diagrammatically shown dotted within a "delta" connected winding.

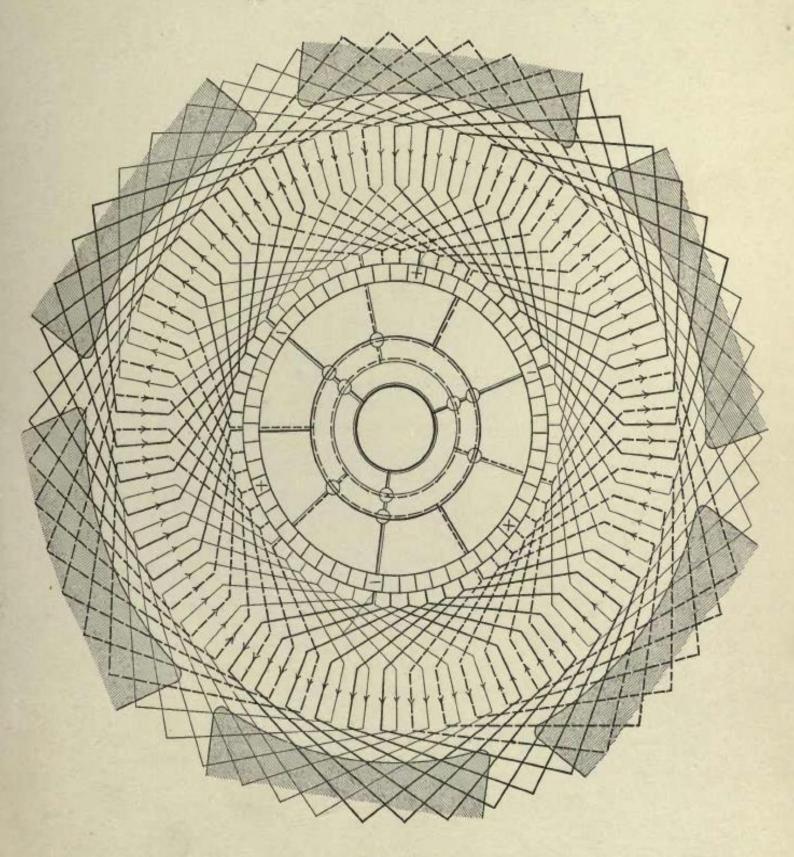
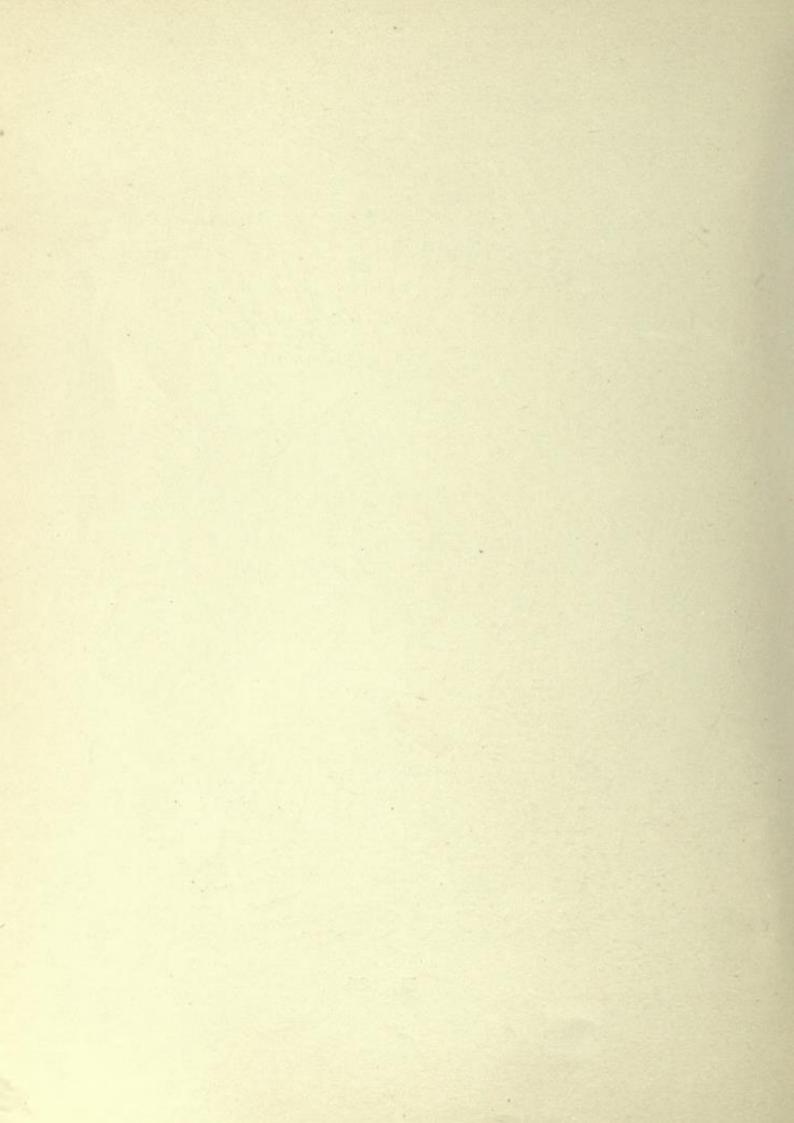


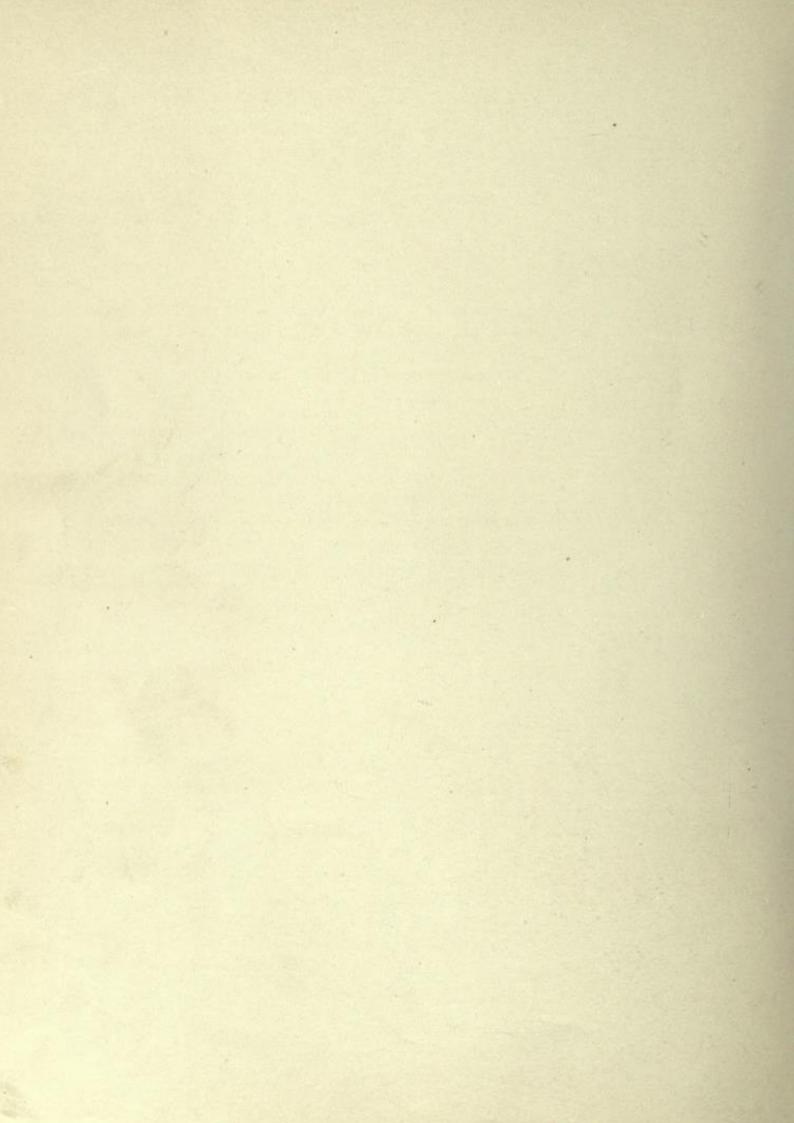
Fig. 129



# PART III.

WINDING FORMULÆ AND TABLES.





### CHAPTER XVI.

#### FORMULÆ FOR ELECTROMOTIVE FORCE.

Comprehensive formulæ for the calculation of the electromotive force set up in armatures may be derived from the formula for the voltage in a circuit, in which the variation of magnetic flux is a simple harmonic function of the time. These formulæ are:—

1. V = 6.28 TNM 10<sup>-8</sup>, the maximum voltage set up in a cycle; 1.000

2.  $V = 4.44 \ TNM \ 10^{-8}$ , the effective voltage set up in a cycle; . 707

3. V=4.00 TNM  $10^{-8}$ , the mean or average voltage set up in a cycle,  $\frac{3}{2}$ 

where V is the voltage generated, in volts; T the number of turns in series, M the number of egs lines included or excluded by each of the T turns in a magnetic cycle, and N the number of magnetic cycles per second.

In armatures of alternators, the effective, or square root of the mean square of the electromotive forces is required, since this is proportional to the effective voltage, i.e. the voltage to maintain current C (square root of the mean square of the current), in a non-inductive resistance. In this case it is supposed that the T turns are so situated as to be simultaneously affected by any change of the magnetic flux, otherwise the voltage for each of the turns differently situated must be calculated separately and properly combined to obtain the resultant voltage.

In the case of multi-phase alternating-current machines, the voltage in each circuit should be calculated, and the resultant voltage derived according to the method of connection, and addition of vectors according to the angle by which the several phases differ from each other.

In quarter-phase machines with common connection, the resultant voltage is  $\sqrt{2}$ , or 1.414 times the voltage generated in one circuit.

In three-phase apparatus, the resultant voltage is the same as the voltage generated in one circuit when the circuits are connected "delta"; and  $\sqrt{3}$ , or 1.732 times the voltage generated in one circuit when the circuits are connected "Y."

In alternating-current commutating machines, the ratio of the voltage between the continuous and the alternating current circuits is 1:.707 in the case of single-phase and quarter-phase commutating machines, and 1:.612 in the case of three-phase commutating machines. In other words, if the voltage at the con-

tinuous current side is known, the voltage between collector rings will be .707 times as great in the case of single and quarter phase commutating machines, and will be .612 times as great in the case of three-phase commutating machines.

In armatures of continuous-current dynamos, the voltage at the terminals is constant during any period considered, and is the integral of all the voltages successively set up in the different armature coils according to their position in the magnetic field; and since in this case only average voltages are considered, the resultant voltage is independent of any manuer in which the magnetic flux may vary through the coils.

Formula 3 is applicable to all continuous-current armatures, whether ring, drum or disc, two-circuit or multiple circuit, and whether the winding be single or multiple.

The simplicity and wide applicability of these formulæ make them preferable to many others that are difficult to interpret, because of the many accessory conditions that must be kept in mind.

Although, by the constants given above, the voltages may be obtained at the alternating current, as well as at the continuous current terminals of commutating machines, the former, i.e. the voltages at the alternating current terminals, may be obtained from the following formulæ, in which V is the required voltage between collector rings, T is the number of turns in series between collector rings, M is the magnetic flux from one pole piece into the armature, and N is the number of cycles per second:—

For single and quarter phase commutating machines, V = 2.83 TNM  $10^{-8}$ . For three-phase commutating machines, V = 3.69 TNM<sup>-8</sup>.

#### CHAPTER XVII.

#### METHOD OF APPLYING THE ARMATURE WINDING TABLES.

THE nature and use of the tables may be most easily understood by applying them to the solution of a few examples.

Example 1. — If we wish a two-circuit, triple winding for a drum armature, with about 670 conductors and six poles, what is the exact number of conductors that must be employed to give us a singly reentrant winding?

Turning to page 312, we find that a two-circuit, triple winding with 670 conductors, is impossible for six poles, but that 672 conductors may be used; and to have the winding singly re-entrant, the front and back pitches must each equal 113. If the front and back pitches should be taken equal to 111, a triply re-entrant winding would result.

EXAMPLE 2. — We next wish to ascertain how many volts this machine will give when the armature is driven at 440 r.p.m., if the flux from each pole piece into the armature equals 2.25 megalines.

The table of Drum Winding Constants on page 280 tells us that with 100 conductors, 100 r.p.m., and a flux equal to one megaline, the terminal volts will, for a six-pole machine, be equal to 1.667. Therefore, in the case before us, we have

#### $V=1.667\times6.72\times4.40\times2.25=111$ volts.

From the same table we find that for a two-circuit, triple winding with six poles, we have .200 average volts between commutator segments per megaline and per 100 r.p.m. So, in this case, we shall have  $.200 \times 2.25 \times 4.40 = 1.98$  average volts between commutator segments.

EXAMPLE 3. — Certain conditions fix the flux of a dynamo from one pole piece into the armature at 8.30 megalines, and the speed at 100 r.p.m. If we wish to employ an eight-pole, two-circuit, double winding, how many conductors do we need, to obtain 150 volts?

Consulting the table of Drum Winding Constants, on page 280, we find that for eight-pole, two-circuit, double windings, we have 3.33 volts per 100 conductors with 100 r.p.m., and one megaline of flux. Therefore, we shall require  $\frac{150}{3.33} \times \frac{100}{8.30} = 544$  conductors.

By reference to page 301, it will be seen that for eight poles, the nearest number of conductors that we can use in order to have a two-circuit, double winding, is 540 or 548. Suppose we use 540 conductors. If we wish a doubly re-entrant winding, we shall take the pitch at one end equal to 67, and that at the other end equal to 69.

EXAMPLE 4. — A slotted armature is to have ten poles, and a two-circuit, triple winding, with eight conductors per slot.

By reference to the table of Summarized Conditions for Two-Circuit, Triple Windings, on page 283, we find that it may be either singly or triply re-entrant, according to the number of conductors used.

The winding is to have 424 conductors. Turning to page 310, it is seen that the pitch must be 43 at both ends, and that for 424 conductors the winding must be singly re-entrant.

If the flux is 20.0 megalines, and the speed 105 r.p.m., we find from page 280 that the voltage will be

$$2.78 \times 4.24 \times 1.05 \times 20.0 = 247$$
 volts.

The average volts per bar are

$$.556 \times 20.0 \times 1.05 = 11.7$$
 volts.

EXAMPLE 5.—An eight-pole armature has a multiple-circuit, double winding, with 1258 conductors. By consulting page 343, we find that it is singly re-entrant, and that the pitch should be 155 at one end, and 159 at the other. It is, of course, understood that these pitches are taken in opposite directions. One of them might have been indicated as positive, and the other as negative. It may be well to point out here that the letters F and B at the head of the tables, meaning respectively, "front" and "back," are interchangeable, meaning merely that the one figure represents the pitch at one end, and the other figure, that at the other end. This is true in regard to all the tables, both two-circuit and multiple-circuit.

Returning to Example 5, the voltage of the machine, assuming the flux equals 7.85 megalines, and a speed of 300 r.p.m., is found by the table of Drum Winding Constants on page 280, to be

$$.833 \times 12.58 \times 3.00 \times 7.85 = 247$$
 volts.

The average volts per bar are

$$.1333 \times 7.85 \times 3.00 = 3.14$$
 volts.

Example 6. - A two-circuit, single winding is wanted, with four conductors per slot.

From the table of Summarized Conditions for Two-Circuit, Single Windings, on page 281, it may be seen that this is only possible with 6, 10, 14, etc., poles; being impossible with 4, 8, 12, 16, etc., poles. The winding is designed for fourteen poles, and 660 conductors. We find from page 329, that the pitch is 47 at both ends. The machine gives 160 volts, and the speed is 75 r.p.m. By the aid of the table on page 280, we find that the flux is equal to

$$\frac{160}{11.67 \times 6.60 \times .75} = 2.77$$
 megalines.

Average volts per commutator segment =  $3.27 \times 2.77 \times .75 = 6.80$  volts.

The above examples have all been chosen merely to illustrate the use of the tables, and the relative magnitudes employed in any one example are not such as would occur in practice.

The tables on pages 280, 281, 282, and 283 are constructed on the assumption that no interpolated commutator segments are employed, and that no portion of the normal number of commutator segments is omitted, and when this is not the case, the results should be properly modified, as may readily be done.

In all the tables, a proper interpretation of the term "conductors" should be made. As stated in the introductory chapter, "groups of conductors" may often be substituted therefor.

It is believed that after becoming familiar with the arrangement of the tables, their use will be found to be of value in a great variety of problems connected with armature windings. Any single result can, however, be obtained by an application of the rules and formulæ given in the text, but after these rules and formulæ are once understood, it will be found that subsequent problems will generally be most conveniently solved by means of the tables.

# CHAPTER XVIII.

ARMATURE WINDING TABLES.

# DRUM WINDING CONSTANTS.

						NUMB	ER OF	POLES		
		CLA	SS OF WINDING.	4	6	8	10	12	14	16
	88	₩ <sub>+</sub>	Single	1.667	1.667	1.667	1.667	1.667	1.667	1.667
	AND LINE.	MULTIPLE	Double	.833	.833	.833	.833	.833	.833	.833
	TS PER 100 CONDUCTORS PER 100 R. P. M. AND FLUX-ONE MEGALINE.	MUL	Triple	.556	,556	.556	,556	.556	.556	.556
si Si	TS PER 100 COND PER 100 R. P. M.	T	Single	3.33	5.00	6.67	8.33	10.00	11.67	13.33
JRE	PER 100 R 100 R.	TWO	Double	1.667	2.50	3.33	4.17	5.00	5,83	6.67
ARMATURES.	PER	CIRC	Triple	1.111	1.667	2.22	2.78	3.33	3.89	4.44
ARM		щ	Single	.1333	.200	.267	.333	,400	.467	.583
	MMI P. N.	MULTIPLE	Double 🛞	.0668	.100	.1333	.1667	.200	.233	.267
DRUM	AVERAGE VOLTS BETWEEN COMMUTA- TOR SEGMENTS PER MEGA LINE & PER 100 R. P. M. (INDEPENDENT OF NO. OF CONDS.)	MUL	Triple ⊗	.0445	.0667	.0888	.1111	.1333	.1555	.1778
_	TWER THE PER I		Single	.267	.600	1.068	1.668	2.40	3.27	4.27
	S BE SEG ENDE	TWO	Double ⊗	.1333	.300	.534	.834	1,200	1.667 .833 .556 11.67 5.83 3.89 .467 .233 .1555	2.14
	VOLT TOR LIN	CIR	Triple &	.0888	.200	.356	.556	.800	1.09	1.42

With Multiple Windings, the maximum Volts per bar is much more greatly in excess of the average Volts per bar than in Single Windings. This may be seen by a careful analysis of such Windings; which also shows that this may be more or less overcome by careful mutual adjustment of the position of the Brushes. This would not, however, be practicable with present methods.

NUMBER OF POLES		FC	OR DE			S PER	1000			VOLTS PER 100 CONDRS. PER 100 R.P.M.WITH FLUX == 1 MEGALINE	AVERAGE VOLTS BETWEEN COMR. SEGTS PER MEGALINE & PER 100 R. P. M
4	1	2		6		10		14		3.33	.267
6	1	2	4		8	10		14	16	5.00	,600
8	1	2		6		10		14		6.67	1,068
10	1	2	4	6	8		12	14	16	8.33	1.668
12	1	2		-		10		14		10.00	2.40
14	1	2	4	6	8	10	12		16	11.67	3.27
16	1	2		6		10		14		13.33	4.27

From the above Table the following Rule may be deduced:

In the ordinary two-circuit single winding, "O" is always such a number that the number of conductors per slot, and "n" the number of poles, cannot have a common factor greater than 2.

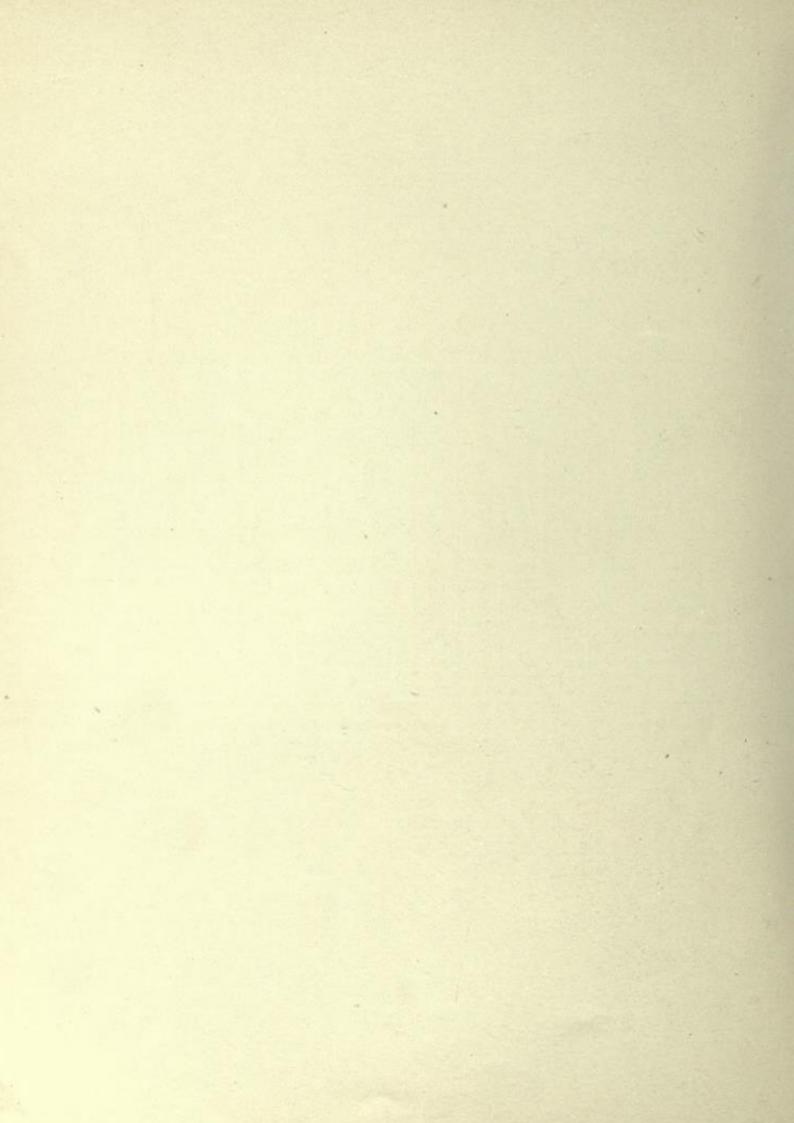


		FO	R DRU	MAR	MATU	RES.				VOLTS PER 100 CONDRS. PER 100 R.P.M. WITH FLUX = 1 MEGALINE	AVERAGE VOLTS BETWEEN COMR. SEGTS. PER MEGALINE & PER 100 R. P. M.
NUMBER OF			(	CONDU	CTORS F	PER SL	ОТ			DRS. DRS. 1.W	AVERAC VOLTS BET COMR. SEGT MEGALINE 100 R. P.
POLES	1	2	4	6	8	10	12	14	16	CON CON	NEG WEG
4	@ ·	@	(0)	(0)	0	00	00	000	@	1.667	.1333
6	00	00	00		00	@	F	000	00	2.50	300
8	00	00	000	@		@	000	000		3.33	.534
10	00	@	00	@	00		00	@	00	4.17	.834
12	000	@	(D)		0	00		00	0	5.00	1,200
14	@	000	00	00	00	00	00		00	5.83	1.635
16	00	@	00	00		@	@	00		6.67	2.14

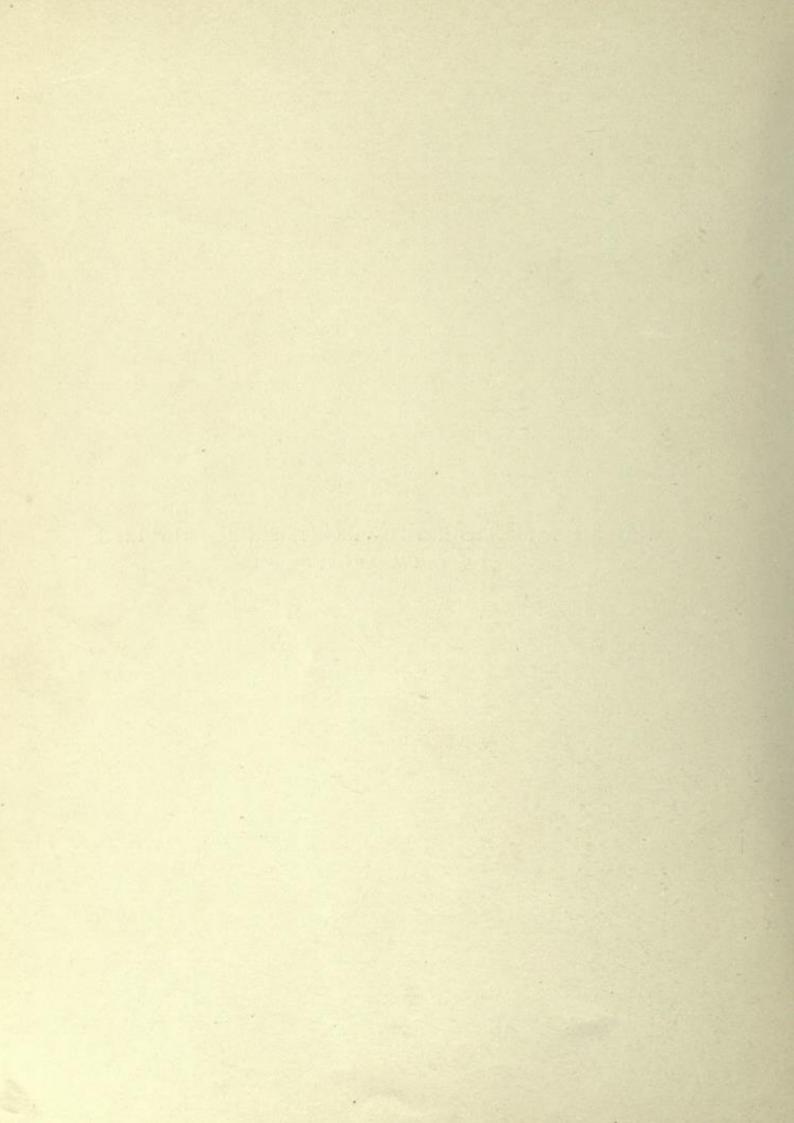
① Moreover, in multiple Windings this value is merely nominal, as a careful analysis of Multiple Windings shows that if this value can be approached at all, it is only by means of more careful mutual adjustment of the Brushes than is practicable with present methods.

1000		FOR	DRU	MARN	IATU	RES.		1002		VOLTS PER 100 CONDRS. PER 100	COMMR. SEGTS.
NUMBER				CONDUC	TORS F	PER SLO	TO			R.P.M. WITH FLUX  = 1 MEGALINE	PER MEGALINE & PER 100 R. P. M.
OF POLES	1	2	4	6	8	10	12	14	16		8
4	(QQ) 000	(D)		000		(00)		000		1,111	.0888
6	@	(00)	(QQ)	@	(00)	@	@	(R)	@	1.667	.200
8	(QQ)	(D)		000		(QQ)		@		2.22	.356
10	(22)	@	(00)	000	@		000	(00)	(000	2.78	,556
12	(QQ)	000		(QQ)		000		000		3.33	.800
14	@	000	000	000	000	(20)	000		(0)	3.89	1.09
16	(00)	000		000		@		@		4,44	1.42

Moreover, in Multiple Windings this value is merely nominal, as a careful analysis of Multiple Windings shows that if this value can be approached at all, it is only by means of more careful mutual adjustment of the Brushes than is practicable with present methods.



WINDING TABLES FOR TWO-CIRCUIT, SINGLE WINDINGS FOR DRUM ARMATURES.



TAB	LE C	FT	VO-C	IRC	JIT,	SINC	LE V	VIND	INGS	, FO	R DF	RUM .	ARM.	ATUI	RES.
rors					FR	ONT	AND	BACE	C PIT	CHES			-	2-	FORS
No. OF CONDUCTORS		LES	Po	LES	100000000000000000000000000000000000000	B LES	10.000	0 LES	POI	2 LES	17700	4 LES	The Assessment	6 LES	No. OF CONDUCTORS
No. O.	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. OF
102 104	20)	_#_	17	17	13	13	9	11							102 104
106 108	37	9T 9T	17	19	13	13	11	11	9	9					106 108
110	\$ T	27 20	17	19	13	15	I Carrie	1	9	9	7	9	7	7	110
112 114	B	20 29	19	19	13	15	11	11			7	9	7	7	112 114
116 118	20 20	29 m	19 19	19 21	15	15	11	13	9	11					116 118
120			100												120
122 124	200	an an	19 21	21 21	15	15	11_	18	9	_11	9	9			122 124
126 128	31. 30	80 33	21	21	15	17	13	13			9	9	7	9	126 128
130 132	31	51 33	21_	23	15	17_	13	13	11	11	- 4		7	9	130 132
134	38 38	88 80	21	28	17	17	AM		11	11					134
136 138	88	SI. 30	23	23	17	17	13	15			9	11			136 138
140 142	80	18	23 23	23 25	17	19	13	15	11	13	9	11	9	9	140 142
144 146	E II	97 90	23	25	17	19			11	13			9	9	144 146
148			25	25			1.5	15	11	10				-	148
150 152	ST ST	部	25	25	19	19	.15	15			11	11			150 152
154 156	8f	86 33	25	27	19	19			13	13	11	11			154 156
158 160	27	8F)	25 27	27 27	19	21	15	17	13	13			9	11	158 160
162	87	- 65	E	100	19	21	15	17			100		9	11	162
164 166	6	41	27 27	27 29	21	21			13	15	11	13			164 166
168	66 47	62	27	29	21	21	17	17	13	15	11	13			168 170
172 174	8	65	29	29	21	23	17	17					11	11	172 174
176			29	29				- 12	-					A CONTRACTOR	176
178 180	0	45	29	31	21	23	17	19	15	15	13	13	11	11	178 180
182 184	45	91	29 31	31 31	23	23	17	19	15	15	13	13			182 184
186 188	0	6	31	31	23	23	19	19						0	186
190	47	6) 6)	31	33	23	25			15	17			_11_	13	188 190
192 194	17	40	31	33	23	25	19	19	15	17	13	15	11	13	192 194
196 198	69 40	18	33	33	25	25	19	21			13	15			196 198
200		-	33	33	-		-				40	10			200
	4		(	5	8	3	1	0	1	2	1	4	1	6	



	BLE (	OF T	WO-0	CIRC	UIT,	SINC	GLE \	WIND	ING	s, FO	R DI	RUM	ARM	IATU	
TORS					FRO	NT A	ND B	ACK I	PITCH	IES.					TORE
No. OF CONDUCTORS	Po	LES		6 LES	1 3	8 LES	70	0 LES	42.77	2 LES		4 LES	1 2 2 2 3	6 LES	No. OF CONDUCTORS
No. 0F	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. 0F
202	49 51	21	33	35	25	25	19	21	17	17					202 204
206 208	88 53	61	33 35	35 35	25	27	21	21	17	17	15	15	13	13	206 208
210 212	61	63 50	35	35	25	27	21	21			15	15	13	13	210 212
214 216	8.0° 6.0°	12	35	37	27	27			17	19		-			214 216
218 220	\$5 \$6	10	35 37	37	27	27	21	23	17	19					218 220
222	55	27	37	37	27	29	21	23			15	17	13	15	222 224
226 228	#	57 82	37	39	27	29	23	23	19	19	15	17	13	15	226 228
230 232	62 62	6T 5-9	37 39	39 39	29	29	23	23	19	19	-				230 232
234 236	87 89	50	39	39	29	29					17	17			234 236
238 240	85	61	39	41	29	31	23	25	19	21	17	17	15	15	238 240
242 244	8	61	39 41	41	29	31	23	25	19	21			15	15	242 244
246 248	61	60	41	41	31	31	25	25							246 248
250 252	61	54 54	41	43	31	31	25	25	21	21	17	19			250 252
254 256	53	54	41	43	31	33			21	21	17	19	15	17	254 256
258 260	65	65 60	43	43	31_	33	25	27					15	17	258 260
262 264	86 88	66 67	43	45	33	33	25	27	21	23	19	19			262 264
266 268	67	67	43	45 45	33	33	27	27	21	23	19	19			266 268
270 272	67 67	60	45	45	33	35	27	27	0.0	200			17	17	270 272
274	67 69	69	45	47_	33	35	05	00	_ 23	23	10	0.1	17	17	274
278 280 282	69 69	11	45 47	47	35	35	27	29	23_	23	19	21 21			278 280
284	99 21	11	47	47	35	37	21	29	0.0	0.5	19	21	10	10	282 284 286
286 288 290	71	18	47	49	35	37	29	29	23	25 25			17	19	288 290
292 294	19	11	49	49	37	37	29	29	20	20	21	21	17	19	292 294
296 298	11	78 76 75	49	49 51	37	37	29	31	25	25	21	21			294 296 298
300	- 18	10	10	94	01	01	20	OI.	40	20					300
	4	ŧ	(	3		8	1	0	1	2	1	4	1	6	-

TAE	BLE (	OF T	WO-C	CIRC	UIT,	SINC	LE V	VIND	INGS	s, FO	R DF	RUM	ARM	ATU	RES.
roas					FRO	ONT A	ND E	BACK	PITC	HES		TET	4		ORS
No. OF CONDUCTORS		4 LES	The second second	6 LES		8 LES		0 LES		2 LES	F 100 155 1	4 LES	The same of the sa	6 LES	No. OF CONDUCTORS
No.OF	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No.OF
302 304	H	10	49 51	51 51	37	39	29	31	25	25			19	19	302 304
306	19	H	51	51	37	39	31	31			21	23	19_	19	306 308
310 312	- #	77	51	53	39	39	31	31	25	27	21	23			310 312
314 316	77	70	51 53	53 53	39	39			25	27					314 316
318 320	19	TI.	53	53	39	41	31	33			23	23	19	21	318 320
322 324	I.	at it.	53	55	39	41	31	33	27	27	23	23	19	21	322 324
326 828	N N	67	53 55	55 55	41	41	33	33	27	27		-	1000		326 328
330 332	61 52	4	55	55	41	41	33	33	000	200	0.0		-04	0.1	330
334 336 338	88 98	94 95	55	57	41	43	33	35	27	29	23	25	21	21	334
340 342	65 66 58	66 87	57	57	43	43	33	35	27	29	23	25	21	21	338 340 342
344 346	65 87	85 85	57 57	57 59	43	43	90	00	29	29					344
348 350	47	5/7 g/ll	57	59	43	45	35	35	29	29	25	25	21	23	348 350
352 354	H7	8M 89	_59	59	43	45	35	35			25	25	21	23	352 354
356 358	01 01	91	59 59	59 61	45	45	85	37	29	31					356 358
360 362	# 10	94 93	59	61	45	45	35	37	29	31	25	27			360 362
364 366 368	N.	UL UL	61	61	45	47	0.5	-0.7			25	27	23	23	364 366
370 372	91 80	9.5 341	61	63	45	47	37	37	31	31			23	23	368 370
374	93	163	61	63 63	47	47	37	37	31	31	27	27			372 374 276
378 380	18 99	12.	63	63	47	47	37	39			27	27			376 378 380
382 384	90 90	95 97	63	65	47	49	37	39	31	33	41	2.	23	25	382
386 388	06 00	90	63 65	65 65	47	49	39	39	31	33			23	25	386
390 392	85	NT 90	65	65	49	49	39	39			27	29			390 392
391	97 98	97	65	67	49	49			33	33	27	29			394 396
398 400	96 91	161 50	65 67	67 67	49	51	39	41	33	33			25	25	398 400
	4	ŧ	(	3	8	3	1	0	1	2	1	4	1	6	

TA	BLE	OF T	wo-	CIRC	UIT,	SIN	GLE '	WINI	DING	S, FC	OR D	RUM	ARN	IATU	RES.
FORS		100			FRO	NT A	ND B	ACK	PITCI	HES		1		-	rors
No. OF CONDUCTORS	1 200	4 LES	7.77	6 LES		8 LES		0 LES		2 LES	17/23/19	4 LES	10000	6 LES	No. OF CONDUCTORS
No.0F	F	В	F	В	F	В	·F	В	F	В	F	В	F	В	No.OF
402	101	381.	67	67	49	51	39	41			29	29	25	25	402
406 408	100	101	67	69	51	51	41	41	33	35	29	29			406
410 412	101	100	67 69	69	51	51	41	41	33	35					410
414	100	108 106	69	69	51	58	-	40	0.5	0.5	00	01	25	27	414 416
418 420 422	110 100 100 100	105 106	69	71	51	53	41	43	35	35	29	31	25	27	418 420 422
424 426	165  06  07	107 107 107	71	71	53	53	41	9.0	90	00	2.0	01			424 426
428 430	lest lest	307	71 71	71 73	53	55	43	43	35	37			27	27	428 439
432 434	EZ.	109	71	73	58	55	43	43	35	37	31	31	27	27	432 434
436 438	100 100	200 111	73	73	55	55	43	45			31	31			436 438
440 442	209	111	73 73	73 75	55	-55	43	45	37	37					440
444 446 448	111	101	73 75	75 75	55	57	45	45	37	37	31	33	27	29	444 446 448
450 452	111	115	75	75	55	57	45	45			31	33	27	29	450 452
454 456	118	113 115	75	77	57	57	10	10	37	39					454 456
458 460	H	115	75 77	77	57	57	45	47	37	39	33	33			458 460
462 464	116	115	77	77	57	59	45	47			33	33	29	29	462 464
466 468 470	118 117	117	77	79	57	59	47	47	39	39			29	29	466 468
472 474	111 111	110 110 110 110	79	79	59 59	59	47	47	39	39	33	35			470 472 474
476 478	110	119 121	79 79	79 81	59	61	47	49	39	41	33	35	29	31	476 478
480 482	110	12	79	81	59	61	47	49	39	41			29	31	480 482
484	ter ter	121 122	81	81	61	61	- 10	10							484 486
488 490 492	191	169	81 81	81 83	61	61	49	49	41	41	35	35			488
494 496	韻	193 125	81 83	83 83	61	63	40	40	41	41	35	35	31	31	492 494 496
498 500	197	155 115	83	83	_61_	63	49	51					31	31	498 500
		ŧ	6	3	8	3	1	0	1	2	1	4	1	6	

TAB	LE C	FTV	vo c	IRCU	JIT, S	SING	LE W	IND	INGS	, FO	RDR	RUM .	ARM	ATUI	RES.
CORS					FRO	NT A	AND:	BACK	PITO	CHES					rors
No. OF CONDUCTORS	Po	LES	Po	LES		B LES		0 LES	4.5	2 LES	1500000	4 LES	72.00	6 LES	No. OF CONDUCTORS
No. OF	F	В	F	В	F	В	F	В	F	В	F	В	F	B	No. 0
502 504	156 155	125	83	85	63	63	49	51	41	43	35	37			502 504
506 508	12	18	83 85	85 85	63	63	51	51	41	43	35	37			506 508
510 512	12f 127	121	85	85	63	65	51	51					31	33	510
514 516	127 127	109 129	85	87	63	65	31	01	43	43	37	37	31	33	514
518 520	123	197	85 87	87 87	65	65	51	53	43	43	37	37			516 518
522 524	H	H	87	87	65	65	51	53			31	91			520 522 524
526 528	105- 131	123	87	89	65	67	53	53	43	45			33	33	524 526 528
530 532	13	12	87 89	89 89	65	67	53	53	43	45	37	39_	33	33	530
534 536	111	110	89	89	67	67	9-0	00			37	39			532 534
538 540	133 335	135 145	89	91	67	67	53	55	45	45					536
542 544	130 138	100	89 91	91 91	67	69	53	55	45	45	20	20	33	35	540 542
546 548	10	377	91	91	67	69	55	55			39	39	33	35	544 546
550 552	10	1 at 133	91	93	69	69	55		45	47	39	39			548 550
554 556	197	139 120	91 93	93	69	69	50	55	45	47					552 554 556
558 560	100	147	93	93	69	71	55	57			39	41	35	35	558 560
562 564	129	111	93	95	69	71	55	57	47	47	39	41	35	35	562 564
566 568	14I 14I	140	93 95	95 95	71	71	57	57	47	47					566 568
570 572	141	163	95	95	71	71	57	57			41	41			570 572
574 576	343 143	140 145	95	97	71	73	7,	-	47	49	41	41	35	37	574 576
578 580	140	143 145	95 97	97 97	71	73	57	59	47	49	34	1	35	37	578 580
582 584	100	146	97	97	73	73	57	59							582 584
586 588	146	H.	97	99	78	73	59	59	49	49	41	43			586 588
590 592	16 16	147 149	97	99 99	73	75	59	59	49	49	41	43	37	37	590 592
594 596	14T 249	149 149	99	99	73	75	00	Ų.					37	37	594 596
598 600	120	125	99	101	75	75	59	61	49	51	43	43			598
	4	t	(	3	(	3	1	0	1	2		4	1	6	000

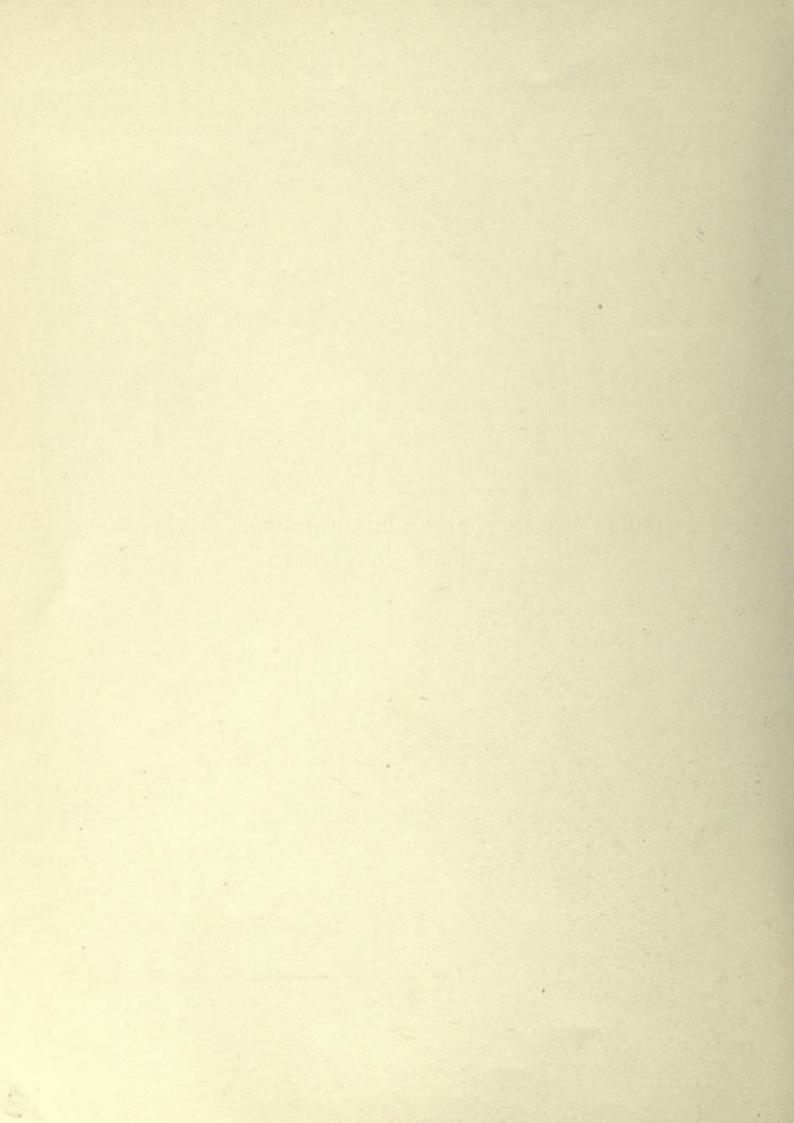
TA	BLE	OF T	wo	CIRC	TIUS	SINC	ILE V	VINE	ING	S FO	R DR	UM .	ARM	ATUE	RES.
TORS					FRC	NTA	ND E	BACK	PITC	HES					TORS
No. OF CONDUCTORS	CONTRACTOR OF THE PARTY OF THE	4 LES	007	B LES	PEDATO-D	8 LES	1 POI	V-751 / V-120 / L	0.0000000000000000000000000000000000000	2 LES	544440000000	4 LES	CANADA	6 LES	No. OF CONDUCTORS
No. 06	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. O
602	155	161	99 101	101 101	75	75	59	61	49	51	43	43			602 604
606	161	151 153	101	101	75	77	61	61					37	39	608
610 612	192	150	101	103	75	77	61	61	51	51			37_	39_	610
614	151	150 104	101	103	77	77	22		51	51	43	45			614
618 620 622	150 150 150	155 156 156 157	103	103	77	77	61	63	24	53	43	45	39	39	618 620 622
624 626	155	182 187 188	103	105	77	79	- 61	00	51	53			39	39	624 626
628 630	167 167	150 150	105	105	79	79	63	63	01	00	45	45	-	00	628 630
632 634	167	150	105 105	105 107	79	79	63	63	53	53	45	45			632 634
636 638	169 180	158	105	107	79	81	63	65	53	53			39	41	636 638
640 642	169 343	161 161	107	107	79	81	63	65			45	47	39	41	640
644 646 648	105 101	141 140	107 107	107 109	81	81	65	65	53	55	45	47			644 646 648
650 652	161 168	160 160	107 109	109 109	81	81	65	65	53	55					650 652
654 656	568 143	165	109	109	81	83	-00				47	47	41	41	654 656
658 660	104	165 163	109	111	81	83	65	67	55	55	47	47	41	41	658 660
662	10a 15a	105	109	111	83	83	65	67	55	55					662
666	101	167	111	111	83	83	67	67				10		***	666
670 672 674	107 101 102 100	150 150 150	111	113	83	85 85	67	67	55	57	47	49	41	43	670 672 674
676 678	160 160	165 171	113	113	85	85	67	69	55	57	47	49	41	43_	676 678
680 682	161	171	113 113	113 115	85	85	67	69	57	57					680
684 686	191	171 172	113	115	85	87			57	57	49	49	43	43	684 686
688	171 173	]78 113	115	115	85	87	69	69			49	49	43	43	688 690
692 694 696	173	173 175	115 115	115 117	87	87	69	69	57	59					692
098 700	172	374 375	115 117	117 117	87	87_	69	71	57	59	49	51			696 698 700
	1	ŧ		3	8	3	1	0	1	2	1	4	1	6	1.55

TAI	BLE	OF T	wo (	CIRC	UIT	SINC	LE V	VIND	INGS	FOI	RDR	UM A	ARMA	ATUF	RES.
TORS					FRO	NTA	AND I	BACK	PITC	HES.					TORS
CONDUCTORS	775	4 LES	Pol	LES	10000	B LES	100000	0 LES		2 LES	100000	4 LES	-100 C 1 PM	6 LES	Ma, OF CONDUCTORS
No. 0F	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. OF
702 704	123	177	117	117	87	89	- 69	71			49	51	43	45	702 704
706	115 177	H	117	119	87	89			59	59			43	45	706
708	900 877	117 119	117	119	89	89	71	71	59	59					708 710
712	\$27 319	120	119	119	89	89	71	71			51	51			712
716 718	179 120	120	119 119	119 121	89	91	71	73	59	61	51	51	45	45	716 718
720	The state of the s								1000						720
722 724	137	181	119 121	121 121	89	91	71	73	59	61			45	45	722 724
726	141	151	121	121	91	91	73	73			51	53			726 728
730	10	162	121	123	91	91	2		61	61	51	58			730
734	160	168	121	123	91	93	73	73	61	61			45	47	732 734
736 738	101	1/15 1/15	123	123	91	93	73	75					45	47	736 738
740 742	130	165	123 123	123 125	93	93	73	75	61	63	53	53			740
744							10	10			53	53			742
746	185	347 347	123 125	125 125	93	93	75	75	61	63		-			746 748
750 752	187	100	125	125	93	95	75	75					4.7	47	750 752
754 756	100	100	125	127	93	95			63	63	53	55	47	47	754
758	120	100	125	127	95	95	75	77	63	63	53	55			756 758
760 762	100	19	127	127	95	95	75	77							760
764 766	191 161	197	127 127	127 129	95	97			63	65			47	49	764 766
768 770	1903 1903	100	127	129	95	97	77	77			55	55			768
772	10000000		129	129			77	77	63	65	55	55	47	49	770 772
774 776	193	196 196	129	129	97	97	-								774
778 780	123	100	129	131	97	97	77	79	65	65					778 780
782 784	170	107	129 131	131 131	97	99	77	79	65	65	55	57	49	49	782
786	195	197			97	99					55	57	49	49	784 786
788 790	107	107	131 131	131	99	99	79	79	65	67					788 790
792 794	100	100	131	133	99	99	79	79	65	67					792
796		100 2014	133	133			-	0.1	00	- 01	57	57			794 796
798 800	199	900	133	133	99	101	79	81			57	57	49	51	798 800
	4	t	6	3	8	3	1	0	1	2	1	4	1	6	





WINDING TABLES FOR TWO-CIRCUIT, DOUBLE WINDINGS FOR DRUM ARMATURES.



TA	BLI	E O	FT	wo	-CI	RC	UIT	, D	OUI	BLE	w	ND	INC	as,	FO	R D	RU	M A	RM	IAT	UR	ES.
FORS							F	RON	IT A	ND	BA	CK :	PIT	CHE	S							ORS
CONDUCTORS	P	4 OLI	ES	P	6 OL	ES	P	8 OLI	ES	P	10 OLI	ES	P	12 OLI		P	14 OLE		P	16 OLI		No. OF CONDUCTORS
No.0F	F	PE- ENTRANCI	В	F	AE- ENTRANC	В	F	HE- ENTARHE	В	F	HE- ENTRANCY	В	F	RE- ENTRAC	В	F	RE-	В	F	RE-	В	-
102 104 106	請	00	90 21	17	00	19				9	00	11	9	@	9	7	0	7				102 104 106
108 110	26 37	00	27 23	19	00	19	H	88	10		00	**				7	00	9	7	@	7	108 110
112 114 116	20 20 20 20 20 20 20 20 20 20 20 20 20 2	00	E	17	00	19	13	88	10	11	(D)	11 13	9	00	9	7	00	9	7	@	7	112 114 116
118 120	20	@	25 81	19	@	19			10			10								30		118 120
$     \begin{array}{r}       122 \\       124 \\       126     \end{array} $	29	00	14	19	00	21	15	878	19	11 13	。 @	13 13	9	00	11_	9	@	9	7	00	9	122 124 126
128 130 132	31 15 15 15 15 15 15 15 15 15 15 15 15 15	00	#1 #2 #2	21	00	23	No.	88	II.				11	@	11	9	00	9	7	00	9	128 130 132
134 136	89	@	15	23 21	00	23 23		CILD	11	13 13	00	13 15	11	ගව	11	9	00	11		00	3	134 136
138 140 142	10 50	00	80	23	00	25 28	H	88	19				11	00	13				9	@	9	138 140 142
144 146	10 Mar.	00	85 87	25 23	(0)	25	ST ST	40	10	13 15	00 (D)	15 15	-		10	9	00	11		-	9	144 146
148 150 152	87 89	00	17 10 12 12	25	00	25	j)	88	12				13	(D)	13	11	0	11	9	@	9	148 150 152
154 156 158	187 190	00	19 41	25	@	25	12	8	29 21	15 15	00	15 17				11	(3)	11	9	00	11	154 156 158
160 162	31	@	45	25	00	27		0.0					13	@	13							160 162
164 166 168	- 11 - 11 - 11	00 (B)	41 41 42	27	@	29	17	%	21	15	(0)	17	13	00	15	11	00	13	9	0.0	11	164 166 168
170 172 174	41.	00	431	29	00	29 29	1	88	21 21	17	(2)	17	13	0.0	15	11	0.0	13	11	@	11	170 172 174
176 178	42	ග	43 46	29 29	ං ග	31 29				17	00	19	15	@	15	13	00	13				176 178
180 182 184	45 45 47	00	10 T	31 29	00	31	21	96	II.	17	00	19	15	@	15				11	@	11	180 182 184
186 188 190	45	00	47	31	00	33	22	88	15 25	19	@	19	15	00	17	13	00	13	11	00	13	186 188
192 194	47 49	@	45	33	0	33				19	@	19				13	00	15				190 192 194
196 198 200	45 49 - 01	ග	49 51 40 61	31	00	33	25	88	能	19	00	21	15	00 (D)	17	13	00	15	11	00	13	196 198 200
		4			6			8			10			12			14			16		



	1	rw(	o-c	IRC	Uľ	т, п	2000		_							JM	ARI	MΑ	ΓUΈ	RES.		1
TORS							F	RON	IT A	AND	BA	CK	PIT(	CHE	S							LORS
CONDUCTORS	P	4 OLE	ES	P	6 OLI	ES	P	8 OLI	ES	P	10 OLI		P	12 OLI	ES	F	14 OLI	ES	P	16 OLE	S	No OF CONDICTORS
No.0F	F	RE*	В	F	RE- ENTRANC	В	F	HE- EHTHANG	В	F	RE-	В	F	RE-	В	F	ME-	В	F	RE- ENTRANCE	В	
202	19	00	11	33	0	33	26 25	88	11	19	00	21							13	(3)	13	20
206 208	12	@	11	35	00	35 35				21	0	21	17	0	17	15	0	15				20
210	68				00									00								21
212	12	0.0	66	35	00	37	10	86	all all	21	0	21	17	00	19	15	0	15	13	00	13	21
216	17 55	@	13 14							21	00	23				10		10				21
218 220	22	00	56	37	00	37	gr gr	88	#				17	00	19	15	00	17	13	00	15	21 22
222 224	- 17	0	SE SE	37	00	39				21	00	23	19	0	19							22 22
226	C. C.		-000	37	0	37	95	6.0	- 05	23	0	23		-	- 44			17	10		15	22
228 230	66	0.0	80	39	0	39	27	86	88					-		15	00	.17	13	0.0	15	22
232 234	粧	(0)	85 85	37	00	39				23	@	23	19	00	19	17	@	17				23
236	17	00	80	39	00	41	25 25	88	17	23	00	25	19	00	21	71	GO	- 1.1	15	@	15	23
238 240	111	@	80	39	(0)	39				-									1000		- 111	23
242 244		00		41	0	41	-	0.0		00		0.5	10		01	17	0	17			4.5	24
246	87	1-23	21	39	00	41	30	88	it	23 25	00 00	25 25	19	00	21				15	(3)	15	24 24
248 250	55	00	63	41	00	43 41							21	0	21	17	0.0	19				24
252 254	8	00	63				20 85	8%	31 33										15	00	17	25
256	· ·	0	27	43	00	43	-			25 25	00	25	21	0	21	17	0.0	19				25
258 260	63	00	55	43	00	45	B	88	8				21						15	00	17	25 26
262				43	@	43	29	60	22	No.			21	00	23	19	@	19	10	00	11	26
264 266	- N	00	61	45	0	45				25 27	00	27 27										26 26
268 270	51	00	27	43	00	45	203 203	88	100				21	00	23	10	-	10	17	00	17	26
272	NI On	@	57 69	45	00	47			-	-			23	0	23	19	0	19				27
274 276	67	00	60 71	45	00	45	100	98	36	27	@	27 29				10	0.0	0.1	17	(5)	17	27
278	-	- mercentage		47	00	47	35	40	36	21	00	20				19	00	21	17	0	17	27
280 282	n	@	71	45	00	47							23	0	23							28
284 286	n	00	71 73	47	00 @	49	56 30	88	B	27	00	29	23	00	25	19	00	21	17	00	19	28
288	11	00	1)							29	8	29										28
290 292	11	00	10	49	00	49	*	88	B				23	00	25	21	@	21	17	00	19	29
294 296	IS.			49 00 51			81	~		29	(3)	29	-						11	.00	10	29
298		@	378	49	00 00	51 49				29	00	31	25	0	25	21	@	21				29
300							H	88	65 89										19 @ 19			30
		4			6			8			10			12			14		16			

		TW	0-0	IR	cui	т, г	oot	JBL	ΕV	VIN	DIN	IGS	, F(	OR	DR	UM	AR	MA	TUI	RES		
TORS							FF	RON	TA	ND	BAC	CK F	PITC	HE	S		33				-	TORS
CONDUCTORS	P	4 OLI	ES	P	6 OLI	ES	P	8 OLE	ES	P	10 OLI	10345	P	12 OLE	ES	P	14 OLE	ES	P	16 OLE		CONDUCTORS
No.0F	F	es- ENTRACO	В	F	RE- ENTHANC		F	RE- ENTRANC	В	F	BE- ENTRANCY	В	F	RE- ENTRANCY	В	F	RE- ENTRANCY	В	F	AS- ENTRANCE	В	No.OF
302 304 306	19	0	19	51 49	00	51				29 31	00 ©	31	25	@	25	21	00	23				302 304 306
308 310	II.	00	Fig.	51 51	(D)	53 51	35_	86	30				25	00	27	01		00	19	യ	19	308 310
312 314 316	- H	00	13	53 51	©	53 53	30 30	88	30 41	31 31	@	31	25	00	27	21_	00	23	19	00	21	312 314 316
318 320 322	75	0	79	53 53	00	55 53							27	@	27	23	@	23				318 320 322
324 326 328	200	00	- EL	55 53	@	55 55	11	88	11:	31 33	(D)	33	27	@	27	23	0	23	19	00	21	324 326 328
330 382 334	81 83	00	20 20 20 20 20 20 20 20 20 20 20 20 20 2	55 55	00	57 55	3	88	13	33		00	27	00	29	23	00	25	21	0	21	330 332
336 338	92	00	2	57	0	57				33	00	33										334 336 338
340 342 344	88 88	00 (D)	86 87	55	00	57	됩	88	11	33	00	35	27	00 (D)	29	23	00	25	21	@	21	342 344
346 348 350	100	00	67 80	57	@	57	11	88	2	35	ග	35				25	00	25	21	0.0	23	346 348 350
352 354 356	5	00	97 80 80	57	00	59	9	86	16	35 35	(D)	35	29	(D)	29 31	25	0	25	01	00	23	352 354 356
358 360	90 91	00	W.	59	0	59	3	90		20	00	37	20	00	- 51	25	00	27	21_	0.0	20	358 360
362 364 366	80 91	00	91 98	61 59	00	61	世	88	91	35 37	00	37 37	29	00	31				23_	@	23	362 364 366
368 370 372	55 55 55 55	000	25	61	(D)	63	6 47	88	4				31	0	31	25	00	27	23	0	23	368 370 372
374 376 378	2	0	8	63 61	00	63 63				37 37	00	37 39	31	0	31	27	@	27				374 376 378
380 382	99 96	00	- 17	63 63	00 (D)	65 63	#	88	5	0.5		8.0	31	00	33	27	@	27	23	00	25	380 382
384 386 388	10 mm	00	\$15 \$15 \$15	65 63	(3) 00	65 65	3	88	40	37	<u>@</u>	39	31	00	33	27	00	29	23	00	25	384 386 388
390 392 394	81	@	88	65 65	65 00 6					39	(3)	319	33	@	33							390 392 394
396 398 400	88	65 CD 65 67 CD 67 CD 65 00 67					40	88	15	39	00	41	33	@	33	27	00	29	25	@	25	396 398 400
		4			6			8		10			12				14			100		

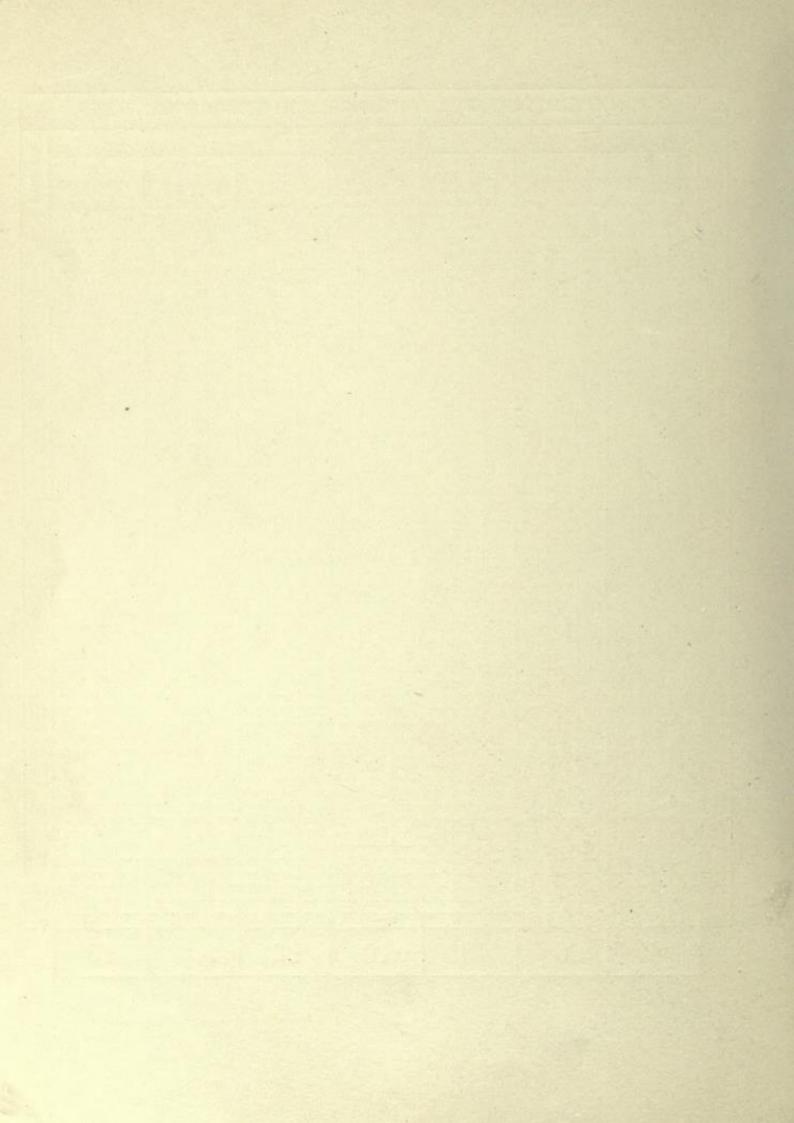
		TW	0-0	CIR	cu	т, і	DOI	JBL	ΕV	VIN	DIN	NGS	, F	OR	DR	UM	M ARMATURES.					
TORS							F	RON	IT A	ND	BA	CK I	PITO	CHE	S					Bul		TORS
CONDUCTORS	P	4 OLE	ES	P	6 OLI	ES	P	8 OLE	ES	P	10 OLI	ES	P	12 OLE		P	14 OLE		P	16 OLE	S	CONDUCTORS
No.OF	F	NE-	В	F	ME- ENTRANC	В	F	RE- SENTRANCY	В	F	RE- ENTRANCE	В	F	RE-	В	F	RE- ENTRANCE	В	F	RE- EMTRANCY	В	No. OF
402	100	00	105	67	00	69	17	88	87	39	00	41	33	00	35	29	00	29	25	0	25	402
406 408	11	@	101	67	(0)	67				41	(3)	41				00						406
410 412 414	12	00	106	69	00	69	61	8	2)	41	(D)	41	33	00	35	29	(0)	29	25	00	27	410 412 414
416	1/0	0	101	69 69	00 (D)	71 69				41	00	43	35	0	35	29	00	31				416 418
420 422	12	00	100	71	@	71	53 53	88	13				-		n F	06			25	00	27	420. 422
424 426 428	106 107	00	107	69	00	71 73	\$0 100	88	49	41	· · ·	43	35	00	35	29	00	31	27	@	27	424 426 428
430 432	107	8	100 107 109	71	@	71	-10	90	34				30	00	01	31	00	31	21	00	41	430 432
434 436	107	00	191	73 71	71 00 73			88	55 50	43 43	(D)	43	35	00	37				27	@	27	434 436
438	109	0	791	73	00	75							37	@	37	31	@	31				438
442 444 446	300 111	00	m	73	8	73	12	88	65 67	43	00	45 45				31	00	33	27	00	29	442 444 446
448 450	101	@	111	73	00	75				40		10	37	0	37							448
452 454	113	00	111	75 75	00 @	77 75	50	3.6	67	45	@	45_	37	00	39	31	00	33	27	00	29	452
456 458	111	@	1115	77	@	77	- 62	(0)		45	0.0	47	0.00		20	33	യ	33	00	700	00	456
460 462 464	115	· ·	111	75	00	77	Bi	88	13	45	00	47	39	00	39				29	(0)	29	460 462 464
466 468	115	00	H	77	@	77	25	88	10 14	47	@	47				33	@	33	29	@	29	466 468
470	317 917 917	@	111	79 77	00	79 79							39	@	39	33	00	35				470
474 476 478	115	00	\$10 \$21	79 79	。 @	81 79	95	88	80	47	00	49	39	00	41				29	00	31	474 476 478
480 482	_IM_	@	1M	81	@	81										33	00	35				480 482
484	110 121	00	121	79	00	81	智	88	81	47	00 (D)	49 49	39	00	41	35	@	35	29	00	31	484 486
488 490 492	101 101 101 100	@	193	81 @ 81			21	88	61 63				41	@	41				01	00	01	490
494 496	湿	@	00 H 83 @ 8				61	60	68	49	(D)	49 51	41	0	41	35	0	35	31	00	31	492 494 496
498 500	眉	00	101	83	00	85	61	86	60 63				41 00 43			35	00	37	7 31 @ 31		31	498 500
		4			6			8			10			12			14					

		TW	o-c	IRC	CUI	т, г	oou	BL	E W	VIN	DIN	GS,	, FC	OR I	DRU	JM	AR	MA	TUI	RES		
FORS							FI	RON	TA	ND	BAC	CK I	PITC	HE	S							TORS
No. OF CONDUCTORS	P	4 OLE	ES	P	6 OLI	ES	P	8 OLE	S	P	10 OLE	S	P	12 OLE	S	P	14 OLE		P	16 OLE	S	No. OF CONDUCTORS
THE REAL PROPERTY.	F	RE- ENTRANCE	В	F	RE-	В	F	HE- ENTRANCY	В	F	NE-	В	F	AC- ENTRANCE	В	F	DE- ENTOUNCY	В	F	AE- ENTRANSY	В	_
502	710 127	(3)	326 227	83	@	83				49	00	51			-						-	502 504
506 508	138	00	国	85 83	00	85 85	08 08	8	45 56	51	@	51	41	00	43	35	00	37	31	00	33	506 508
510							- 08	00	- 10						and the same	00		01	0.		- 00	510
512 514	127	0	133	85 85	00	87 85				51	00	51	43	00	43	37	0	37				512
516 518	335	00	177	87	0	87	68	88	08 08	51	00	53							31	0.0	33	516 518
520 522	320	(0)	15	85	00	87							43	@	43	37	00	37				520 522
524 526	123	00	333	87 87	00 (D)	89 87	66	R	2	51	ං ග	53 53	43	0.0	45				33	(2)	33	524 526
528 530	ISI.	@	[3]	89	0	89				-						37	00	39				528 530
532	1%	00	提	87	00	89	65	ಜಿ	計		05		43	00	45				33	00	33	532
534 536	150 130	(3)	101	89	00	91				53	00	53 55	45	00	45	37	00	39				534 536
588 540	D83	00	238 237	89	00	89	SE .	88	8										33	00	35	538 540
542 544	浙	0	125	91 89	00	91 91				53	00	55	45	@	45	39	00	39				542 544
546 548	135 181	00	181	91	00	98	9	868	25	55	(W)	55	45	00	47				33	00	35	546 548
550		Page 1	- WE	91	@	91	99	000	(3)			-	40	00	*1	39	(3)	39	00	00	00	550
552 554	-15	@	100	98	0	93		-		55	@	55										552 554
556 558	111	0.0	間	91	00	93	60	88	91	55	00	57	45	0.0	47	39	00	41	35	0	35	556 558
560 562	189	00	122	93	00	95 93							47	.00	47							560 562
564 566	Hi	0.0	141	95	0	95	19	18	11	55 57	00 (D)	57 57				39	00	41	35	(3)	35	564 566
568 570	15	0	141 163	93	00	95				01	00		47	@	47	41	0	41				568 570
572	18_	00	143 143	95	00	97	N	88	13	-			47	00	49	11	00	41	35	00	37	572
574 576	12	(D)	117	95	0	95				57 57	00	57 59										574 576
578 580	113	00	100	97 95	00	97	11.	8.6	15				47	00	49	41	@	41	35	00	37	578 580
582 584	165 167	00	145	97	00	99				57	00	59	49	(2)	49	41	00	43				582 584
586 588	149	00	147	97	@	97	13	88	18 75	59	@	59							37	@	37	586 588
590 592	147		92	99	@	99	- 11	- 00	75				10	000	10	41		10	01	(8)	31	590
594	-	@	10	97				0.0	-	59	0	59	49	0	49	41	0.0	43	-			592 594
596 598	187	00	100	99	00	101	75	86	13	59	00	61	49	00	51	43	(3)	43	37	0	37	596 598
600								8		10				12			14			600		

		TW	0-0	CIR	CUI	т, і	DOL	JBL	ΕV	VIN	DIN	IGS	, F	OR	DR	UM	AR	MA	TU	RES	3.	
rons							F	RON	TA	ND	BA	CK 1	PITO	CHE	S							TORS
CONDUCTORS	P	4 OLI	ES	P	6 OLI	ES	P	8 OLE	ES	P	10 OLE	ES	P	12 OLE	ES .	P	14 OLI	ES	P	16 OLE		No. OF CONDUCTORS
No, OF	F	HE-	В	F	SE-	В	F	RE+ ENTHANCE	В	F	DE- ENTRANCY	В	F	RE- ENTRANCY	В	F	SE- ESTRANCE	В	F	RE* ENTHANCS	В	-
602	125	0.0	\$61 160	101	00	101	10	88	16	59	0.0	61	49	00	51			- 10	37	00	39	602
606 608 610	181	00	101	101	00	103				61	@	61	51	@	51	43	@	43				606
612	180 153	00	155	103	0	103	H	88	H	61	0	61				43	00	45	37	00	39	612
616 618	353 150	@	105	101	00	103				61	0.0	63	51	ග	51							616
620 622 624	161	ග	169	103	00	105 103	H	3%	II	61	00	63	51	00	53	43	00	45	39	@	39_	620 622 624
626 628	150 150 157	00	191	105 103	00	105 105	H	86	79 79	63	@	63	51	0.0	53	45	0	45	39	0	39	626 628
630 632	387 109	00	100	105	00	107						200	53	@	53							630 632
634 636 638	107	00	169 161	105	8 8	105	13	8	19	63	00	63 65				45	@	45	39	0.0	41	634 636 638
640 642	161	0	169 161	105	0.0	107							53	@	53	45	00	47				640
644	360 181	00	101	107 107	00 00	109 107	19	83	100	63 65	00 (D)	65 65	53	00	55	44		15	39	00	41	644
648 650 652	161	00	100	109	(Q)	109 109	11	88	8				53	00	55	45	00	47	41	0	41	648 650 652
654 656	152	0	100	109	00	111				65 65	@ ••	65 67	55	(3)	55	47	ග	47				654 656
658 660 662	168	00	1/00 1/02	109	@	109	3	88	65							47	@	47	41	0	41	658 660 662
664	禁	@	166 197	109	00	111				65 67	00	67 67	55	@	55	11	00	41				664
668 670	100	00	107	111 111	00 (2)	113 111	11	8	101				55	00	57	47	00	49	41	00	43	668 670
672 674 676	100	@	100	113 111	00	113	12	86	22	67	@	67				177		40	41		10	672 674 676
678 680	100	00 (B)	173 179 113	113	00	115	-	90	-15	01	00	00	55	00 (D)	57	47_	00	49	41	0.0	,43	678
682	183	0.0	111 179	113	(0)	113	8	88	90 FE	67	00	69				49	0	49	43	@	43	682 684
686 688 690	177 178	@	174	113	00	115 115				69	@	69	57	@	57	49	0	49				686 688 690
692 694	111	115 @ 11				117 115	iii	86	#5 #2	69	<b>@</b>	69	57	00	59	20		24	43	00	43	692 694
696 698 700	188	117 @ 11					5	88	C.	69	00	71	57	00	50	49	00	51	49	00	45	696 698
100	113	4	167	110	6	111	AT.	8	- 20	10				12	59		14		43	700		

		TW	0-0	CIRC	CUI	т, г	001	JBL	E V	VIN	DIN	ıcs	, F(	OR I	DR	UM	AR	MA	TU	RES	;.	
TORS							F	RON	TA	ND	BAG	CK I	PITC	HE	S		ni i					TORS
No. OF CONDUCTORS	P	4 OLE	ES	P	6 OLI	ES	P	8 OLE	S	P	10 OLE	S	P	12 OLE	ES	P	14 OLE	S	P	16 OLE	S	No. OF CONDUCTORS
- Contract of the last of the	F	RE- ENTRANGE	В	F	PE-	В	F	RE- ENTRINEY	В	F	AE+ ENTRANCY	В	F	HE- ENTRANCY	В	F	RE-	В	F	RE" ENTRANCY	В	_
702	\$10 \$77	ග	175 177	117	00	119	1			69	00	71	59	0	59	49	00	51				702
706 708	186	0.0	\$77 \$70	117	0	117	17 00 00	88	10 30	71	0	71						2	43	00	45	706 708
710				119	0	119	-00	100	- 30				F-0			51	0	51	20		20	710
712 714	471	@	48	117	0.0	119				71	@	71	59	0	59			2				712
716	31t 170	0.0	101	119 119	(D)	121	B	88	- 11	71	00	73	59_	00	61	51	0	51	45_	0	45	716
720 722	48	GD.	177	121	0	121				-												720 722
724 726	485	0.0	110 260	119	00	121	%	83	OL M	71 73	00 (D)	73 73	59	00	61	51	00	53	45	0	45	724 726
728	101	0	315 710	121	00	123					-00	10	61	0	61							728
730 732	捌	0.0	160	121	(0)	121	盐	88	引							51	00	53	45	00	47	730 732
734 736	1 N.S. 7 (0)	(D)	188	123 121	00	128 123				73	00	78 75	61	0	61					9		734 736
738 740	110	00	18	123	00	125	45	25	81				61	00	63	53	@	53	45	00	47	738 740
742	115		180	123	0	123				73		75	-		00				-			742 744
746		00	100	125	0	125	-			75	00	75		•	40	53	0	53				746
748 750	場	00	133	123		125	93	88	12				61	00	63				47_	@	47	748 750
752 754	187	0	15	125 125	00 127 (D) 125			-57	1	75	0	75	63	0	63	53	00	55				752 754
756 758	10	00	100	127			91	88	tt	75_	00	77							47	0	47	756
760	300	0	191	125	00	127							63	00	63	53	00	55				760
762 764	149 101	0.0	191	127	00	129	90	88	3)	75	00	77	63	0.0	65				47	00	49	762 764
766 768	100	0	100 100	127	0	127				77	00	77				55_	@	55				766 768
770	40	00	101	129 127	00	129 129	96 97	88	Ş.				63	00	65				47	00	49	770
774 776	173	(3)	199	129	00	131	71		Man	77	@	77	65		65	55	0	55				774 776
778				129	(3)	129	-	7.66	-	1.1	0.0	10	00	@	.00	-		-				778
780 782	193	00	100	131	0	131	Ħ	88	19							55	00	57	49	0	49_	780 782
784 786	107	@	100	129	00	131				77	00	79	65	0	65						1	784 786
788 790	107 107	00	185	131	00 (D)	133 131	N.	88	20				65	00	67	55	00	57	49	00	49	788 790
792 794	197	00	107	133		-				70	(5)	200					28					792
796 798	11/17	00	109	131	00	133	55	88	损	79	00	79 81	65	00	67	57	(D)	57	49	00	51	794 796
800						135																798 800
		4			6			8		10				12			14					





WINDING TABLES FOR TWO-CIRCUIT, TRIPLE WINDINGS FOR DRUM ARMATURES.

# TABLE OF TWO-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF 10+ HE-F F B F B B B F B B No. 野 計 H (00) 苦 (20) (00) (00) @ 器 (00) 甚 H 120 No. (00) (00) 뫱 (00) (00) 33 발 풮 (00) ñ (00) 첉 型 (00) (00) @ (00) H SER! 社 (22) 敖 (00) (00) $\frac{146}{148}$ (00) (22) (20) (00) 器 計 먑 节 (00) (00) (00) 27 司 41 (00) 13 (22) 발 (00) 166 (22) 63 @ (00) 新 빏 @ @ -17 F 持 (00) 祖 (00) $\frac{178}{180}$ 計 報 @ $\frac{182}{184}$ (00) (00) 教 188 (22) (00) (00) 計 32 (00) (00) 益 数 (00) 計 #



# TABLE OF TWO-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES No.OF OF B F F F F B B B F B B B 54 (00) (00) 경 H B (00) (00) (00) $\frac{210}{212}$ 38 इस्स 50. (20) 計 (RD) (00) $\frac{220}{222}$ 000 3W $\frac{224}{226}$ (1) @ (00) $\frac{228}{230}$ 提 @ (00) (00) (00) 23 H 20° NI NI 1 533 (00) (00) $\frac{238}{240}$ 쫛 $\frac{242}{244}$ (00) (00) $\frac{246}{248}$ SFR. (00) 48 (00) 양 900 E (00) (II) (22) (00) (00) E 988 (00) (00) F 档 な 71 (00) Til 碧 FFS. 55 (00) $\frac{286}{288}$ (00) B (00) (00) (00) (00) 296 (00) (00) 21 000

	BL	E C	FT	W	o-c	IRC				OF COLUMNS		Contraction of the last of the	2012000	STITUTE S		R DI	RUN	A N	RM.	ATU	JRE	
TOR							F	RON	T A	ND	BA	CK I	PITC	HE	S			-			-	TOR
CONDUCTORS	P	4 OLE	ES	P	6 OLE	ES	P	8 OLE	S	P	10 OLE	ES	P	12 OLE	ES	P	14 OLI	ES	P	16 OLE	S	CONDUCTORS
No.0F	F	RE- EATHART	В	F	RE+ ENTRARCE	В	F	Ag- ENTRAGE	В	F	HE- ENTRANCY	В	F	RE- ENTRANCY	В	F	HE- ENTHARCH	В	F	PE- ENTRANCY	В	No.0F
302 304	17	@	静				37	00	37	31	(2)	31				21	@	23		-		302
306	75	000	12 79	22	(00)	11	39	000	39	29	000	31	8	92	87							306
308	Tis Tiv	(00)	16				37	00	39										19	@	19	308
312	37 70	(0)	77	13	GEO.	12	39	@	41	31	00	33				21	@	23	19	@	21	312
316 318		000	100	15	557	12	39	000	39	31	@	31	- 15	æ	11	23	00	23				316 318
320	211			01	900	30	-		41				97	000	47							320
322	10 10 10	(00)	II.	27	00	II	41	@		33	000	33					-					322
326 328	19	(30)	23				39	(00)	41	31	(00)	33				23	(D)	23	19	00	21	326 328
330	80	000	10	65	620	#	41	000	43				\$ E	653	易	23	000	25	21	000	21	330
334 336	91. 50	@	65 85	- 6 <u>6</u>	8778	No.	41	00	41	33	@	35										334
338	83 80	@	80 51		6176	17	43	@	43	00	000	00										336
340 342	101 17	000	50 71	赞	00	10	41	000	43				H	88	22	23	000	25	21	000	21	340
344	新	@	80 80				43	(00)	45	35	(0)	35				25	@	25	21	(00)	23	344
348 350	50	(20)	62 80	95	500	17	43	(00)	43													348
352					655	50	-	-	material con-	0.8		07	-00	- Co	- 69							352 354
354 356	17	000	- VI	00 00	हम्म <u>ु</u>	故	45	000	45	35 35	(90)	37	- 第	883	Tin .	25	00	25				356
358 360	- 55	@	77	10	(00)	- 17	43	00	45							25	@	27	21	00	23	358
362 364	77	00	W				45	000	47	37	(20)	37							23	00	23	362
366 368	81	000	10	號	800	8	45	000	45	35	000	37	20	833	11							366
370	93	00	93 103		- Indian	-	47	00	47							25	00	27				370
372 374	15 50	(92)	90 Mo	4	55%	01 01	45	00	47	37	00	39				27	000	27	23	(1)	23	372 374
376 378	99 90	000	- 0	15	@	12	47	000	49	37	00	37	20	88	8				23	000	25	376 378
380 382	WI VS	(0)	W)				47	00	47				-									380
384				64 60	888	00 00	-			39	000	39				27	000	27				384
386 388	10	00	\$0 \$0 \$0				49	(0)	49	37	@	39				27	(2)	29				386
390	100 100	000	77	92	888	65 - 67	47	000	49				81 88	883	103 203		50		23	000	25	390
394 396	W.	00	- 201	GB 57	@	10 to	49	(00)	51	39	000	41 39					7		25	(00)	25	394 396
398 400	203	@	99 101	ur	350	61	49	00	49	00	500	00				27	@	29				398
400		4			6			8			10			12		29	14	29		16		400

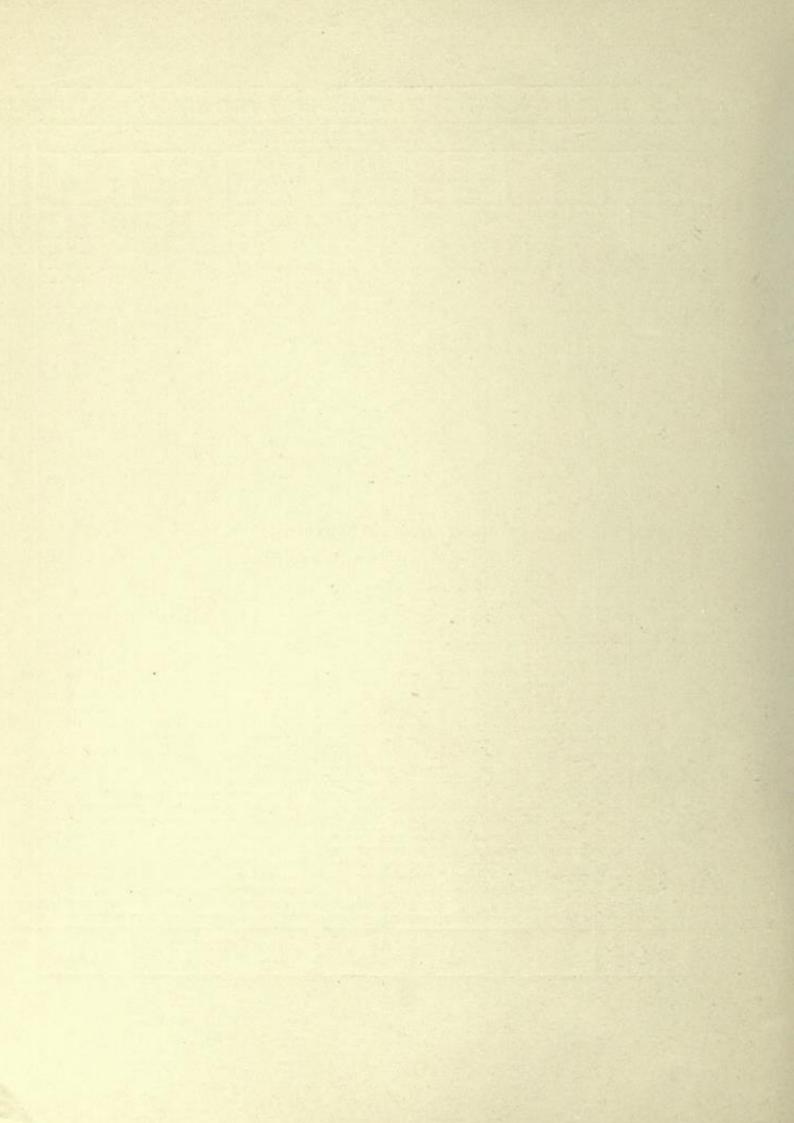
	BL	EO	FT	W	o-c	IRC	רוט	г, т	RIP	LE	WII	NDI	NG	S, F	OF	D	RUN	ИΑ	RM	ATL	JRE	
TORS							F	RON	TA	ND	BAG	CK I	PITC	HE	S							TORS
CONDUCTORS	P	4 OLE	ES	P	6 OLE	ES	P	8 OLE	S	P	10 OLE	ES	P	12 OLE	ES	P	14 OLE	cs	P	16 OLE	S	CONDUCTORS
No.OF	F	HS- ENTRAHCY	В	F	RE- ENTRARCY	В	F	HE- ENTRANCY	В	F	RE- ENTRANCT	В	F	PE- ENTRANCY	В	F	AH- ENTRANCY	В	F	ME- ENTRANCE	В	No. OF
402	iii	000	98 100	St.	000	GI AN	51	000	51	41	@	41	25 33 .	000	12							402
406	96 108	@	101	65	563	61	49	@	51	39	@	41							25	@	25	406
410 412	101	00	102			-	51	@	53							29	@	29	25	00	27	410
414 416	134	000	102	85	(1)	%	51	000	51	41	000	43 41	754 50	223	35	29	000	31				414 416
418 420	108 100	@	107	93	800 Guan	背	53	@	53										200			418 420
422 424	腊	(00)	100				51	00	53	43	@	43							25	00	27	422 424
426 428	307	000	106	71	<i>१</i> ५८	n	53	000	55	41	000	43	20 10	<i>इसर</i>	37	29 31	000 (D)	31 31	2.7	000	27	426
430 432	400	@	13	1)	(00)	33	53	30	53	-												430 432
434	300	@	111	-	000		55	@	55	43	(D)	45							0.00			434
438 440 442	100	000	100	43	85	#3	53	000	55				II.	883	H	31	@	31	27	000	27	440
444 446	甜	@	118	报	883	Vi	55	@	57	45	000	45				31	@	33	27	000	29	444
448	甜	000	111	78 70	@	76 77	55	000	57	43	@	45	12	ge de	17							446 448 450
452 454	HI	@	Ha Ha	18	- CC	-11	55	@	57	45	@	47	- 85	GI.D		31	(20)	33	27	(00)	29	452 454
456 458	111	@	419	33	850	70	57	@	59	45	000	45				33	000	33	29	(0)	29	456 458
460 462	115	000	119	35	888	75	57	000	57				1	5278	27 27					-		460 462
464	119	@	111				59	@	59	47	(Q)	47										464 466
468 470	118	@	HI	11.	@	70	57	@	59							33	000	33	29	@	29	468 470
472	111	000	UK	17	900	ij	59	000	61	47	000	49	75	000	77				29	000	31	472
476 478 480	出	@	411	29 88	क्र	100	59	@	59	47	@	47										476
482 484	W	@	133	-91	000	.61	61	@	61	49	@	49				33	@	35 35				480 482 484
486 488	110 121	000	鸖	Th.	(20)	£7. E2	59	000	61	47	000	49	Al .	器	a	00	@	30	29	000	31	486
490 492	121 121	@	榋	F1 60	900	300	61	@	63										31	(QQ)	31	490 492
494 496	121 123	@	塩				61	@	61	49 49	@	51 49				35	@	35				494 496
498 500	居	000	蜡	- 13	883	163 160	63	000	63				А	<i>হন্দ</i>	13	35	000	87				498 500
		4			6			8			10			12			14			16		

E							F	RON	IT A	AND	BA	CK I	PITO	CHE	S							-
CONDUCTORS	P	4 OLI	ES	P	6 OLI	ES	P	8 OLI	ES	P	10 OLE	S	P	12 OLE	s	F	14 OLI	ES	P	16 OLI	ES	
No.OF	F	BE- ENTRARC	В	F	HE- ENTRANES	В	F	RE- ENTRANCY	В	F	RE- ENTRANCY	В	F	HE- ENTRANCY	В	F	ne- entwents	В	F	HE- ENTRANGY	В	
502 504	155 157	@	127	-	@	10	61	00	63	51	000	51				-			31	@	31	-
506	126 327	00	贈		000	10	63	00	65	49	@	51							31	@	33	I
508	135 139	000	100	12	929	NI NI	63	000	63				3	888	10	35	000	37				
512	器	@	155				65	(20)	65	51	@	53				37	@	37			-	+
16	107	00	120	10	889	W.	63	00	65	51	000	51							31	@	33	1
520				-		IF.							-	(TO)	48				33	- Control of the Cont	33	I
522 524	721	000	122	- BP	00	17	65	000	67	53	00	53	#	888	**	37	00	37	50	000	00	B
26	盟	00	111	E	000	25	65	@	65	51	000	53				37	00	39				1
530 532	lan ins	00	12				67	@	67													
534	185	000	139 136	#1 #0	589	10	65	000	67	53	000	55	#	हुन्छ इन्ह	#5 45				33	000	33	I
536 538	138	00	133				67	œ	69	53	00	53				37	@	39	33	00	35	1
40	137	@	187	141	(0)	教	67	@	67							39	000	39				13
44	186	000	125	17	223	2	69	000	69	55	(00)	55 55	99	000	40 40 40							1
48				31	G2S	15			Lucia	00	000	00	46	(CD)	47				-		-	1
550	100	00	157	12	क्र	11	67	@	69							39	000	39	33	@	35	1
554	139	(20)	111				69	(00)	71	55 55	(B)	57 55				39	(00)	41	35	@	35	
558	H	000	177	11	(2)	10	69	000	69	00	300	00	60 61	333	17							1
560	139	00	110				71	00	71	200												
564 566	123	(00)	111 143	242	820	200	69	(20)	71	57	000	57 57				39	@	41	35	@	35	R
568	10	000	111	100	853	9	71	000	73				4	84%	55	41	00	41	35	000	37	l
572				50	305	el					-		47	200	49				00	500	31	R
574 576	111	@	100	77	000	Vib Vit	71	(00)	71	57	000	59 57										1
578 580	163 146	(00)	143				73	00	73							41	00	41				
582 584	140	000	10	96 97	223	80	71	000	73	100	(55)	59	47	000	#	41	000	43	35	000	37	
586	140	@	146	-	- una		73	00	75	59	(Q) (Q)	59							37	<b>(20)</b>	37	I.
188	15	00	147	11	898	95	73	@	73							-	-					E
92 94	147	000	115	- 1	@	5/8 210	75	000	75	59	000	61	1	88	49	41	000	43				20,00
596	1/0				0.00	210	20000			59	@	59	46	0.0	101	43	000	43			-	U
98	_ISi_	(30)	121	99	920	00 191	73_	@	75										37_	@	37	0
600	_151	4	101	99	6	3/13	-13_	8	10		10			12			14			16		37

				-	CI	no	_				BA					-	101	1 //	T IVI /	ATU	NE	-
CONDUCTORS	P	4 OLE	ES	P	6 OL	ES		8 OLI			10 OLE			12 OLI		P	14 OLI	ES	P	16 OLE		COMPLICTORS
No. 0F	F	RE- ENTRANCE	В	F	HE-	В	F	HE- ENTRANGE	В	F	ENTRANCE	В	F	RE-	В	F	RE- ENTAINE	В	F	HE- ENTHANCY	В	PO CN
602	110	@	169				75	@	77	61	00	61							37	(00)	39	60
606 608	189	000	161 163	177	888	101	75	000	75	59	000	61	N	868	11	43	600	43				60
610	100	00	151 164	VIII	63	1/4	77	00	77							43	88	45				61
614	164 150	00	161 184	10	@	增	75	(00)	77	61	00	63							37	(0)	39	61
616 618	151 186	000	159	310 500	823	P100	77	000	79	61	@	61	61 21	000	81 88				39	000	39	61
320 322	161	@	102				77	@	77							43	@	45				62
624 326	154 167	@	154-	-112	888	103	79	@	79	63	000	63 63				45	000	45				62 62
328 330	154	000	157 16-9	100	(00)	109	77	000	79				31	956	18				39	000	8.9	62
332 334	提	@	141				79	000	81	63	1	65							39	@	41	63
536 538	167 161	@	157	19	200	10	79	(00)	79	63	000	63_				45	000	45 47	3			63
340 342	1/1	000	150	900 100	888	185	81	000	81				15	888	13						7	64
344	101	@	160 161 165	*/4	000	109	79	@	81	65 63	@	65 65	la_	000	- 13				39	@	41	64
548 350	161	@	105	100	00)	109	81	@	83	00	WU.	00				45	@	47	41	@	41	64
552 354				102	900	109		000	3	0.5		07	4.0	one	54.	47	@	47	11	_(QQ)_	-11	65
356	161	000	165	5/10	600	109	81		81	65 65	(Q)	67 65	10	000	20						-	65
58 60	160	@	168	100	8F8	100	83	@	83			-										66
362	127	00	122				81	@	83	67	@	67				47	@	47	41_	00	41	66
366 368	160	000	165	111	@	1111	83	000	85	65	000	67	批	555	500 51	47	000	49	41	000	43	66
370 372	18	@	127	411	200	10	83	@	83													67
574 576	101	@	111				85	@	85	67 67	@ @	69 67										67
378 380	M	000	栅	101	FFR	111	83	000	85				89	893	81	47	000	49 49	41	000	43	68
582 584	100	@	167	199	@	H	85	@	87	69	000	69				10	788/		43_	@	43	68
386 588	100	00)	M	110	000	110	85	@	85	67	@	69										68
390 392	m	000	12)	110 110	800	313	87	000	87				ţi.	9.00	57 50	10	65	40				69
594	TA.	@	B	110	67	711	85	@	87	69	@	71				49	(B)	49 51	43	(00)	43	69
596 598	H	@	III .	118	88	出	87	@	89	69	000	69							43	@	45	69
700		4			6			8			10			12			14			16		70

	AB	LE (	OF '	TW	0-0	IR	CUI	т, т	RII	PLE	W	ND	INC	as,	FO	R D	RU	M A	RN	TAT	UR	
roas							FI	RON	TA	ND	BAC	CK F	PITC	HE	S					- 10		TORS
CONDUCTORS	P	4 OLE	ES	P	6 OLE	S	P	8 OLE	S	P	10 OLE	S	P	12 OLE	ES	P	14 OLE	ES	P	16 OLE	S	CONDUCTORS
No. 0R	F	ME- ENTRANCY	В	F	NE-	В	F	AE- VURAFTES	В	F	NE" ENTEANCY	В	F	RE- BATRANCY	В	F	HE- ENTINHEY	В	F	RE- ENTRANGS	В	No.OF
702	辨	000	1/16 177	耕	000	137	87	000	87	71		71	10	233	15			-	-			702
704 706	177	@	125 115				89	00	89	69	@	71 71				49	00	51				706
708 710	lit.	@	400	486	883	117	87	@	89							51	000	51	43	@	45	708
712 714	110	000	311	111	STR	뱀	89	000	91	71	000	73	42	FFR.	69				45	000	45	712
716		-		1119		- 123				71	@	71		000	- 01				10		20	716
718 720	371	@	tal .	100	00	1117	89	(00)	89							51	000	51				718 720
722 724	48	00	575 161				91	00	91	78	@	73				51	@	58	-			722 724
726 728	455	000	個	那	833	122	89	000	91	71	000	73	60	883	85				45	000	45	726 728
730 732	145	00	100	相	85%	103	91	00	93										45	@	47	730
734	415	@	180	7115	200	161	91	@	91	73	00	75				51	000	53				734
736 738	107	000	165	19	00	121	93	000	93	78	00	73	8	966	81 68	53	(D)	53				736
740	12	@	100	-			91	(D)	93										45	@	47	740
744 746	123		160	18	85	10	93		95	75 73	000	75 75							47	and the	47	746
748		@	10000	-01	-	-		(0)		10	00	10				53	@	53	41	@	41	748
750 752	125	000	100	瑞	888	127	93	000	93				- #i	883	53	53	000	55				750
754 756	100	00	101	195 127	@	125	95	(00)	95	75 75	000	77 75										754
758	101	(Q)	贈	117	000	£37	93	00	95	-		10							47	00	47	758
760 762	1sy 3%	000	160	105 200	600	157	95	000	97				(1)	929	61	53	000	55	47	000	49	769
764 766	189	@	191				95	@	95	77	(00)	77	-			55	(20)	-55				764
768 770	191	(00)	122	15	888	- 183	97	(2)	97									-				768
772 774	101	000	105	117	(60)	109	95	000	97	77	000	79	04	ONO	46				4"	000	40	77
776			Marie II	177	(00)	109				77	(22)	77	65	355	322	55	@	55	47	.000	49	774
778 780	100	@	301	179 103	900	137	97	(00)	99							55	(Q)	57	49	@	49	778 780
782 784	19	@	496 196				97	@	97	79	@	79										782 784
786 788	107	000	-15h -190	13	मर	120	99	000	99	77	000	79	65	878	69							786
790	100	00	197 319				97	@	99			-				55	@	57	49	@	49	788 790
792 794	101	(III)	197 201	133	@	123	99	00	101	79	(00)	81				57	000	57	49	00	51	792 794
79€ 798	107	000	190	115	000	103	99	000	99	79	@	79	65	888	36					-		796
800			-																-			800
		4			6			8			10			12			14			16		





WINDING TABLES FOR MULTIPLE-CIRCUIT, SINGLE WINDINGS FOR DRUM ARMATURES.



I	MULT	ripli	E-CIF	RCUI	T, SI	NGL	E WII	NDIN	GS, I	FOR	DRU	JM A	RMA	TURI	ES.
OBS					FRO	NT A	ND B	ACK :	PITCE	HES		TE T			ORS
CONDUCTORS	I would	4 LES	Pol	ES	Marine and Co.	8 LES	POI	0 LES	100	2 LES		4 LES	11	6 LES	No, OF CONDUCTORS
No, OF	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. OF
202	49	51	33	35	25	27	19	21	15	17	13	15	11	13	202
204	49 51	51	33	35	25 25	27	19	21 21	15	17 19	13	15 15	11	13	204
206	51	53	33	35	25	27	19	21	17	19	13	15	11	13	208
210	51	53	33	35	25	27	19	21	17	19	13	15	13	15	210
212	51	53	35	37	25	27	21	23	17	19	15	17	13	15	212
214 216	53	55 55	35	37	25 25	27	21 21	23	17	19 19	15 15	17	13	15	214
218	53	55	35	37	27	29	21	23	17	19	15	17	13	15	218
220	53	55	35	37	27	29	21	23	17	19	15	17	13	15	220
222	55	57	35	37	27	29	21	23	17	19	15	17	13	15	222
224 226	55	57 57	37 37	39 39	27 27	29	21	23 23	17	19 19	15 15	17	13	15	224
228	55	57	37	39	27	29	21	23	17	19	15	17	13	15	228
230	57	59	37	39	27_	29	21	23	19	21	15	17	13	15	230
232	57	59	37	39	27	29	23	25	19	21	15	17	13	15	232
234	57	59 59	37	39 41	29	31	23 23	25 25	19	21 21	15 15	17	13	15 15	234
238	39	61	39	41	29	31	23	25	19	21	15	17	13	15	238
240	59	61	39	41	29	31	23 -	25	19	21	17	19	13	15	240
242	59	61	39	41	29	31	23	25	19	21	17	19	15	17	242
244 246	59 61	61	39	41	29 29	31	23 23	25 25	19	21 21	17	19	15	17	244 246
248	61	63	41	43	29	31	23	25	19	21	17	19	15	17	248
250	61	63	41	43	31	33	23	25	19	21	17	19	15	17	250
252	61	63	41	43	31	33	25	27	19	21	17	19	15	17	252
254 256	63	65 65	41	43	31	33	25 25	27	21 21	23 23	17	19	15	17 17	254 256
258	63	65	41	43	31	33	25	27	21	23	17	19	15	17	258
260	63	65	43	45	31	33	25	27	21	23	17	19	15	17	260
262	65	67	43	45	31	33	25	27	21	23	17_	_19_	15	17	262
264	65	67	43	45	31	33 35	25 25	27 27	21	23 23	17	19	15	17	264 266
268	65	67	43	45	33	35	25	27	21	23	17	21	15 15	17	268
270	67	69	43	45	33	35	25	27	21	23	19	21	15	17	270
272	67	69	45	47	33	35	27	29	21	23	19	21	15	17	272
274 276	67	69	45	47	33	35 35	27	29	21	23	19	21	17	19	274
278	69	69 71	45	47	33	35	27 27	29 29	21 23	23 25	19	21 21	17	19	276 278
280	69	71	45	47	33	35	27	29	23	25	19	21	17	19	280
282	69	71	45	47	35	37	27	29	23	25	19	21	17	19	282
284 286	69 71	71 73	47	49	35 35	37 37	27	29	23	25	19	21	17	19	284
288	71	73	47	49	35	37	27 27	29 29	23	25 25	19 19	21 21	17	19 19	286 288
290	71	73	47	49	35	37	27	29	23	25	19	21	17	19_	290
292	7.1	73	47	49	35	37	29	31	23	25	19	21	17	19	292
294	73	75	47	49	35	37	29	31	23	25	19	21	17	19	294
296 298	73 73	75 75	49	51	35	37	29 29	31	23	25 25	21	23 23	17	19	296
300	73	75	49	51	37	39	29	31	23	25	21	23	17	19	300

### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES. POLES POLES POLES POLES POLES POLES POLES OF F F F F F B F F B B B B B В d N 79 51 $\frac{27}{27}$ 19 21 $\frac{21}{21}$ 338 21 $\frac{21}{21}$ 37 37 35 29 $\frac{21}{21}$ 23 $\frac{25}{25}$

### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF OF F F F B F B F B F B F B B B S. 27 $\frac{25}{25}$ 418 $\frac{71}{71}$ 27 71 71 71 55 $\frac{27}{27}$ 73 75 75 55 73 111 $\frac{448}{450}$ 73 $\frac{27}{27}$ 75 $\frac{448}{450}$ $\frac{470}{472}$ 79 480

### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF OF F F F B F В F B F B B B F B No No 546 $\frac{546}{548}$ $\frac{71}{71}$ $\frac{71}{73}$ 71 73 75 73 75 75 75 51 41

N	NULT	IPLE	E-CIF	CUI	T, SII	NGLE	a wii	NDIN	GS, I	FOR	DRU	M AR	MAT	URE	s.
TORS					FRO	NT A	ND E	BACK	PITCI	HES					TORS
CONDUCTORS	4		(	Distriction of	The second	3	- metrico	0	_1	2000	1000	4	100000000000000000000000000000000000000	6	OF CONDUCTORS
0F CO!	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	F C03
o Z	F	В	F	В	F	В	F	В	F	В	F	В	F	В	g Z
602 604	149 149	151 151	99	101 101	75 75	77	59 59	61 61	49	51 51	41 43	43 45	37 37	39 39	602
606	151 151	153 153	99	101	75 75	77	59 50	61	49	51 51	43	45	37	39	606
610 612	151 151	153 153	101 101	103 103	75 75	77	59 61	61	49	51 51	43	45 45	37 37	39 39	610 612
614 616	153 153	155 155	101 101	103	75 75	77	61 61	63 63	51 51	53 53	43	45 45	37 37	39 39	614 616
618	153 153	155	101	103	77	79	61	63	51	53	43	45	37	39	618
620 622	155	155 157	103	105	77	79 79	61	63	51 51	53 53	43	45 45	37	39	620
624 626	155 155	157 157	103	105 105	77	79 79	61	63 63	51 51	53 53	43	45 45	37	39 41	624 626
628 630	155 157	157 159	103 103	105 105	77	79 79	61	63 63	51 51	53 53	43	45 45	39	41 41	628
632 634	157 157	159 159	105 105	107 107	77 79	79 81	63	65 65	51 51	53 53	45 45	47	39	41	632 634
636 638	157 159	159 161	105 105	107 107	79	81 81	63	65 65	51 53	53 55	45 45	47	39 39	41	636 638
640	159 159	161	105	107	79	81	68	65	53	55	45	47	39	41	640
642	159	161	107	109	79 79	81	63	65 65	53 53	55 55	45 45	47	39	41	642 644
646 648	161 161	163 163	107	109	79 79	81 81	63 63	65 65	53 53	55 55	45 45	47	39	41	646
650 652	161 161	163 163	107	109	81	88 83	63	65	53	55 55	45	47	39	41	650
654 656	163 163	165 165	107	109	81 81	83 83	65 65	67 67	53 53	55 55	45 45	47 47	39 39	41	654 656
658 660	163 163	165 165	109	111 111	81 81	83 83	65	67 67	53	55 55	45	47	41	43	658
662	165 165	167 167	109	111	81	83	65	67	55	57	47	49	41	43	660
666	165	167	109	111	83	85	65 65	67 67	55 55	57 57	47	49	41	43	664 666
668 670	165 167	167 169	111	113 113	83 83	85 85	65 65	67	55 55	57 57	47	49	41	43	668
672 674	167 167	169 169	111	113 113	83 83	85 85	67	69	55 55	57	47	49	41	43	672
676 678	167 169	169	111	113	83 83	85 85	67	69 69	55 55	57 57	47	49	41	43	676 678
680 682	169 169	171 171	113 113	115 115	83 85	85 87	67 67	69 69	55 55	57	47	49	41	43	680 682
684	169 171	171 173	113	115 115	85 85	87	67	69	55	57	47	49	41	43	684
688	171	173	113	115	85	87 87	67 67	69 69	57 57	59 59	47	49 51	41	43	686 688
690 692	171 171	178 173	113 115	115 117	85 85	87 87	69	69 71	57 57	59 - 59	49	51 51	43	45 45	690
694	173 173	175 175	115 115	117 117	85 85	87 87	69	71 71	57 57	59 59	49	51 51	43	45 45	694 696
698 700	173 173	175 175	115 115	117 117	87 87	89 89	69 69	71 71	57 57	59 59	49	51 51	43	45 45	698 700
	-		V. C.C.			-			1001	00	2.0	0.1	247	10	1011



### MULTIPLE-CIRCUIT SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF F B F B F F B F B F B F B B è N 59 $\frac{71}{71}$ $\frac{211}{211}$ 71 71 69 73 71 71 71 73 55 876 75 75 73 878 87 147 75 $\frac{223}{223}$ 75 73

# MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF OF F F F F F B F B F B B B В B og. 111 113 113 $\frac{227}{227}$ $\frac{149}{151}$ 115 77 $\frac{151}{151}$ 67 115 233 157 237 $\frac{157}{157}$ 119 95 79 59 237 $\frac{239}{239}$ 159 79 119 $\frac{121}{121}$ 95 $\frac{241}{241}$ $\frac{243}{243}$ $\frac{123}{123}$ 71 71 $\frac{972}{974}$ $\frac{165}{165}$ $\frac{121}{121}$ 71 81 71 $\frac{73}{73}$ 71 GI

### MULTIPLE CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF F F B F В F F B F В F В B В 71 71 71 171 $\frac{171}{171}$ $\frac{127}{127}$ 73 75 $\frac{1024}{1026}$ 75 $\frac{173}{173}$ 75 $\frac{177}{177}$ 75 77 77 $\frac{177}{177}$ $\frac{179}{179}$ $\frac{105}{105}$ $\frac{107}{107}$ 75 77 77 77 79 $\frac{273}{273}$

Above choice of Pitches will prove most satisfactory, although, as stated in text, the absolute magnitude of average pitch may be varied within reasonable limits.



# MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF F B F B F B F B F B F В F B No. 139 79 $\frac{71}{71}$ $\frac{277}{277}$ $\frac{277}{279}$ $\frac{279}{281}$ 141 187 $\frac{141}{141}$ 111 113 113 $\frac{1120}{1122}$ 71 79 143 $\frac{1134}{1136}$ $\frac{115}{115}$ 115 $\frac{143}{145}$ 71 73 $\frac{115}{115}$ 145 71 $\frac{1156}{1158}$ 291 $\frac{1166}{1168}$ 193 $\frac{195}{195}$ $\frac{145}{145}$ 115 99 73 $\frac{147}{147}$ $\frac{117}{117}$ $\frac{1166}{1168}$ $\frac{1170}{1172}$ $\frac{145}{147}$ 149 149 117 119 119 $\frac{117}{117}$ $\frac{1180}{1182}$ $\frac{75}{75}$ $\frac{295}{297}$ $\frac{297}{299}$ 119 73 $\frac{1194}{1196}$ 75

### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF F B F B F B F B F B F F В В $\frac{1206}{1208}$ 77 75 $\frac{1214}{1216}$ $\frac{1220}{1222}$ $\frac{151}{151}$ $\frac{153}{153}$ $\frac{123}{123}$ $\frac{121}{121}$ $\frac{101}{101}$ 75 155 79 $\frac{1252}{1254}$ $\frac{127}{127}$ 79 $\frac{1272}{1274}$ $\frac{107}{107}$

### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTOR POLES POLES POLES POLES POLES POLES POLES OF SP F F F F F F B B B B B B F B 219 221 $\frac{133}{133}$ $\frac{219}{219}$ 111 $\frac{1328}{1330}$ 113 115 $\frac{137}{137}$ 115 113 229 231 231 231 137 $\frac{173}{173}$ 137 139 171 $\frac{115}{117}$ $\frac{231}{231}$ 139 101 $\frac{231}{233}$

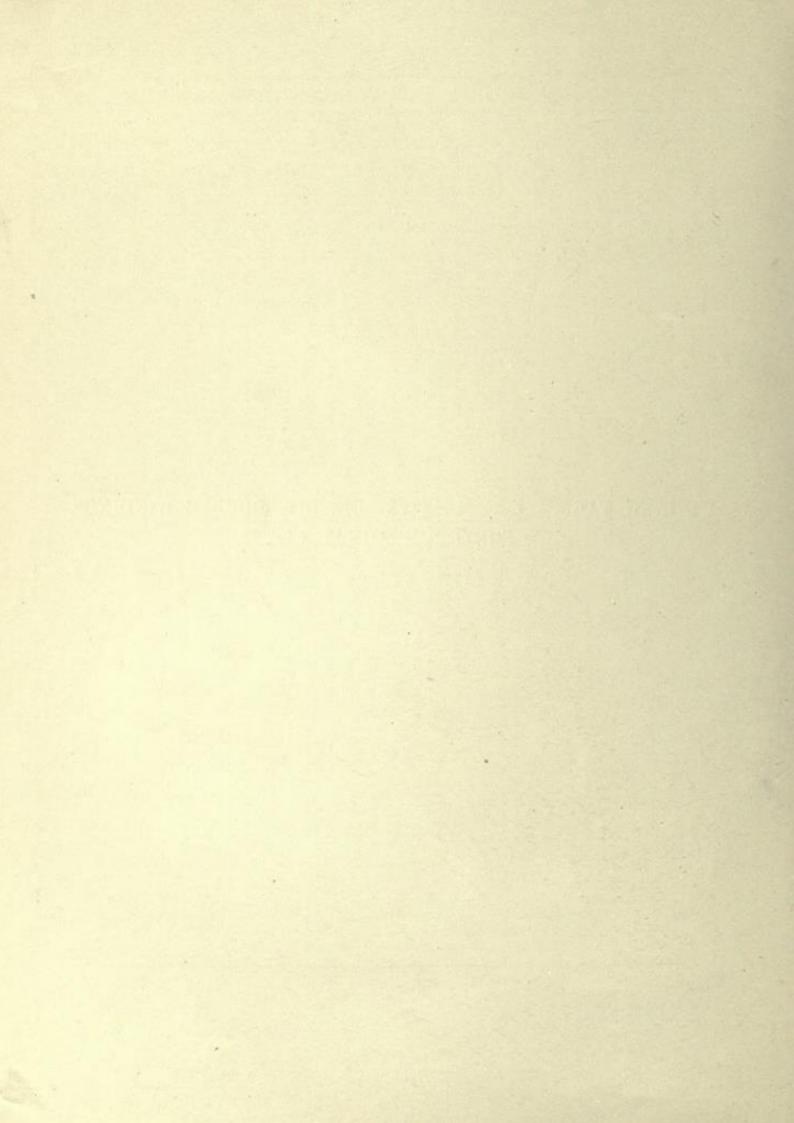
### MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF OF F B F B F B F B F В F B F B S. $\frac{175}{175}$ $\frac{177}{177}$ $\frac{175}{175}$ $\frac{139}{141}$ $\frac{141}{143}$ 117 $\frac{177}{177}$ $\frac{235}{235}$ 237 237 $\frac{1420}{1422}$ 143 177 $\frac{179}{179}$ 141 $\frac{1420}{1422}$ 143 $\frac{1444}{1446}$ $\frac{119}{119}$ $\frac{1450}{1452}$ 1486 373 $\frac{249}{249}$ $\frac{247}{247}$ $\frac{149}{149}$



# MULTIPLE-CIRCUIT, SINGLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES POLES POLES POLES POLES POLES POLES POLES OF F B F B F F F F B F B B B B $\frac{149}{149}$ $\frac{251}{251}$ $\frac{153}{153}$ $\frac{253}{253}$ 189 151 $\frac{125}{125}$ $\frac{127}{127}$ $\frac{153}{158}$ $\frac{1552}{1554}$ $\frac{155}{155}$ $\frac{129}{129}$ 131 113 $\frac{1554}{1556}$ $\frac{129}{129}$ $\frac{261}{263}$ 197 $\frac{155}{155}$ 199 $\frac{133}{133}$ $\frac{157}{157}$ $\frac{159}{159}$ 113 131

Above choice of Pitches will prove most satisfactory, although, as stated in text, the absolute magnitude of average pitch may be varied within reasonable limits

WINDING TABLES FOR MULTIPLE-CIRCUIT, DOUBLE WINDINGS FOR DRUM ARMATURES.



# MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. FRONT AND BACK PITCHES CONDUCTORS CONDUCTORS RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F F F B F F F B F B B -B B B o'N 23 13 17 $\frac{218}{220}$ 29 13 $\frac{17}{17}$ 11 19 $\frac{13}{13}$ (3) (0) 15 (0) 0.0 (3) 27 27 $\frac{23}{23}$ 13 15 19 (0) $\frac{27}{27}$ 23 19 15 21 21 $\frac{25}{25}$ 25 19 $\frac{21}{21}$ 25 19 $\frac{21}{23}$ 15 (0) (0) 71

Above choice of Pitches will prove most satisfactory, although, as stated in text, the absolute magnitude of average pitch may be varied within reasonable limits.

23

 $\frac{27}{27}$ 

19

17

 $\frac{21}{21}$ 

73 73



### MULTIPLE-CIRCUIT, DOUBLE WINDING, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F F F B F F F В B F В B B В So. 0.0 79 75 27 27 75 77 53 29 23 21 $\frac{25}{25}$ 312 21 81 77 79 17 19 $\frac{21}{23}$ 51 $\frac{21}{21}$ $\frac{25}{25}$ 57 53 19 $\frac{79}{79}$ $\frac{27}{27}$ 19 85 23 21 23 $\frac{27}{27}$ 21 0.0 $\frac{25}{25}$ 47 47 93 29 $\frac{25}{25}$ 63 29 $\frac{25}{25}$ 0.0 21 21 $\frac{25}{25}$ 29 $\frac{25}{25}$ 29 (0) 97 97 65 69 47 47 51 23 41

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F F B F B F B F B F B B F B 27 43- $\frac{27}{27}$ $\frac{103}{103}$ 71 71 103 51 73 29 $\frac{25}{25}$ 107 71 75 29 29 109 113 73 77 53 29 35 113 $\frac{115}{117}$ 75 79 79 27 27 119 $\frac{115}{115}$ 75 57

Above choice of Pitches will prove most satisfactory, although, as stated in text, the absolute magnitude of average pitch may be varied within reasonable limits,

# MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES.

1	ORS				FI	RONT	'ANI	BAC	CK PI	TCHI	ES					TORS
RE-ENTRANCY	CONDUCTORS	100	4		3	1 23	3	100000000000000000000000000000000000000	0		2		4		6	No. OF CONDUCTORS
E-EN	0F CO		LES		LES		LES		LES		LES		LES		LES	8
	No. 0F	F	В	F	В	F	В	F	В	F	В	F	В	F	В	The second second
(3)	502	123 123	127	81	85 85	61	65	49	53	39	43	33	37	29	33	502 504
(D)	506	125	129	83	87	61	65	49	53	41	45	35	39	29	33	506
00	508	125	129	83	87	61	65	49	53	41	45	35	39	29	33	508
0	510	125	129	83	87	61	65	49	53	41	45	35	39	29	33	510
00	512	125	129	83	87	61	65	49_	53	41	45	35	39	29	33	512
0	514	127	131	83	87	63	67	49	53	41	45	35	39	31	35	514
00	516	127	131 131	83 85	87	63	67	49	53 53	41	45 45	35 35	39	31	35 35	516 518
00	518 520	127 127	131	85	89	63	67	49	53	41	45	35	39	31	35	520
0	522	129	133	85	89	63	67	51	55	41	45	35	39	31	35	522
00	524	129	133	85	89	63	67	51	55	41	45	35	39	31	35	524
0	526	129	133	85	89	63	67	51	55	41	* 45	35	39	31	35	526
00	528	129	133	85	89	63	67	51	55	41	45	35	39	31	35 35	528
00	530 532	131	135 135	87	91	65	69	51	55 55	43	47	35 35	39	31	35	530 532
@	534	131	135_	87	91	65	69	51	55	43	47	37	41	31	35	534
00	536	131	135	87	91	65	69	51	55	43	47	37	41	31	35	536
(0)	538	133	137	87	91	65	69	51	55	43	47	37_	41	31	35	538
00	540	133	137	87	91	65	69	51	55	43	47	37	41	31	35	540
0	542	133	137	89	93	65	69	53	57	43	47	37	41	31_	35_	542
00	544	133 135	137	89 89	93	65	69	53	57	43	47	37	41	31	35_ 37	544
00	546 548	135	139	89	93	67	71	53	57	43	47	37	41	33	37	548
(0)	550	135	139	89	93	67	71	53	57	43	47	37	41	33	37	550
00	552	135	139	89	93	67	71	53	57	43	47	37	41	33	37	552
@	554	137	141	91	95	67	71	53	57	45	49	37	41	33	37	554
00	556	137	141	91	95	67	71	53	57	45	49	37	41	33	37	556
(0)	558 560	137	141	91	95	67	71	53	57	45 45	49	37	41	33	37	558 560
00	562	139	143	91	95 95	69	71 73	53 55	59	45	49	39	43	33	37	562
00	564	139	143	91	95	69	73	55	59	45	49	39	43	33	37	564
0	566	139	143	93	97	69	73	55	59	45	49	39	43	33	37	566
00	568	139	143	93	97	69	73	55	59	45	49	39	43_	33	37	568
@	570	141	145	93	97	69	73	55	59	45	49	39	43	33	37	570
00	572	141	145	93	97	69	73	55	59	45	49	39	43	33	37	572
00	574 576	141	145 145	93	97	69	73 73	55 55	59 59	45	49	39	43	33	37	574 576
0	578	143	147	95	99	71	75	55	59	47	51	39	43	35	39	578
00	580	143	147	95	99	71	75	55	59	47	51	39	43	35	39	580
0	582	143	147	95	99	71	75	57	61	47	51	39	43	35	39	582
00	584	143	147	95	99	71	75	57	61	47	51	39	43	35	39	584
@	586	145	149	95	99	71	75	57	61	47	51	39	43	35	39	586
00 @	588 590	145 145	149 149	95	99	71 71	75 75	57 57	61	47	51 51	39 41	43	35 35	39	588 590
00	592	145	149	97	101	71	75	57	61	47	51	41	45	35	39	592
0	594	147	151	97	101	73	77	57	61	47	51	41	45	35	39	594
00	596	147	151	97	101	73	77	57	61	47	51	41	45	35	39	596
00	598	147	151	97	101	73	77	57	61	47	51	41	45	35	39	598
00	600	147	151	97	101	73	77	57	61	47	51	41	45	35	39	600

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTOR REFENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F F F F B F B F F В В B В В No. ó 79 75 43 55 0.0 161 111 O 118 (0) 115 $\frac{115}{115}$ (6) 115 71 €D 71 71 113 57 $\frac{117}{119}$ 71 71



# MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES.

-	O RS				FI	RONT	ANI	BAG	CK PI	тсні	ES		-			ORS
RE-ENTRANCY	CONDUCTORS		4		6		3	000	0	13.00	2	1000	4	175	6	No. OF CONDUCTORS
PENT		Pol	LES	Po	LES	Po	LES	Po	LES	Po	LES	Pol	LES	Po	LES	A CON
82	No.OF	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No.0
00	702	173 173	177	115 115	119 119	85 85	89 89	69	73 73	57	61	49	53	41	45	702
(3)	706	175	179	115	119	87	91	69	73	57	61	49	53	43	47	706
(D)	708 710	175 175	179 179	115	119	87	91	69	73 73	57	61	49	53	43	47	708
0.0	712	175	179	117	121	87	91	69	73	57	61	49	53	43	47	712
00	714 716	177	181 181	117	121 121	87	91 91	69	73 73	57	61	49	53	43	47	714
@	718	177	181	117	121	87	91	69	73	57	61	49	53	43	47	718
00	720 722	177	181	117	121	87	91	69	73	57	61	49	53	43	47	720
00	724	179 179	183 183	119	123 123	89 89	93	71	75 75	59 59	63	49	53	43	47	722
(0)	726	179	183	119	123	89	93	71	75	59	63	49	53	43	47	726
(D)	728 730	179	183 185	119	123 123	89	93	71 71	75 75	59 59	63	49 51	53	43	47	728
00	732	181	185	119	123	89	93	71	75	59	63	51	55	43	47	732
00	734 736	181 181	185	121 121	125	89	93	71	75	59	63	51	55	43	47	734
@	738	183	185 187	121	$\frac{125}{125}$	89 91	93 95	71 71	75 75	59 59	63 63	51 51	55	43	47	736 738
00	740	183	187	121	125	91	95	71	75	59	63	51	55	4.5	49	740
00	742	183 183	187 187	121 121	125 125	91 91	95 95	73 73	77	59 59	63	51 51	55 55	45	49	742
@	746	185	189	123	127	91	95	73	77	61_	65	51	55	45	49	746
@	748 750	185 185	189 189	123 123	127 127	91	95 95	73 73	77	61	65	51	55	45	49	748
00	752	185	189	123	127	91	95	73	77	61	65	51	55 55	45	49	750 752
@	754	187	191	123	127	93	97	73	77	61	65	51_	55_	45	49	754
00	756 758	187	191 191	123 125	127 129	93	97	73 73	77	61	65	51 53	55	45	49	756 758
00	760	187	191	125	129	93	97	73	77	61	65	53	57	45	49	760
@	762 764	189 189	193 193	125 125	129 129	93	97 97	75 75	79 79	61 61	65 65	53 53	57	45	49	762
(D)	766	189	193	125	129	93	97	75	79	61	65	53	57	45	49	766
00	768	189	193	125	129	93	97	75	79	61	65	53	57	45	49	768
00	770 772	191 191	195 195	127 127	131 131	95 95	99	75 75	79 79	63	67 67	53 53	57 57	47	51	770
@	774	191	195	127	131	95	99	75	79	63	67	53	57	47	51	774
(Q)	776 778	191 193	195 197	127 127	131 131	95 95	99	75 75	79 79	63 63	67	53 53	57	47	51 51	776
00	780	193	197	127	131	95	99	75	79	63	67	53	57 57	47	51	780
@	782	193	197	129	133	95	99	77	81	63	67	53	57	47	51	782
(O)	784 786	193 195	197 199	129 129	133 133	95 97	99 101	77	81 81	63	67	53 55	57 59	47	51 51	784 786
00	788	195	199	129	133	97	101	77	81	63	67	55	59	47	51	788
00	790 792	195 195	199 199	129 129	133 133	97	101	77	81	63	67	55	59	47	51	790
@	794	195	201	131	135	97 97	101	77	81 81	63 65	67	55 55	59 59	47	51 51	792 794
0.0	796	197	201	131	135	97	101	77	81	65	69	55_	.59	47	51	796
00	798 800	197 197	201 201	131	135 135	97 97	101	77	81 81	65 65	69	55 55	59 59	47	51 51	798 800
	550	201		407	100	0.1	101		O.L	00	.00	100	00		OI	000

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTOR RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES POF F B F B F В F В F B F В F В ġ 79 53 141 0.0 73 $\frac{145}{145}$ 141, 107 53 63 7) 75 57 (3) $\frac{145}{145}$ 75

Above choice of Pitches will prove most satisfactory, although, as stated in text, the absolute magnitude of average pitch may be varied within reasonable limits.

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTORS RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F B F F F B B F B B F B F B No. 끪 $\frac{149}{149}$ 115 73 55 113 93 $\frac{117}{117}$ 75 $\frac{79}{79}$ 59 918 229 922 91 75 229 $\frac{233}{233}$ 91 $\frac{153}{153}$ 113 $\frac{235}{237}$ 159 @ @ 117 119 71 $\frac{245}{247}$ $\frac{245}{245}$ 123 $\frac{127}{127}$ 97

# MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES RE.ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F B F B F B F B F B F B F B $\frac{123}{123}$ 73 $\frac{127}{127}$ 71 75 75 $\frac{253}{253}$ $\frac{257}{257}$ 65 75 169 $\frac{173}{173}$ $\frac{127}{127}$ 131 71 71 75 75 $\frac{127}{127}$ $\frac{131}{131}$ $\frac{101}{101}$ 73 73 73 73 77 77 77 173 73 $\frac{177}{177}$ $\frac{1050}{1052}$ $\frac{73}{73}$ 63 $\frac{67}{67}$ 77 77 77 77 77 79 $\frac{175}{175}$ $\frac{179}{179}$ 91 75 0.0 177 75 $\frac{267}{267}$ 177 $\frac{137}{137}$ $\frac{105}{105}$ 75 1078 75 79 79 $\frac{275}{275}$ 111 77 273 $\frac{135}{135}$ $\frac{67}{67}$

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTOR RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F B B F B F F F B F B F B B 77 77 1108 $\frac{275}{275}$ $\frac{279}{279}$ 71 137 0.0 77 $\frac{1118}{1120}$ 141 73 $\frac{277}{279}$ $\frac{281}{283}$ $\frac{95}{95}$ 79 79 79 79 83 $\frac{1136}{1138}$ $\frac{73}{73}$ 0.0 73 97 0.0 141 0.0 75. 115 119 0.0 119 71 75 1178 $\frac{149}{149}$ 75 $\frac{1182}{1184}$ (2) 1198 1198 $\frac{147}{147}$

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTORS REFINTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F B F B F B F B F B F В F В $\frac{123}{123}$ $\frac{1218}{1220}$ $\frac{75}{75}$ 0.0 $\frac{203}{203}$ $\frac{203}{205}$ $\frac{121}{121}$ $\frac{75}{75}$ $\frac{153}{153}$ $\frac{157}{157}$ $\frac{125}{125}$ 101 $\frac{105}{105}$ 79 $\frac{1238}{1240}$ $\frac{205}{205}$ $\frac{209}{209}$ 91 $\frac{75}{75}$ 321 $\frac{211}{211}$ $\frac{125}{125}$ $\frac{105}{105}$ $\frac{215}{215}$ $\frac{157}{157}$ $\frac{129}{129}$ 89 $\frac{161}{161}$ 77 79 $\frac{213}{213}$ $\frac{159}{159}$ 1300 $\frac{323}{323}$ $\frac{327}{327}$ $\frac{215}{215}$ $\frac{219}{219}$ $\frac{161}{161}$ $\frac{165}{165}$ 79

### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTORS RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F B F В F B F B F F В В F B No. $\frac{107}{107}$ 79 111 $\frac{217}{217}$ $\frac{129}{129}$ $\frac{133}{133}$ $\frac{107}{107}$ 111 131 81 $\frac{113}{113}$ 133 137 115 $\frac{171}{171}$ $\frac{1366}{1368}$ 343 229 $\frac{169}{169}$ 173 115

 $\frac{135}{135}$ 

 $\frac{137}{137}$ 

113

117

83

 $\frac{227}{227}$ 

 $\frac{231}{231}$ 

 $\frac{169}{171}$ 

171

 $\frac{171}{173}$ 

175 175

 $\frac{175}{177}$ 

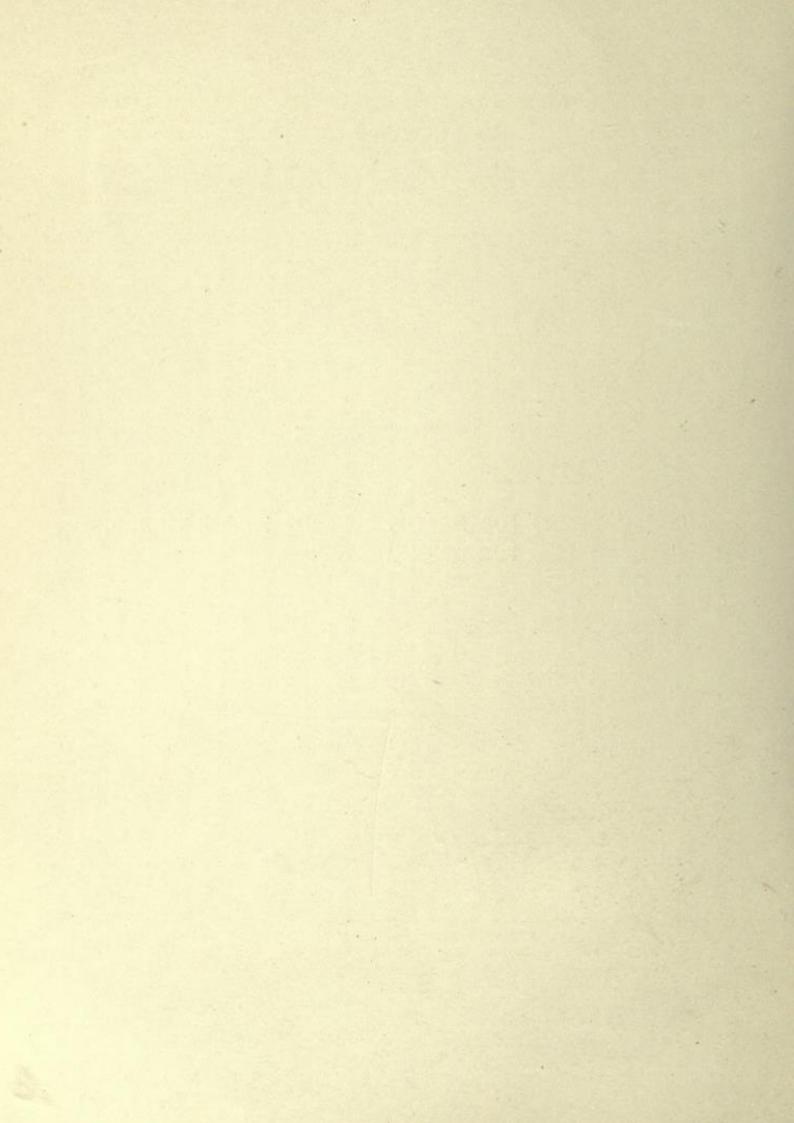
### MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES F B F B F B F B F F B F B B No. 0.0 $\frac{115}{115}$ 349 $\frac{173}{173}$ $\frac{177}{177}$ 143 $\frac{119}{119}$ $\frac{1406}{1408}$ 0.0 239 $\frac{235}{235}$ 357 $\frac{175}{175}$ $\frac{179}{179}$ $\frac{121}{121}$ $\frac{143}{145}$ $\frac{1420}{1422}$ $\frac{141}{141}$ 141 0.0 $\frac{179}{179}$ $\frac{241}{241}$ $\frac{245}{245}$ 181 $\frac{147}{149}$ 183 0.0 $\frac{1476}{1478}$ 0.0 373 $\frac{245}{245}$ $\frac{247}{247}$ $\frac{251}{251}$ $\frac{375}{375}$ $\frac{251}{251}$ 373 (0)



## MULTIPLE-CIRCUIT, DOUBLE WINDINGS, FOR DRUM ARMATURES.

>	ORS	FRONT AND BACK PITCHES												ORS		
OF CONDUCTORS		J.M. GOALL	4		6	8		10		12		14		16		DOG
ENT	OON	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	Po	LES	NOO
CE PER PER PER PER PER PER PER PER PER PE	No.OF	F	В	F	В	F	В	F	В	F	В	F	В	F	В	No. OF CONDUCTORS
@	1502	373	377	249	253	185	189	149	153	123	127	105	109	91	95	1502
00	1504	373	377	249	253	185	189	149	153	123	127	105	109	91	95	1504
00	1506 1508	375 375	379 379	249	253 253	187	191 191	149 149	153 153	123 123	127 127	105	109	93	97	1506
CD	1510	375	379	249	253	187	191	149	153	123	127	105	109	93	97	1508 1510
00	1512	375	379	249	253	187	191	149	153	123	127	105	109	93	97	1512
00	1514	377	381	251	255	187	191	149	153	125	129	107	111	93	97	1514
00	1516	377	381	251	255	187	191	149	153	125	129	107	111	93	97	1516
0	1518	377	381	251	255	187	191	149	153	125	129	107	111	93	97	1518
(0)	1520 1522	377 379	381	251 251	255 255	187 189	191 193	149 151	153 155	$\frac{125}{125}$	129 129	107	111	93	97	1520
00	1524	379	383	251	255	189	193	151	155	125	129	107	111	93	97	1522 1524
0	1526	379	383	253	257	189	193	151	155	125	129	107	111	93	97	1526
00	1528	379	383	253	257	189	193	151	155	125	129	107	111	93	97	1528
(0)	1530	381	385	253	257	189	193	151	155	125	129	107	111	93	97	1530
00	1532	381	385	253	257	189	193	151	155	125	129	107	111	93	97_	1532
00	1534 1536	381	385 385	253 253	257 257	189 189	193 193	151	155	125	129	107	111	93	97	1534
(0)	1538	383	387	255	259	191	195	151 151	155 155	125 127	129 131	107	111	93	97	1536 1538
00	1540	383	387	255	259	191	195	151	155	127	131	107	111	95	99	1540
@	1542	383	387	255	259	191	195	153	157	127	131	109	113	95	99	1542
00	1544	383	387	255	259	191	195	153	157	127	131	109	113	95	99	1544
00	1546	385	389	255	259	191	195	153	157	127	131	109	113	95	99	1546
(0)	1548 1550	385 385	389 389	255 257	259 261	191 191	195 195	153	157 157	127	131	109	113	95	99	1548
00	1552	385	389	257	261	191	195	153	157	127 127	131 131	109	113	95 95	99	1550 1552
0	1554	387	391	257	261	193	197	153	157	127	131	109	113	95	99	1554
00	1556	387	391	257	261	193	197	153	157	127	131	109	113	95	99	1556
@	1558	387	391	257	261	193	197	153	157	127	131	109	113	95	99	1558
00	1560	387	391	257	261	193	197	153	157	127	131	109	113	95	99	1560
00	1562 1564	389	393	259 259	263 263	193 193	197 197	155	159	129	133	109	113	95	99	1562
@	1566	389	393	259	263	193	197	155 155	159 159	129 129	133 133	109	113 113	95 95	99	1564 1566
00	1568	389	393	259	263	193	197	155	159	129	133	109	113	95	99	1568
0	1570	391	395	259	263	195	199	155	159	129	133	111	115	97	101	1570
00	1572	391	395	259	263	195	199	155	159	129	133	111	115	97	101	1572
00	1574	391	395	261	265	195	199	155	159	129	133	111	115	97	101	1574
(3)	1576 1578	393	395 397	261 261	265 265	195 195	199	155 155	159	129 129	133 133	111	115	97	101	1576
00	1580	393	397	261	265	195	199	155	159 159	129	133	111	115 115	97	101	1578 1580
0	1582	393	397	261	265	195	199	157	161	129	133	111	115	97	101	1582
00	1584	393	397	261	265	195	199	157	161	129	133	111	115	97	101	1584
(0)	1586	395	399	263	267	197	201	157	161	131	135	111	115	97	101	1586
00	1588 1590	395	399	263	267	197	201	157	161	131	135	111	115	97	101	1588
00	1592	395	399	263 263	267	197	201	157 157	161	131	135	111	115	97	101	1590
@	1594	397	401	263	267	197	201	157	161	131 131	135 135	111	115	97	101	1592 1594
00	1596	397	401	263	267	197	201	157	161	131	135	111	115	97	101	1596
(0)	1598	397	401	265	269	197	201	157	161	131	135	113	117	97	101	1598
00	1600	397	401	265	269	197	201	157	161	131	135	113	117	97	101	1600
			Alexander	at star	P DOLL							THE RESERVE AND ADDRESS OF THE PERSON NAMED IN	at the same of	Cr. U.C.		

WINDING TABLES FOR MULTIPLE-CIRCUIT, TRIPLE WINDINGS FOR DRUM ARMATURES.



### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTORS RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F B F B F B F B F B F B F B é Z (00) 23 (00) 21 $\frac{216}{218}$ (QQ) $\frac{218}{220}$ (W) (M) 13 (QQ) 4] (00) O 19 W $\frac{21}{21}$ æ $\frac{246}{248}$ 21 29 15 (1) $\frac{248}{250}$ (22) œ (20) $\frac{23}{23}$ (00) (20) 21 23 (20) $\frac{270}{272}$ (90) 71 @ 23 $\frac{278}{280}$ 73 78 21 W $\frac{278}{280}$ $\frac{15}{15}$ œ (20) @ @ 75 77 (III) $\frac{298}{300}$ $\frac{25}{25}$ $\frac{71}{71}$ 15 ගුන



### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. FRONT AND BACK PITCHES CONDUCTORS CONDUCTORS RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F F F F B B B F B F B B F B 73 @ 79 41 19 $\frac{25}{25}$ (00) @ W 23 29 $\frac{25}{25}$ 57 43 19 23 (20) (22) 79 79 29 $\frac{27}{27}$ 23 @ $\frac{25}{25}$ 19 $\frac{21}{21}$ 25 w 27 27 $\frac{25}{25}$ Q 27 27 91 (00) (00) 27 25 39 33 W 33 $\frac{27}{27}$ 29 W (W) $\frac{27}{29}$ Q @ 43 45 $\frac{23}{23}$ 29 @ 29 25 $\frac{27}{27}$ (00) (00) $\frac{53}{53}$ Q

### MULTIPLE-CIRCUIT TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTOR RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES OF OF F B F F B F B F F F B B B B No. 71 (20) 71 @ $\frac{107}{107}$ 73 73 (00) 23 @ 73 (20) 75 51 29 (00) (00) 113 75 (20) 77 (20) 29 77 $\frac{25}{25}$ (20) (00) 79 @ 79 117 59 (00) 73 (00) 113 119 119 75 75 81 49 (00) $\frac{472}{474}$ (00) $\frac{121}{123}$ 77 77 (20) (00) (20) @ 79 79 (20) 39 (20) 127 12759 59 53 53 (22) $\frac{498}{500}$ 81 39 33

### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES REFENTRANCY POLES POLES POLES POLES POLES POLES POLES PP F F B B F F F F F B B B B B o'N 29 83 55 @ $\frac{127}{127}$ 89 55 $\frac{524}{526}$ $\frac{127}{129}$ 63 æ (1) @ $\frac{137}{137}$ $\frac{536}{538}$ 51 71 Q 139 71 $\frac{71}{71}$ $\frac{137}{137}$ 143 67 73 59 53 139 145 (00) 141 75 $\frac{45}{45}$ 141 (00) 149 101 55 55 (00) (DE) $\frac{145}{147}$ 5.7 57 $\frac{53}{53}$

### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES REENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F F B F B F F F F B В В В В No (20) 79 35 (00) (00) (22) (00) 75 75 (00) 55 75 (00) (W) 77 77 W (00) 79 Ш (00) (00) W (W) 171 87 71 53 45 51 $\frac{45}{45}$ (00) (00) œ 173 117 71 71 @ (00) 73 $\frac{175}{177}$ 113 119 89 53 (1) (20) (W) (W)



MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES.																	
75	TORS		FRONT AND BACK PITCHES														
RE-ENTRANCY	CONDUCTORS	4 POLES		6 POLES		8 POLES		10 POLES		12 POLES		14 POLES		16 POLES		No. OF CONDUCTORS	
E-EN																F CO	
	No.OF	F	В	F	В	F	В	F	В	F	В	F	В	F	В	0 2	
000	702 704	173 173	179	113 115	119 121	85 85	91	67	73 73	55 55	61	47	53	41	47	702	
000	706 708	173 173	179 179	115 115	121 121	85 85	91 91	67 67	73 73	55 55	61	47	53 53	41	47	706	
@	710 712	175 175	181 181	115 115	121 121	85 85	91 91	67 69	73 75	57 57	63 63	47	53 53	41 41	47	710 712	
000	714 716	175 175	181 181	115 117	121 123	87 87	93 93	69 69	75 75	57 57	63 63	47	58 55	41	47	714 716	
(gg)	718 720	177	183 183	117	123 123	87 87	93 98	69	75 75	57	63	49	55 55	41	47	718 720	
@	722 724	177	183 183	117	123 123	87 87	93	69	75 75	57	63	49	55	43	49	722 724	
000	726 728	179 179	185 185	117	123 125	87 87	93	69 69 -	75 75	57 57	63	49	55 55	43	49	726 728	
000	730 732	179 179	185 185	119 119	125 125	89 89	95 95	69 71	75 77	57 57	63 63	49	55 55	43	49	730	
@	734 736	181 181	187 187	119 119	125 125	89 89	95 95	71 71	77	59 59	65 65	49	55 55	43	49	734 736	
000	738 740	181 181	187 187	119 121	125 127	89 89	95 95	71 71	77	59 59	65	49	55 55	43	49	738 740	
000	742 744	183 183	189 189	121 121	127 127	89 89	95 95	71 71	77	59 59	65 65	49 51	55 57	43	49	742 744	
(B)	746 748	183 183	189 189	121 121	127 127	91 91	97 97	71 71	77	59 59	65 65	51 51	57 57	43 43	49	746 748	
000	750 752	185 185	191 191	121 123	127 129	91 91	97 97	71 73	77 79	59 59	65 65	51 51	57 57	43	49	750 752	
000	754 756	185 185	191 191	123 123	129 129	91 91	97 97	73 73	79 79	59 59	65 65	51 51	57 57	45	51	754 756	
(B)	758 760	187 187	193 193	123 123	129 129	91 91	97 97	73 73	79 79	61 61	67 67	51 51	57 57	45	51 51	758 760	
000	762 764	187 187	193 193	123 125	129 131	93	99 99	73 73	79 79	61	67 67	51 51	57 57	45 45	51	762 764	
000	766 768	189 189	195 195	125 125	131 131	93 93	99	73 73	79 79	61	67 67	51 51	57 57	45 45	51	766 768	
@	770 772	189 189	195 195	125 125	131	93 93	99	73 75	79 81	61	67 67	51 53	57, 59	45 45	51 51	770 772	
000	774 776	191 191	197 197	125 127	131 133	93 93	99 99	75 75	81 81	61	67 67	53 58	59 59	45	51	774 776	
000	778 780	191 191	197 197	127 127	133 133	95 95	101 101	75 75	81 81	61	67	58 53	59 59	45 45	51 51	778 780	
@	782 784	193 193	199 199	127 127	133 133	95 95	101	75 75	81	63	69	53 53	59 59	45 45	51	782 784	
000	786 788	193 193	199 199	127 129	133 135	95 95	101	75 75	81 81	63	69	53 53	59 59	47	53	786 788	
000	790 792	195 195	201	129 129	135 135	95 95	101	75 77	81 83	63	69	53 53	59 59	47	53 53	790 792	
(Q)	794 796	195 195	201	129 129	135 135	97 97	103	77	83 83	63	69	53	59 59	47	58 58	794 796	
000	798 800	197 197	203 203	129 131	135 137	97 97	103	77	83 83	63	69	58 55	59 61	47	53 53	798	

### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS CONDUCTORS FRONT AND BACK PITCHES RE-ENTRANCY POLES POLES POLES POLES POLES POLES POLES PO OF F F F B F F F B B B B B F B No. (W) 133 79 (QQ) $\frac{207}{207}$ (00) (00) 135 79 79 71 (00) 141 79 $\frac{211}{211}$ (20) 143 73 (00) (00) @ 87 73 57 55 848 (20) W 75 @ (00) (00) 219 $\frac{105}{105}$ (00) $\frac{147}{147}$ 83 89 69 75 59 (00) 113 113 (22) (00) 71 113 53 ® $\frac{107}{107}$ 113 (QQ) 71 (00) $\frac{115}{115}$ $\frac{115}{115}$ 71 53

### MULTIPLE-CIRCUIT, TRIPLE WINDINGS, FOR DRUM ARMATURES. CONDUCTORS FRONT AND BACK PITCHES CONDUCTORS REFENTRANCY POLES POLES POLES POLES POLES POLES POLES OF F F B F B F F B F B F B B B (00) 73 @ 111 @ $\frac{155}{157}$ $\frac{117}{117}$ 73 73 79 55 229 119 119 95 75 75 75 55 $\frac{157}{159}$ 113 119 $\frac{151}{153}$ (22) (00) 75 55 75 55 (00) $\frac{71}{71}$ (W) 77 77 71 71 @ 161 യ (00) 77 77 239 $\frac{245}{245}$ 159 165 117 73 (00) @ @ 73 73 $\frac{241}{241}$ 119 79 $\frac{159}{161}$ $\frac{101}{101}$ (22) 167 $\frac{121}{121}$ 101 73 (W) œ

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111 113

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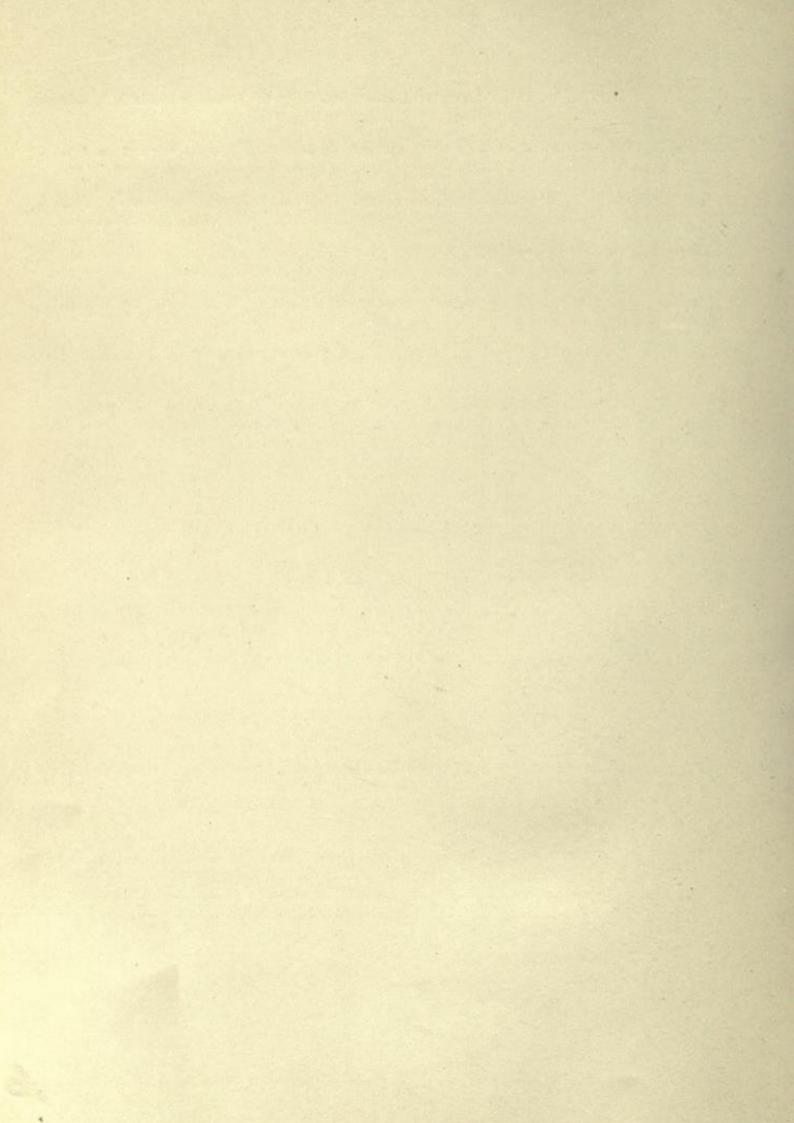
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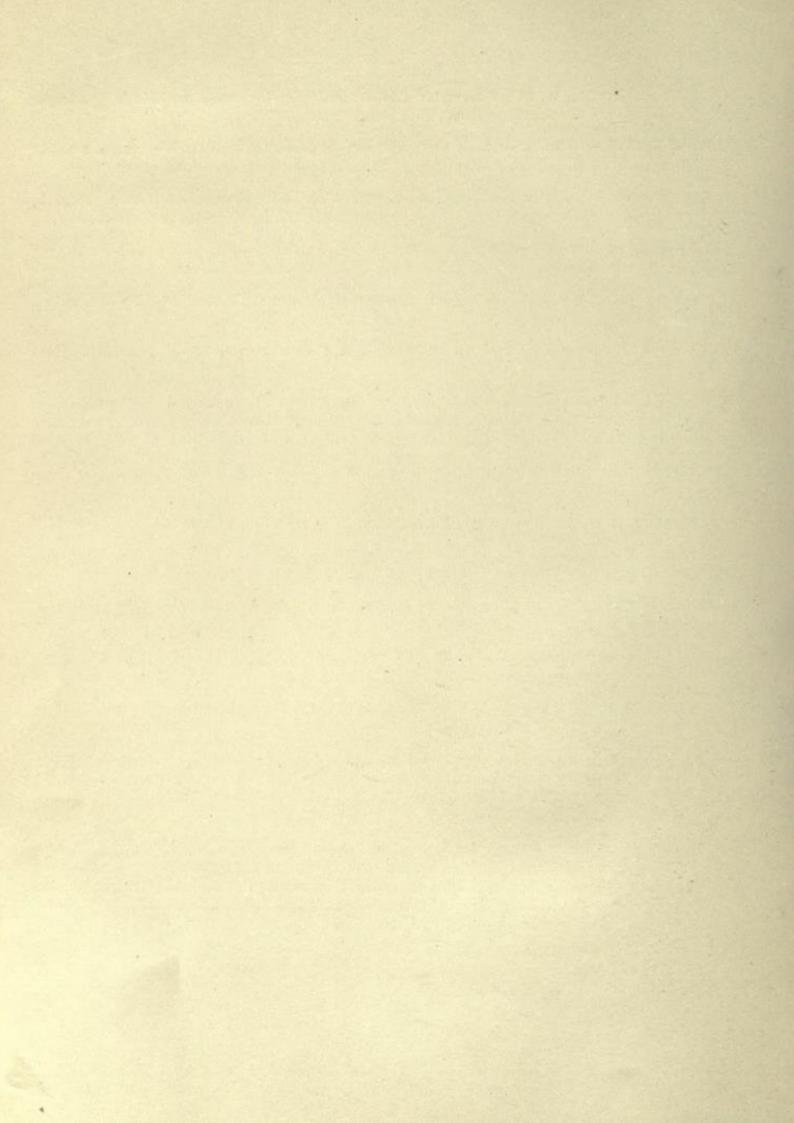
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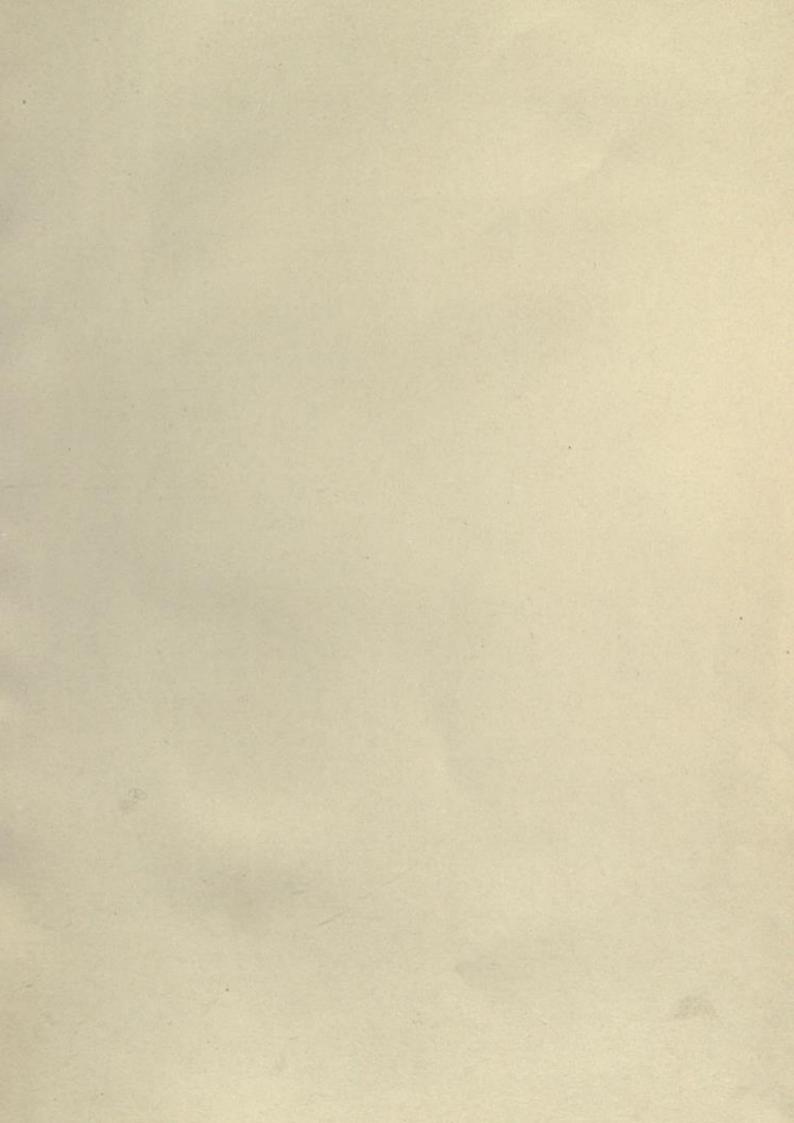
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